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ALTINBAS UNIVERSITY

Electrical and Computer Engineering

DESIGN OF FRACTAL BASED MONOPOLE ANTENNA ARRAY WITH ULTRA-MUTUAL COUPLING REDUCTION FOR MIMO APPLICATIONS

Zainab Jasim Mohammed AL-Dulaimi

Master's Thesis

Supervisor

Dr. Dogu Cagdas Atilla

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by

Zainab AL-Dulaimi

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Academic Title Name SURNAME	Academic Title Name SURNAME
Co-Supervisor	Supervisor

Examining Committee Members (first name belongs to the chairperson of the jury and the second name belongs to supervisor)

Academic Title Name SURNAME	Faculty, University	_____
Academic Title Name SURNAME	Faculty, University	_____
Academic Title Name SURNAME	Faculty, University	_____
Academic Title Name SURNAME	Faculty, University	_____
Assoc. Prof. Name SURNAME	Faculty, University	_____

I certify that this thesis satisfies all the requirements as a thesis for the degree of Master of Science.

_____

Academic Title Name SURNAME

Head of Department

Approval Date of Graduate School of Science and Engineering: ____/____/____

_____

Academic Title Name SURNAME

Director

DEDICATION

To the one and only the one who teach the human what he wasn't know , The bone of his praise, God Almighty sanctified His throne.

To the man who delivered the message, Advised the hall Nation, he is the Prophet of mercy and Lighting of the world . prophet Muhammad peace be upon him

To who is fighting his pain , resuming his struggles , In order to regain the position that he deserves , home for Islam , castle for The Arabs nation , Factory for Civilization , Shelter for dignity My beloved, homeland Iraq

To the one who harvested the thorn To pave me up The Way of Knowledge and Science The great heart that I am proud of being his daughter my father God prolong his age and Keep his healthy and comfortable .

He let me enjoy his generosity am replaying him for his favor The fruit is planted by his hands..

To who devoted her life to the performance of the message which was made from Patience paper and stitch it in the darkness of life , The candle that lit our way ,which drank the cup empty to feed us Drop of love My dear mother, I ask God to extend her life with good deeds

To the one which the life took him quickly and he couldn't see my achieving this success Absent from the world Always present in the heart My dear grandfather

may Allah have mercy on him To the bond from after the LORD my brothers and my sisters (Sara , Monther , Reem ,Zahraa ,Mohammed,Rahaf , al mouatasem belallah) i ask Allah for success in their ways

To symbol courage which they was Fell on the homeland With their pure blood My uncles the Martyrs (Ali , Hussein)

To a free man, but a prisoner My uncle (Salih) May Allah delight us By releasing him

To a beautiful heart and smile My dear uncle (Mohammed)Allah bless him and make him happy.

To the maker of glory And one of builders of the nation Green Spring cloud Which falls , And his generous hand Who did not spare me His effort any day My dear professor (Taha Ahmed Elwi)

To the oppressed people of the world. .To all those who stood with me and supported me in my affairs …

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ABSTRACT

DESIGN OF FRACTAL BASED MONOPOLE ANTENNA ARRAY WITH ULTRA-MUTUAL COUPLING REDUCTION FOR MIMO APPLICATIONS

AL-Dulaimi Zainab ,

M.Sc., Electrical and Computer Engineering, Altınbaş University,

Supervisor: Dr. Dogu Cagdas Atilla

Co-Supervisor: Taha A. Elwi

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In this thesis, a design of meander line monopole antenna array based on a Hilbert-Shaped Reject Band (HSRB) fractal of the third order is proposed for MIMO applications. The antenna array [32] is constructed from two monopole antennas of meander lines that are fed with a quarter wavelength (λ/4) micro-strip line transformer at 2.45GHz to suit the Wi-Fi applications. The antenna structure is printed on an FR4 substrate of 1mm thickness. The ground plane is defected with the proposed HSRB structure as mounted on the back profile of the substrate. The HSRB structure is proposed to reduce the mutual coupling between the proposed antenna elements within a limited area [32]. Moreover, the direction of the mender line is curved toward an opposite direction to increase the radiation diversity at the antenna far-field. Therefore, the maximum correlation factor due to the mutual coupling, surface wave effects, is found about 0.02 at 2.45GHz with a radiation diversity coefficient of 1. The proposed antenna design is fabricated to achieve excellent matching bandwidth, |S11|<-10dB, around 2.45GHz. The proposed antenna radiation pattern is found to be directed toward the array side away from the center with a gain of 2dBi. The maximum mutual coupling, given by |S12|, between the antenna elements [32] within a distance of 0.16λo at 2.45GHz is found bellow -30dB. Finite Integral Technique (FIT) of CST Microwave Studio (CSTMWS)[32] and Finite Element Method (FEM)

based HFSS simulations are invoked to evaluate the antenna performance. Finally, the measured results from the fabricated antenna array are agreed very well with simulations.

Keywords: MIMO, fractal, monopole, meander.

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LIST OF ABBREVIATIONS

MIMO : Multi input Multi output

EBG : Electromagnetic band gap

FIT : Finite Integral Technique

FEM : Finite Element Method

1. INTRODUCTION

Since 1950's, miniaturized scale strip reception apparatuses got an amazing consideration despite the fact that the primary models were showed up in the 1970's [32]. They were utilized in numerous telecom applications [32]: handheld gadgets, airplane, satellite, rocket, and so forth. The microstrip received wires show low profile, low assembling expense, precisely hearty, incorporate capable with PCB circuits and light pause. They are moderately adaptable as far as thunderous frequencies, polarization, example, transmission capacity and coordinating impedance. The span of the miniaturized scale strip reception apparatus moderately relies upon the wavelength at which the radio wire is full. Consequently, the scaling down of the reception apparatus measure is stile a test [32]. Utilizing substrate materials with high relative permittivity [32], shorting structures, snick way crisscross or winding shapes with opened geometries. Shorting pins or varactor diodes between the fix and the ground plane permit an extra tuning movement. Be that as it may, microstrip radio wires demonstrate a few impediments: Low taking care of intensity limit, low effectiveness, fake feed radiation, thin data transfer capacity, trouble of accomplishing unadulterated cross roundabout polarization, and moderately low increase [1].In this undertaking, the plan of a wearable microstrip received wire exhibit to be a well used on the human body is imagined for biomedical applications. The execution of the proposed radio wire cluster is tried in shut to the human body [32]. Since the dire interest of a wearable microstrip received wire that works with advanced properties, for example, fixable, low profile, less impact on the human body, low weight, simple mix with garments, a tremendous potential research is connected to wearable reception apparatuses and microwave gadgets. Normally received wire components should be divided by so as to have reasonable seclusion and low connection between's the components [32]. Be that as it may, debasement would result after setting the reception apparatus components in closeness, due to close handle impacts, diffraction from limited ground planes, and solid inductive and capacitive coupling between the components [2], [4].

1.2. PROBLEM STATEMENT

This ponder is conducted to optimize the proposed cluster parameters [32]. To begin with, the separate between the received wire components (S) is expanded to 32mm. The parametric think about is conducted on for cases as: No fractal [32], 0th emphasis, 1st emphasis, and 2nd cycle. In that matter, it is considered to screen the S-parameters as it were [32]. It is found the received wire components based 2nd emphasis give the finest coordinating, |S11|=- 25dB, [32] at 3.78GHz and 8.24GHz. In addition, the received wire cluster gives of the 2nd emphasis a shared coupling, |S12|<-30db, at 3.78GHz and 8.24GHz [32].

2. LITERATURE SURVEY

Single-negative attractive meta-materials are utilized in arrange to decrease common coupling between high-profile received wires utilized in multiple-input multiple-output frameworks. The attractive porousness of the created single-negative incorporations have negative viable reaction over a particular recurrence band. The considerations considered here are composed of broadside coupled split-ring resonators [5]. they report on the utilize of UC-PBG absconds on the ground plane with inclining spaces on the fix to essentially diminish the common coupling between received wire components inside a miniaturized profile radio wire cluster. The execution of the proposed cluster is inspected numerically utilizing CST MWS [6]. the proposed BCSRR decreases the common coupling to −15 dB between two patches with a partition of as it were 1/11th free-space wavelength for applications including Multi Input Multi Yield (MIMO) system [7]. They were examined the utilize of mandate divisions based on three patches to act as radiation chiefs to diminish the MIMO systems.

3. MIMO ANTENNA THEORY

3.1. INTRODUCTION

MIMO Innovation is a remote innovation. This window uses different transmitters and recipients to share additional information in the same times [23]. The MIMO innovation took advantage of the radio wave phenomenon, which indicated that it was multi-track as soon as the information transmitted to the dividers, ceilings and objects was reflected. The reception wire was accepted at different times by different points at different times in a marginal manner. MIMO innovates in multi-track behavior by victimizing differently, "intelligent."[37] Transmitters and receivers with a built-in "spatial"[35] measure to greatly increase execution and playback. MIMO frames allow different receiving wires to send and receive different spatial flows at the same time. MIMO frameworks work to change the work of radio wires more intelligently by imposing sanctions on them to mix data flows from different strategies Completely diverse} and at distinctive times to increase the control of capturing a collector signal. Wired radio wires are used for spatial differences.

3.2. PUBLIC INFORMATION ABOUT THE MIMO:

When we are using multiple antennas, MIMO has been able to innovate remotely expanding the channel's specific capacity. By increasing the number of corrupt and receiving wires transmitted, it is possible to increase channel productivity directly with each set of radio wires embedded in the frame. This makes MIMO wireless technology one of the most important wireless wireless technologies [25] that are subsequently employed for a long time. Because the speed of ghost transport has become an increasingly important product of radio-communication frameworks, the procedures required to use the transport capacity that can be accessed more successfully. MIMO-wireless-technology was one such procedure [12].

3.3 TYPES OF MIMO

3.3.1. SISO

The only shape of radio-link can be defined as SISO of Single-Input-Single-Output. This is often successfully a standard radio channel in which transmitters work with one radio wire as well as the collectors; where is no differences and no additional processing is required. The advantage of a SISO framework is the effortlessness which needs no handling in terms of the different shapes

of differing qualities which will be utilized. In any case, SISO-channel gives restricted execution as obstructions and blurring will affect the framework more than a MIMO framework utilizing a few frame of differing qualities. Productivity depends on channel bandwidth and signal-to-noise ratio.

3.3.2. SIMO

The SIMO or MIMO individual input occurs when the transmitter has one antenna and the collector has several radio wires; usually known as differences. It is used frequently to enable the recipient window Which receives signals from a number of free sources to combat the effects of uncertainty. It has been used several times with the synthesis / acceptance of concise waves to combat the effects of ionic decay and interference. SIMO has the advantage of achieving it moderately, although there are some flaws in the fact that preparation is required within the complex. Using a SIMO card may be very satisfactory in many applications, but when the assembly tool is found in a multi-use device such as a mobile phone, the preparation levels may be restricted by measuring, charging the battery, and depleting the battery. There are two types of SIMO scenarios that can be used:

1- Switched differences SIMO: This frame of SIMO looks for the most grounded flag and switches to that antenna.
2- Maximum proportion combining SIMO: This frame of SIMO takes both signals and entireties them to grant the combination. In this way, the signals from both received wires contribute to the generally flag.

3.3.3 MISO

Different Input Single Yield (MISO) is additionally named transmit differing qualities. In this case, the same information is transmitted needlessly from the two transmitter radio wires. The receiver is at that point able to get the optimum flag which it can at that point utilize to get extricate the desired data. The advantage of utilizing MISO is that the numerous radio wires and the repetition coding / preparing is moved from the recipient to the transmitter. In occasions such as cellphone UEs, this will be a critical advantage in terms of space for the received wires and decreasing the level of handling required within the collector for the repetition coding. This encompasses a positive effect on measure, taken a Number of dead and battery life where minimize the level of preparation requires less battery consumption.

3.3.4 MIMO

MIMO is viably a radio radio wire innovation as a result of it employments totally different received wires at the transmitter and recipient to empower a assortment of flag ways that to hold the knowledge, selecting isolated ways that for every radio wire to empower totally different flag ways that to be used. one in all the middle thoughts behind MIMO remote frameworks reference frame flag getting ready during which time is complemented with the abstraction activity inborn inside the utilize of various spatially sent received wires, i.e. the utilize of various received wires found at various focuses. suitably MIMO remote frameworks will be seen as a coherent enlargement to the keen received wires that are used for varied a protracted time to form strides remote. it's found between transmitters and collectors, the flag will take varied ways that. conjointly by moving the radio wires so a touch take away the ways that used can alter. The assortment approachs|of the way|of how} accessible happens as a results of the amount of objects that show up to the facet or so inside the coordinate way between the transmitter and collector. Already these varied ways that because it were served to gift impedances. By utilizing MIMO, these additional ways that will be used to advantage. they will be used to produce additional vigor to the radio connect by creating strides the flag to commotion proportion, or by increasing the connect data capability. the 2 elementary styles for MIMO ar given underneath:

1-Abstraction differing qualities: Spatial differing qualities used during this smaller sense often alludes to transmit and find differing qualities. These 2 techniques ar used to produce changes inside the flag to commotion proportion and that they ar characterised by progressing the unwavering quality of the framework with respect to the various shapes of blurring.

2-Spatial multiplexing: This form of MIMO is employed to produce additional data capability by utilizing the various ways that to hold additional activity, i.e., increasing the knowledge turnout capability. one in all the key focal points of MIMO abstraction multiplexing is that the truth that it's able to offer additional data capability. MIMO abstraction multiplexing accomplishes this by utilizing the many ways that and with success utilizing them as additional "channels" to hold data. the foremost extreme add of knowledge that may be carried by a radio channels restricted by the physical boundaries characterised below Shannon's Law. Multiple-input, multiple-output (MIMO) received wire frameworks ar used in gift day remote tips, count in IEEE 802.11n, 3GPP LTE, and moveable Wi-MAX frameworks. the tactic underpins improved data turnout so below conditions of impedances, flag blurring, and multipath. The

request for higher data rates over longer separations has been one in all the essential inspirations behind the advancement of MIMO orthogonal- frequency-division-multi (OFDM) communications systems. Shannon's law defines the utmost rate at that error free knowledge will be transmitted over a given information measure within the presence of noise. it's typically expressed within the form:

$$C=(BW)\log2(1+SNR) \quad (3.1)$$

where, C is that the data rate in bits per second, bioattack is that the information measure in Hertz, and SNR is Signal to Noise quantitative relation. The over condition seems associate degree increment during a channel SNR comes regarding in stripped picks up in channel output. As a result, the traditional thanks to attain higher info rates is by increasing the flag transmission capability. Shockingly, increasing the flag transfer speed of a communications channel by increasing the image rate of a balanced carrier increments its unprotectedness to multipath blurring. For wide transfer speed channels, one halfway arrangement to fathoming the multipath challenge is to utilize a arrangement of narrowband covering subcarriers. Not because it were will the utilize of covering OFDM subcarriers progress ghastly effectiveness, however the lower image rates utilised by narrowband subcarriers decreases the have an effect on of multipath flag things. MIMO communications channels offer associate degree curiously arrangement to the multipath challenge by requiring totally different flag ways in which. In impact, MIMO frameworks utilize a mixture of various received wires and diverse flag ways in which to select up info of the communications channel. By utilizing the abstraction activity of a communications connect, MIMO frameworks will accomplish basically higher info rates than typical single-input, single-output (SISO) channels. during a 2×2 MIMO framework, signals engender on other ways from the transmitter to the collector received wires. Utilizing this channel info, a collector will recoup free streams from every of the transmitter's radio wires. A 2×2 MIMO framework produces 2 abstraction streams to with success twofold the foremost extreme info rate of what might be accomplished during a typical 1×1 SISO communications channel. the best data rate of a MIMO framework, the data rate may be assessed as a piece of N abstraction streams. A basic approximation of MIMO data rate could be a perform of abstraction streams, bandwidth, and signal-to-noise (SNR) and is shown within the following:

$$C= N(BW)\log2\ (1+SNR) \quad (3.2)$$

Given the condition for MIMO channel capacity, it is conceivable to explore the relationship between the number of spatial streams and the throughput of different usage of SISO and MIMO setups. As an case, the IEEE 802.11g specs endorse that a wireless-local may be a organize (WLAN) channel employments a SISO setup. With this standard, the most extreme coded information rate of 54 Mb/s requires utilize of a 64-QAM tweak plot and a code rate of 3/4. As a result, the un-coded bit rate is 72 Mb/s (4/3 x 54 Mb/s). With least transmitter blunder vector size (EVM) at -25 dB, an SNR of 25 dB can be evaluated as the prerequisite for a 64-state quadrature-amplitude-modulation (64QAM) plot. Whereas EVM and SNR are not proportionate in all cases, we are able accept that the greatness mistake of a image will rule the flag blunder as the SNR approaches its lower restrain. The maximum data rate of IEEE 802.11g maps closely with the maximum channel capacity dictated by the Shannon- Hartley theorem. According to this theorem, a Gaussian channel with an SNR of 25 dB should produce an un-coded data rate of 94 Mb/s in a 20-MHz channel bandwidth. By differentiate, Eq.3.2 would recommend that a MIMO channel with four spatial streams ought to be able of four times the capacity of the SISO channel. 20-MHz channel with a signal-to-noise proportion (SNR) of 25 dB and four spatial streams ought to have an un-coded bit rate of 4 x 94 Mb/s = 376 Mb/s. This estimation maps closely with the anticipated information rates of the draft IEEE 802.11n physical layer specs. IEEE 802.11n is planned to back MIMO setups with as numerous as four spatial streams. At the most noteworthy information rate, bursts employing a 64QAM tweak conspire with a 5/6 channel code rate deliver a information rate of 288.98 Mb/s and an un-coded bit rate of 346.68 Mb/s. At the most noteworthy information rate, the IEEE 802.11n channel with four spatial streams produces a information rate that's comparable to the hypothetical constrain of 376 Mb/s. It can be watched that the bit rate of a 4 x 4 (four spatial stream) MIMO setup surpasses that of the Shannon- Hartley constrain at all information rates, making MIMO frameworks alluring for higher information throughput. Whereas MIMO frameworks give clients with clear benefits at the application level, the plan and test of MIMO gadgets isn't without critical challenges.

3.4. BENEFITS OF MIMO TECHNOLOGY

A) Different wired wire settings can be used to overcome the effects that impede multiple paths and blurring when attempting to achieve the speed of transmission of long information in limited bandwidth channels. Multi-input and multiple-output (MIMO) radio frames are used in advanced remote instructions and are computed in IEEE 802.11n, 3GPP LTE and Wi-MAX Portable

frameworks. The strategy is based on already improved information productivity under constraints, multiple paths, and uncertainty. The demand for higher information rates over longer intervals was a key inspiration behind the improved OFDM frames. A - Gives prevailing information rates, expansion and constant quality using various receiving wires in transmitters and collectors without the need for additional transfer capacity or movable control. This is usually achieved by creating multiple free channels to send different streams of information. These flows can be combined by shaping the active digital beam and setting up the MIMO receiver (inside the oval oval) to expand the quality and operate it. The number of free channels and related information streams that can be supported via the MIMO channel corresponds to the minimum number of wires received at the transmitter or collector. Thus, the 2x2 frame can reinforce at most two flows, the 3x3 frame can support three flows and the 4x4 frame can support four flows. A few free flows can be combined by forming the active tape and setting up the MIMO collector, as shown in the Rodi Oval, which comes with continuous quality and operation.

3.5. LTE MIMO CONCEPTS

MIMO systems form an essential part of LTE in order to achieve the ambitious requirements for throughput and spectral efficiency. MIMO refers to the use of multiple antennas at transmitter and receiver side.

3.5.1 Downlink MIMO

For the LTE downlink, a 2×2 setup for MIMO is expected as standard arrangement, i.e. 2 transmit received wires at the base station and 2 get received wires at the terminal side. Arrangements with 4 radio wires are moreover being considered. Diverse MIMO modes are conceived. It has got to be separated between spatial multiplexing and transmit differences, and it depends on the channel condition which conspire to choose.

3.5.1.1 Spatial multiplexing

Spatial multiplexing permits transmitting diverse streams of information at the same time on the same downlink asset block(s). These information streams may has a place to a single client (single client MIMO / SU-MIMO) or distinctive clients (multi client MIMO / MU-MIMO). Whereas SU-MIMO increments the information rate of one client, MUMIMO permits expanding the in general capacity. Spatial multiplexing is as it were conceivable on the off chance that the

portable radio channel permits it. Each transmit received wire transmits a distinctive information stream. Each get received wire may get the information streams from all transmit radio wires.

3.5.1.2 Transmit diversity

Rather than expanding information rate or capacity, MIMO can be utilized to misuse differing qualities. Transmit differing qualities plans are as of now known from WCDMA discharge 99 and will moreover shape portion of LTE as one MIMO mode. In case the channel conditions don't permit spatial multiplexing, a transmit differences conspire will be utilized instep, so exchanging between these two MIMO modes is conceivable depending on channel conditions. Transmit differences is utilized when the chosen number of streams (rank) is one.

3.5.2 Uplink MIMO

Uplink MIMO plans for LTE will vary from downlink MIMO plans to require into consideration terminal complexity issues. For the uplink, MU-MIMO can be utilized. Different client terminals may transmit at the same time on the same asset piece. Usually too alluded to as spatial domain multiple get to (SDMA). The plot requires as it were one transmit radio wire at UE side which may be a huge advantage. The UEs sharing the same asset piece need to apply commonly orthogonal pilot designs. To abuse the good thing about two or more transmit received wires but still keep the UE fetched low, antenna subset choice can be utilized. Within the starting, this strategy will be utilized, e.g. a UE will have two transmit radio wires but as it were one transmits chain and intensifier. A switch will at that point select the radio wire that gives the finest channel to the e-Node-B.

3.5.3 Antenna Design for MIMO Applications

3.5.3.1 Information about design

Electromagnetically printed radio wires are created to supply each wideband impedance characteristics. Numerous parameters optimize the impedance transmission capacity of this antenna which has got to be examined. These received wires are built up for advanced wideband remote applications like versatile phones or Wireless LAN, Bluetooth, UWB and RFID innovations. As said within the past chapter, MIMO frameworks perform best when they can reply to the issues related to received wire hypothesis such as cluster setup, radiation design, sort of polarization and common coupling.

To discover out the right plan and setup of the MIMO received wire, it's imperative to fulfill the necessities concerning its last remote application. Be that as it may, it is worthy to characterize a few basic properties that must be affirmed to ensure a good performance and to function within the best conceivable way. These necessities must be taken under consideration to optimize the received wire execution. In any case, these characteristics are not free from each other. In brief:

- Size: The measure (volume) of the received wire and its generally affect on the encompassing environment is amazingly imperative for most remote communication frameworks. Likely the greatest issue with utilizing little received wires for remote communications is the decrease in effectiveness.
- Efficiency: The greater the efficiency, the better the link budget.
- Bandwidth: The designed antenna must satisfy the bandwidth requirements for the wireless system.

The gain bandwidth and return loss bandwidth (frequency range in which the return loss is better than -10dB) must be satisfied.

- Polarization: in order to reduce the multipath fading and probability of error and to increase the channel capacity.
- Power Handling: required to characterize the materials required for the received wire to fulfill its application. Considering these necessities and including the require for low-cost arrangements, printed received wires show up to be the leading choice.

However, by appearing in many forms, the best suited for a specific application may not be clear.

3.5.3.2 Printed antennas

Printed Radio wires, in its most fundamental shape, comprise of a transmitting fix on one side of a dielectric substrate which regularly includes a ground plane on the back side. Due to its reasonable, direct creation, and small measurements, composed received wires are clear to coordinated in portable terminals. For great execution, a thick dielectric substrate with a moo dielectric permittivity is appropriate by giving way better productivity and bigger transfer speed. In spite of the fact that the received wire directivity is autonomous of the substrate thickness, the received wire effectiveness and transmission capacity execution depends on the dielectric permittivity of the substrate. There are three main sorts of level profile printed radio wires because it is appeared on the figure 3.1([19]).

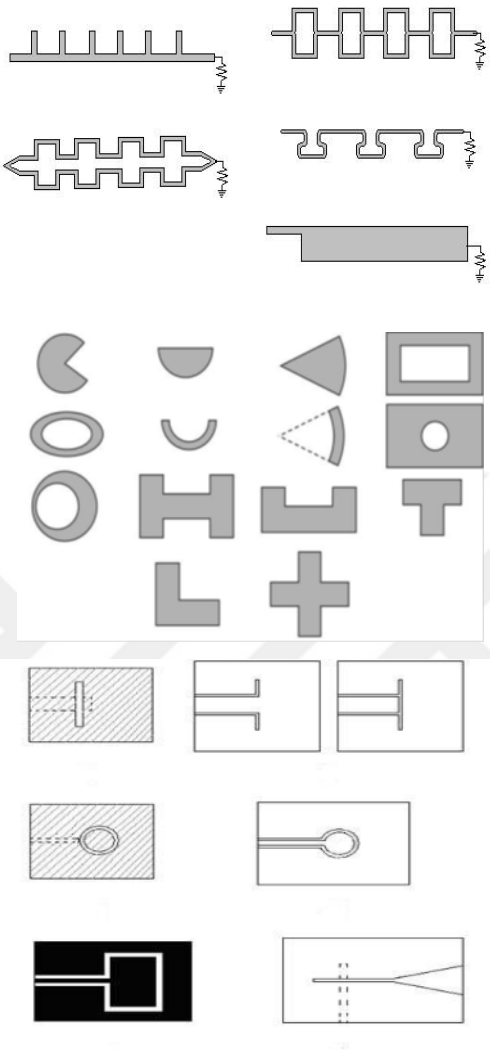

Figure 3.1: Printed Antenna Shapes (a) The travelled antenna's wave (b) Patch Antennas, and (c) Printed Slot Antenna.

All of these received wire sorts have a lean profile and are able to function in more than a single recurrence. In expansion to having a tall execution, micro-strip fix radio wires have the most effortless way of creation (can be fabricated in huge amounts), bolster both straight and circular polarizations and are able to be created in any kind of shape.

3.5.4 Printed Patch Antennas

Among fix radio wires there are distinctive sorts of nourishing procedures for printed received wires, two of which stand out among others: Micro-strip Line Nourish, 8(a) (the micro-strip line and ground plane, made of the same conductor fabric are put on inverse sides, which may have

an discuss crevice between the ground plane and substrate) and Coplanar Waveguide (CPW), (b), (contains a single conductive metallic layer on the substrate that incorporates the radiator and ground plane).

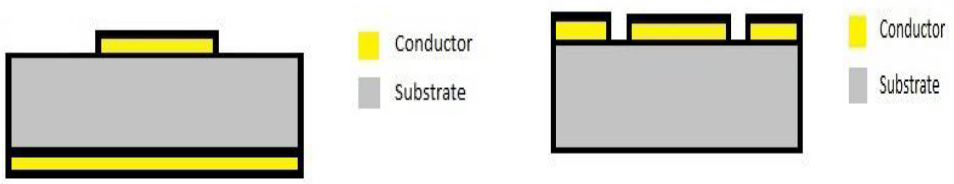

Figure 3.2: Feeding Line Techniques (a) Micro-strip Line and (b) Coplanar Waveguide (CPW).

These topologies impact the antenna's execution and the sort of bolster strategy must be chosen concurring to its application. On the table 1 there are a few contrasts that can be commented between CPW and Micro-strip line:

Table 3.1: Micro-strip line and CPW feeding types [9,10]

Feature	Microstrip Line	Coplanar Waveguide (CPW)
Dispersion	High	Low
Losses	Low	High
Coupling	High	Low
Design Flexibility	Low	High
Circuit Size	Large	Small

3.5.4.1 MIMO antenna solutions

In [11], a few contemplations concerning MIMO Radio wire plan are set up in arrange to optimize viewpoints such as cluster arrangement, radiation design, sort of polarization and shared coupling. This paper proposes distinctive concepts and arrangements for MIMO frameworks such as Received wire cluster arrangement and reconfigurable radio wires. Received wire cluster or staged cluster framework comprises on a set of fix radio wires with diverse formats. The cluster topology is chosen in arrange to maximize capacity and minimize blunder rate. In MIMO clusters, the relationship between the numerous signals must be as slightest as conceivable to neutralize the situation of corruption in channel capacity. Pick up improvement can be accomplished by utilizing a few differing qualities techniques:

1- Spatial Differences: distinctive components are dispersed with ideal separate to extend the number of channels within the interface. In this procedure, the littler the separate, the more the shared coupling between radio wires, that result a reduction of the channel capacity.
2- Polarization Differences: components within the cluster are encouraged with in an unexpected way polarized signals.
3- Pattern Differing qualities: the signals with different points are given to each one of the radio wires present within the array. MIMO received wire frameworks can be utilized in arrange to realize distinctive objectives such as expanding the by and large pick up, cancelling out the obstructions from a specific set of bearings and maximizing the Flag to Impedances Beat Commotion Proportion (SINR) - to set up the most extreme constrain concerning to channel capacity.

3.5.4.2 Standard MIMO configurations

MIMO radio frameworks utilize different received wires in arrange to send and get numerous information streams at once. The number of radio wires required is characterized by the radio producer based on what they decide will work for ideal transmission and gathering with their specific equipment and computer program. Normal setups are:

•2×2 MIMO (two transmit antennas, two receive antennas)

•3×3 MIMO (three transmit antennas, three receive antennas)

•4×4 MIMO (four transmit antennas, four receive antennas)

•8×8 MIMO (eight transmit antennas, eight receive antennas)

For the most part talking, the more radio wires a framework has the more concurrent information streams can be transmitted at once, making strides the radio connect. Be that as it may person framework setups, current physical and RF natural conditions, and progresses in radio innovation implies that more radio wires doesn’t continuously break even with superior framework execution.

3.5.4.3 Antenna choices for MIMO radios

There are numerous received wire choices for MIMO radios, depending on how the client wishes to design their radio framework based on the one of a kind needs of their application. Broadly talking, received wires for MIMO utilize can be broken down the taking after categories:

Traditional vertically polarized dipole received wires. These are the same radio wires utilized on SISO frameworks, the as it were distinction being that the MIMO framework will utilize two or more of the same received wire. In hone nearly any omni-directional radio wire can be utilized as long because it meets the recurrence band prerequisites, pick up, RF control taking care of, and other parameters for the framework being indicated.

A. Multi-antenna items contained inside a single radome. These radio wires contain two or more RF connectors, which each interface to a diverse radio wire inside the radome. These are frequently outlined with a incline cleared out and incline right polarization, with each component balanced 90 degrees (orthogonally) from the other, giving extra received wire polarization differing qualities for active RF situations without expanding framework establishment complexity for the conclusion client.

Each item highlights numerous RF connectors, each associated to a discrete received wire inside the product's arbitrary. These are frequently cross-polarized for extra polarization and spatial differing qualities. Fair as with SISO frameworks, MIMO frameworks can utilize omni-directional radio wires or directional boards and division radio wires for distinctive communication scenarios and scope needs.

3.6 ANTENNA POLARIZATION CHOICES

MIMO radio frameworks can take advantage of different sorts of received wire polarization plans to progress differences, which is one of the key ways MIMO frameworks are able to supply strong network indeed challenging situations that would demonstrate troublesome for single received wire radio systems. All received wires have a particular polarization heading, which is decided by their plan and speaks to the wavering course of the electromagnetic radio waves as they engender from the antenna's emanating component. The 'electric' parcel of the electromagnetic wave is as it were half of the transmitted flag, with a attractive wave that wavers at a 90 degree point at the inverse timing of the electrical wave. As the electrical cycle .wave rises the attractive wave falls, and vise-versa for each. Polarization sorts can be broken down into two key sorts: "linear polarization" and "circular polarization":

1- Linear polarization happens in a straight line, and can be vertical, even, or at any point such as 45 degrees. The electrical wave of the antenna's flag sways up and down along the hub of this straight line.
2- Circular polarization, rather than happening on a single plane, pivots because it takes off the radio wire. Envision a winding corkscrew transmitting out of the radio wire. Circular polarization can either pivot cleared out (counter-clockwise) or right (clockwise). One transformation is completed for each wavelength of the transmission. Radio wires that are circularly polarized are frequently superior suited for working in harsh climate conditions, as they can more effortlessly pass through rain and other climatic unsettling influences compared to straight polarized radio wires.

The choice of received wire polarization is imperative since transmit and get received wires got to be combined by coordinated polarization sort. A vertically polarized received wire works best with another vertically polarized radio wire, and circularly polarized radio wires will work best with other circularly polarized received wires. In the event that a polarization jumble happens, a misfortune of pick up will be presented lessening communication remove and affecting the quality of video or information transmission. Whereas polarization coordinating on both sides of the RF connect is still imperative with MIMO frameworks, since there are two or more sets of received wires utilized with each radio clients can arrange inverse polarizations with each set of radio wires on the radio to present cross polarized segregation between the numerous information streams being transmitted by the radio, expanding differing qualities and progressing transmission quality. For case, a 2×2 MIMO framework can be arranged to utilize two received wires with incline 45 degree polarizations. Setting two received wires at inverse 45 degree points gives a 90 degree orthogonal polarization between the two transmit radio wires, giving each radio wire 30 dB of separation from one another, in expansion to confining those streams from other vertically polarized broadcasts within the region which will meddled with the transmission in spite of not sharing the same recurrence band. On the get side, the received wires are designed to match the same 45 degree polarization in arrange to get the flag without issue. By segregating each information stream, blurring and other quality issues can be relieved.

3.7 ANTENNA SELECTION FOR MIMO SYSTEMS

Each MIMO framework will be arranged in an unexpected way based on the specific needs of the clients working the framework, what sort of information should be transmitted, and what environment the framework will be worked in. Be that as it may a few generalized recommendations for diverse working scenarios are sketched out underneath:

3.7.1 Hand-Held or Body-Worn Tactical MIMO Radio Systems

For clients who will be wearing the radio in a pocket or vest, or carrying the radio in conjunction with other gear (such as a tv camera), light weight omni-directional received wires are an perfect choice. The 360 degree scope of the radio wire permits for a organize association no matter which course the client is confronting, and these received wires are generally little and offer assistance keep the by and large weight of the complete unit to a sensible level. Radio wires with built-in spring bases or coordinates RF coaxial goosenecks are well known choices for body-worn radio frameworks. Another choice that can be utilized body worn received wires. These regularly center the antenna's bar in a specific course, giving improved directional scope and an increment in pick up over omni-directional received wires. These are frequently circularly polarized and worn in a MOLLE pocket on strategic vests by clients on both the front and back of the body, giving total scope around the client.

3.7.2 Vehicle Mounted MIMO Radios

For portable frameworks introduced into vehicles both omni-directional and directional received wires effortlessly be utilized depending on wants of the framework and can be mounted inside or remotely at the caution of the client. Numerous omni-directional radio wires are made to be specifically introduced onto the roof of the vehicle, making a changeless framework. Transitory establishments can moreover be effortlessly made by utilizing attractive mounting packs, permitting for vehicles to be rapidly outfitted for particular needs without the bother of changeless establishment. A solid magnet holds the radio wire in put whereas the vehicle is in movement. For UAVs, airplanes, and helicopters, received wires with extraordinary mounting frameworks are accessible, such as flight-rated STC mounts that are secured to the airframe. Blade-shaped received wires and indeed internally-mounted received wires are choices as well.

3.7.3 Directional and Sector Infrastructure Antennas

For MIMO scope of particular ranges, board and segment radio wires are regularly utilized. The tight beam-width and tall pick up permits for more exact directional arrange scope of particular

zones. These can be utilized as base stations or repeaters for huge scope zones, being encouraged by person MIMO radio hubs inside the organize. Establishment can be changeless for long-term utilize or done as a fast arrangement for particular occasions.

3. 8 THEORY OF MICRO-STRIP ANTENNAS

This area presents and examines a few of the fundamental concepts of micro-strip radio wires with the accentuation on clarifying the general features of fix radio wires at the side the utilize of different bolstering procedures. Moreover, it'll cover three well known strategies of examining micro-strip antennas such as transmission line show, depth demonstrate, and full wave show, all of which are displayed. There's a unfaltering increment in request for this smaller than expected gadget with the progresses of novel remote communication approaches. Within the setting of these improvements, compact measure, light weight, moo profile and moo fetched are the foremost critical components that architects ought to take into consideration for any remote versatile component. Micro-strip received wire or fix received wire has numerous focal points over the other radio wires, since of its moo profile, conformability, light weight, ease-of-production and moo fetched [14]. Numerous applications make utilize of micro-strip radio wires such as Worldwide Situating frameworks (GPS), inaccessible detecting, portable communication, wearable and implantable therapeutic applications, as well as air ship and shuttle applications. Be that as it may, in show disdain toward of their appealing highlights, micro-strip received wires have characteristic impediments, for case, their limit impedance transmission capacity, moo proficiency, destitute polarization immaculateness and moo power-handling capability. Numerous analysts have created and amplified the applications of micro-strip radio wires by tending to a few of the above-mentioned impediments [15].

3.8.1 General Structure of a Micro-strip Patch Antenna

By and large, the micro-strip received wire comprises of a level rectangular metal sheet or "fix" fitted over a greater sheet of metal known as a ground plane. Metal Fix and the ground plane are isolated by a lean dielectric substrate as appeared in Figure 3.2.

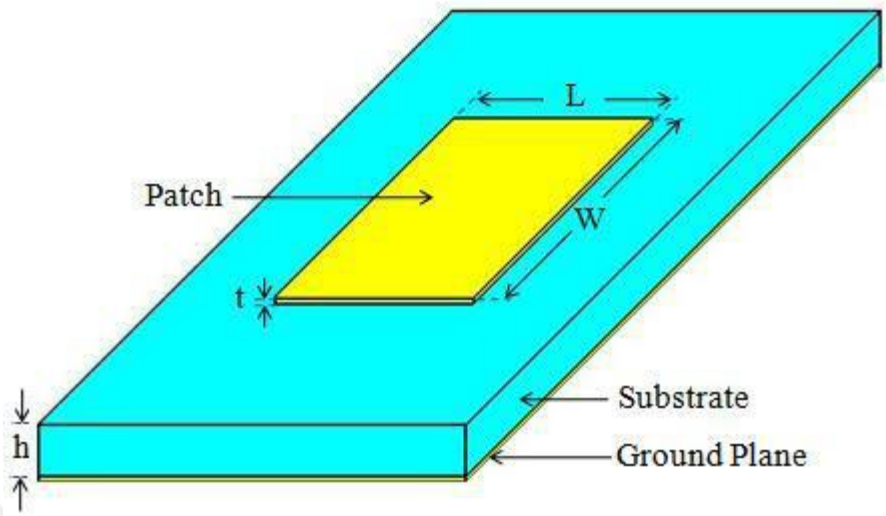

Figure 3.3: The structure of a rectangular micro-strip patch antenna [15].

The fix and the ground plane are more often than not of conducting materials such as gold and copper, which are carved utilizing different methods. Besides, the fix can be outlined of any shape (square, rectangular, dipole and circular). Other common fix shapes are appeared in Figure 3.3 [15]. It ought to be famous that the measurements and shapes of the fix, dielectric consistent of substrate, substrate thickness and nourish area decide the execution of the micro-strip radio wire, counting transmission capacity, effectiveness, pick up and resounding recurrence. Regularly, patch's length, L is around one half wavelength ($0.3333\ \lambda g < L < 0.5\ \lambda g$), where λg is the guided wavelength inside the substrate layer. The fix thickness is exceptionally lean in comparison with the λg ($t << \lambda g$). Various substrates have been utilized within the plan of micro-strip radio wires which can be either inflexible or adaptable. The stature, h of the substrate is ordinarily between $0.003\lambda g \leq h \leq 0.05\lambda g$. Another vital radio wire parameter which influences the received wire execution is the dielectric steady of the substrate, εr which is regularly $2.2 \leq \varepsilon r \leq 12$ [15].

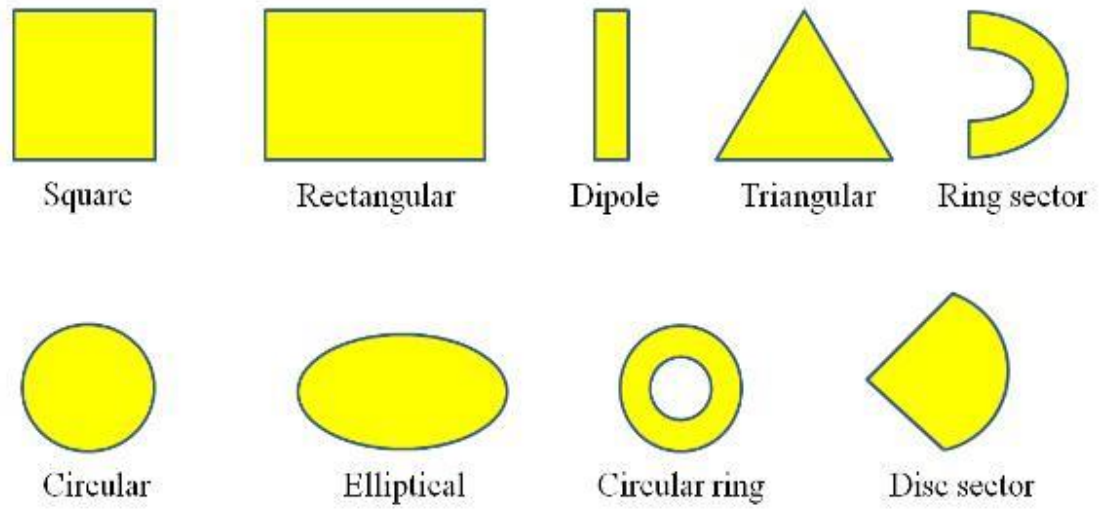

Figure 3.4: The common shapes of micro-strip patch antenna [15].

Micro-strip antenna's radiation emerges from the impacts of the bordering areas between the fix edge and the ground plane which at that point powers the vitality to spill into the discuss through the received wire substrate as demonstrated in Figure 3.3 [15]. To attain best conceivable execution of the micro-strip radio wires it is proposed that a thick substrate be utilized with a little dielectric consistent [26], which expanded the radio wire measure, but to plan a smaller than expected Micro-strip fix received wire, higher dielectric constants which lead to a diminish in antennae performance are required. Thus, there's a require for compromise between the details of the received wire and its execution.

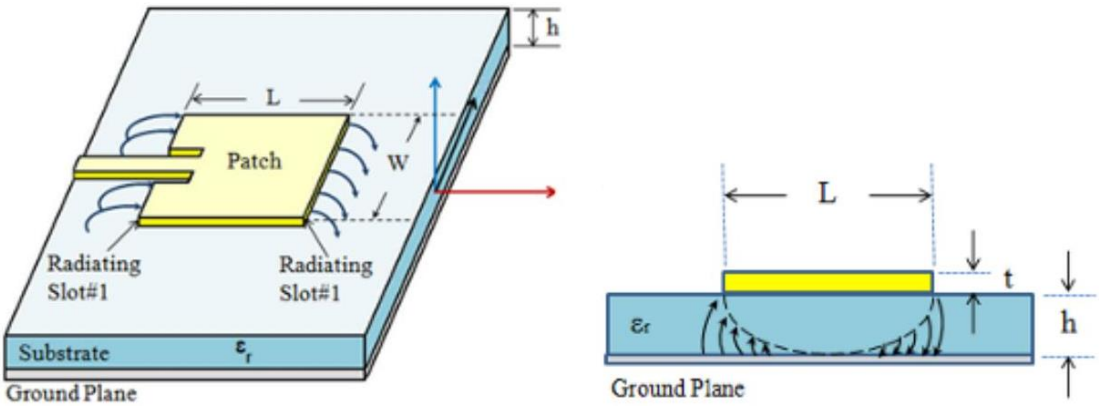

Figure 3.5: The rectangular micro-strip patch antenna with radiating slots.

Micro-strip radio wire works in its fundamental mode, when its bolster is found within the center of the patch's width, W, and some place in resonant length, L [17]. The precise position is built

up by the electromagnetic field dispersion within the fix received wire. The variety of the attractive field (current) and electric field (voltage) along the patch antenna is shown in Figure 2.4, which appears that the current incorporates a most elevated conceivable esteem at the fix center and most reduced esteem at the edges on the cleared out and right sides. In any case, the voltage is zero at the center of the fix, most elevated (positive) at the cleared out edge and least (negative) at the correct edge.

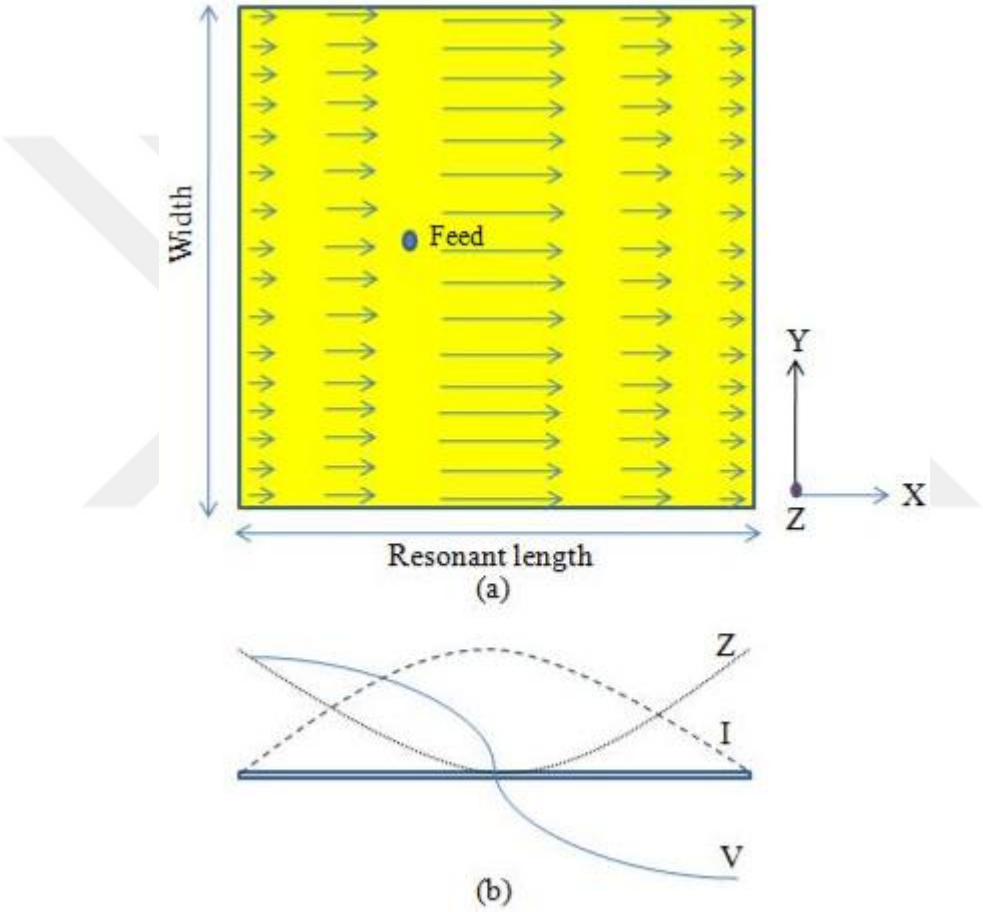

Figure 3.6: (a) The current distribution on the patch surface, (b) the voltage (*V*), the current (*I*) and impedance (|*Z*|) distribution along the patch's resonant length [17].

From the current, I, and voltage, V, the impedance, Z, can be determined, which is able be least (hypothetically zero)[17] at the middle of the fix and most noteworthy (hypothetically infinite) at the fix edges; 50 Ω may well be found some place along the received wire length where the radio wire has got to be encouraged[17].

3.8.2 Feeding Techniques

Numerous procedures have been proposed pointed at bolstering the RF control into micro-strip fix receiving wire. By and large, these approaches are gathered into two categories: reaching and non-contacting. Within the to begin with category [17], the RF control is nourished specifically through a interfacing component between the fix and the source such as a micro-strip line, whereas within the moment category; electromagnetic field coupling is utilized to exchange control between the micro-strip line and the transmitting fix [17]. Four strategies are broadly utilized: micro-strip line, coaxial test (both reaching category), aperture coupling and nearness coupling (both non-contacting category), which is able be clarified. The foremost critical figure for the nourishing plan is to attain most extreme coupling, which in turn leads to a most extreme figured it out pick up, or else wrong radiation, surface wave misfortunes, undesirable radiations and side flaps will be produced which would lower the figured it out pick up [18].

3.8.2.1 Coaxial probe feed

Coaxial bolstering as outlined in Figure 8.5 is the only micro-strip fix nourishing method. With this procedure the inward conductor of the coaxial is patched to the radiation fix of the received wire whereas the external conductor is associated to the ground plane [26]. The benefits are:

- Ease of fabrication,
- Ease of matching,
- Low level of false radiation.
- The limitations are:
- Narrow bandwidth,
- Difficulty of model especially for thick substrate since a hole has to be drilled in the substrate,
- With increasing substrate thickness, the input impedance becomes more inductive which leads to a matching problem.

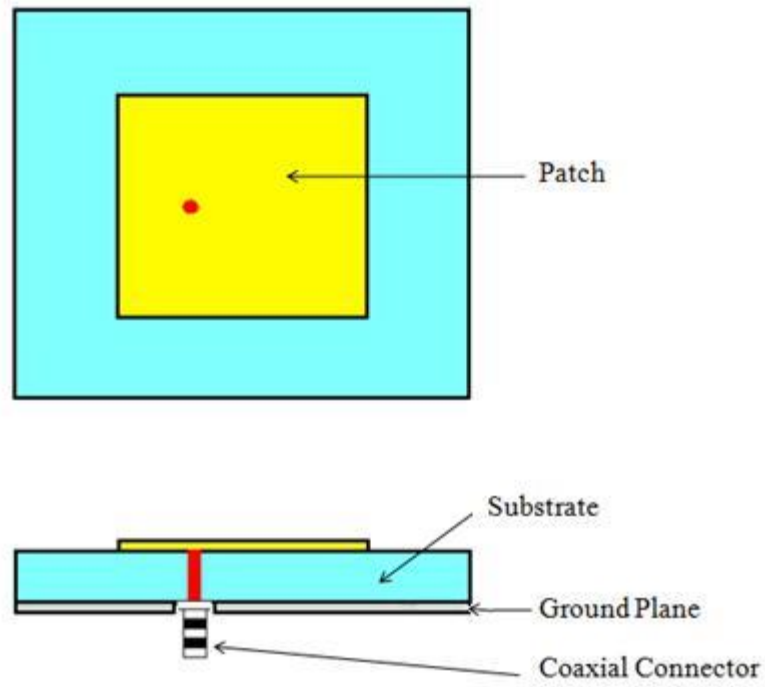

Figure 3.7: The probe fed rectangular micro-strip patch antenna [26].

3.8.2.2 Micro-strip line feed

Another straightforward method for micro-strip fix nourishing is the micro-strip line, which is specifically associated to the patch's edge on the same substrate to permit planar structure as outlined in Figure 3.7. By overseeing the inset position, coordinating the nourish line with the fix can be effortlessly done [26]. The points of interest are:

- Ease of fabrication,
- Simple to model,
- Impedance matching.
- The limitations are:
- Feed radiation,
- Thicker substrate results in increasing the surface waves which limit bandwidth of the antenna.

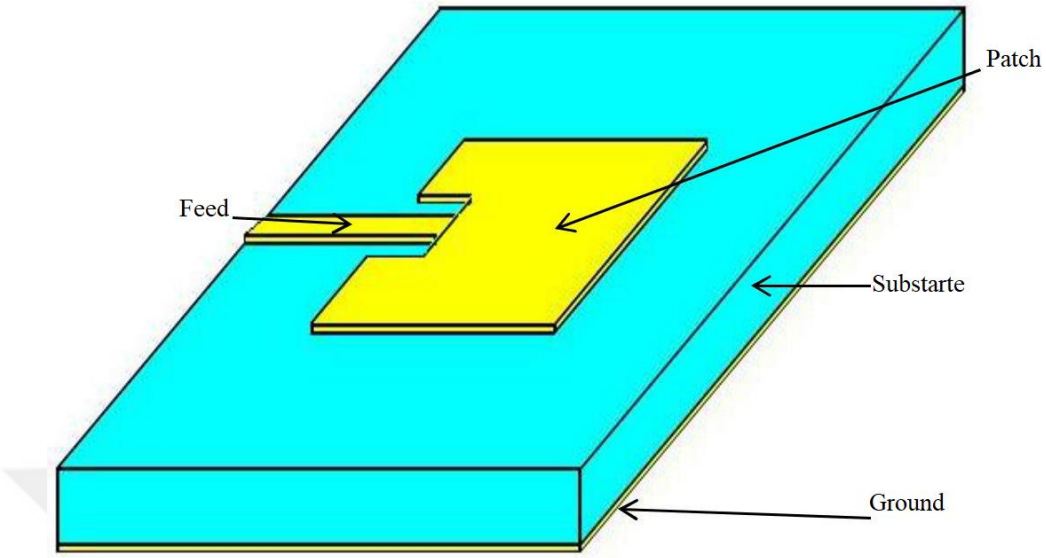

Figure 3.8: General view of the micro-strip line feed technique [26].

3.8.2.3 Proximity-coupling feed

A non-contacting strategy is talked about here. As outlined in Figure 3.8, a bolster line is set between two dielectric substrates and the emanating fix is on beat of the upper substrate [26]. The benefits are:

- Elimination of false feed radiation,
- Providing very high bandwidth (as high as 13%),
- Matching can be done easily by regulating the length of the feed line and the width-to-line ratio of the patch.
- The limitations are:
- Difficulty of fabrication due to the two dielectric layers which require proper alignment,
- Increase in the overall thickness of the antenna.

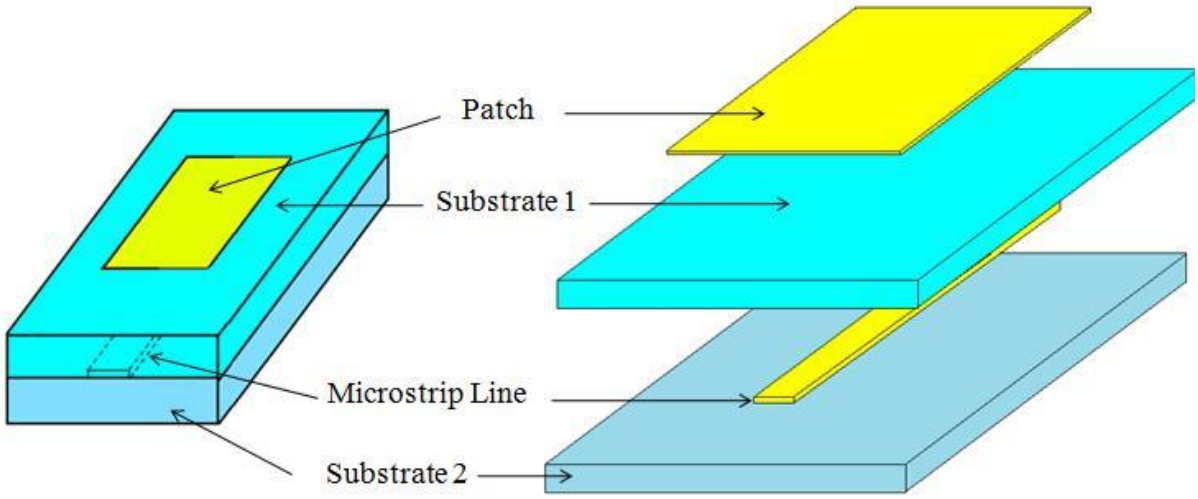

Figure 3.9: General view of the proxy-mity coupling feeding technique.

3.8.2.4 Aperture-coupling feed

Gap coupling nourish could be a non-contacting nourish strategy, in which the transmitting fix and the micro-strip nourish line are isolated by a ground plane as appeared in Figure 3.9 [26], and a tall dielectric fabric is utilized for the foot substrate an a thick moo dielectric steady fabric is utilized for the beat substrate to upgrade the patch's radiation [18]. As a rule, the fix is put on the beat layer, and the ground plane contains a coupling gap situated beneath the middle of the fix for greatest coupling, coming about in lower cross polarization since of the symmetry of the arrangement. The sum of coupling from the bolster line to the fix depends on the shape, estimate and area of the gap.

The advantages are:

- False radiation is minimised since the ground plane separates the patch and the feed line.

The Limitations are:

- The difficulty of manufacturing due to the presence of multiple layers, which increases the thickness of the antenna.
- Narrow bandwidth.

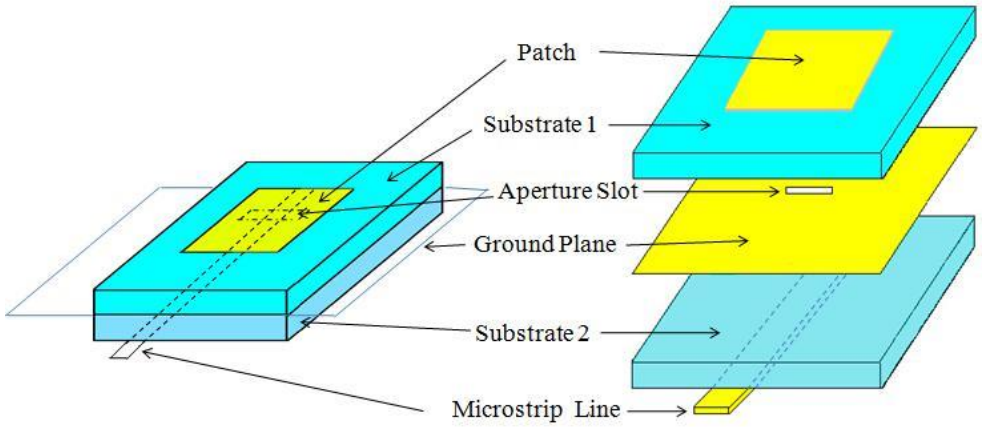

Figure 3.10: General view of the aperture coupling feeding technique [26].

3.9 METHODS OF ANALYSIS

Micro-strip fix receiving wires examination strategies have gone through a long handle of improvement. Micro-strip fix received wire examination could be a exceptionally prevalent strategy in conjunction with transmission line show, depth show, and full wave demonstrate [26], and incorporates basically fundamentally equations/Moment Strategy. In common, the transmission line model can be considered as the best of all and it gives great physical knowledge but it isn't as exact. On the other hand, depth demonstrate is more exact and gives great physical knowledge but is complex in nature. Too, full wave demonstrate can be considered the foremost exact, flexible and with the capability of taking care of single components, limited and unbounded clusters, stacked components, subjective formed components and coupling, but, this show is considered on a very basic level complex.

3.9.1. Transmission Line Model

This demonstrate delineates the micro-strip radio wires as two spaces with stature ‘h’, and width ‘w’ and they are isolated by the transmission line ‘L’, as can be seen in Figure 3.10. The created electric field lines will spread in substrate and small bit within the discuss. This recommends that this medium does not bolster Transverse electromagnetic mode (TEM). Since the electric field lines have distinctive stage speeds, there must be an successful dielectric steady (εreff) to account for the bordering and the wave engendering within the line. It should be mentioned that the esteem of εreff could be a small less than εr as the bordering areas around the fringe of the patch are not as it were limited within the dielectric substrate but they are moreover spread within the discuss as shown in Figure 3.10.

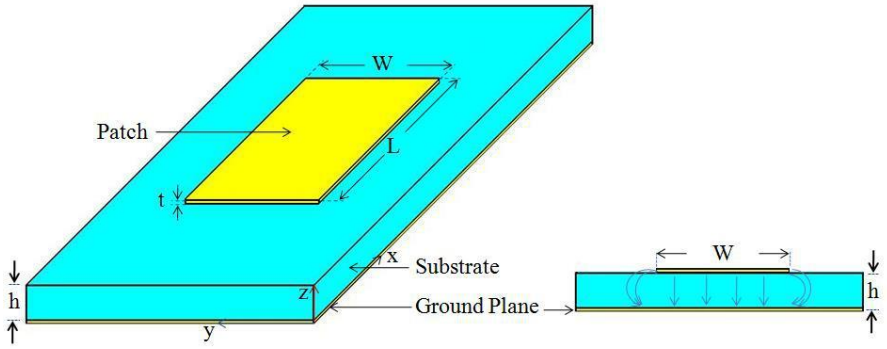

Figure 3.11: The Micro-strip line with electric field lines [26].

The TM10 mode is the fundamental mode where the length of the fix is required to be marginally less than a half wavelength (λg/2). The wavelength (λg) within the dielectric medium is communicated by the condition underneath. As specified prior, this show portrays the microstrip received wire as two openings isolated by the transmission line of length L. Along the structure's width the voltage is most extreme, whereas the current is least. Besides, the components of the field at the patch's edge are settled into extraneous component, along x-axis, and typical component, along z-axis, relative to the ground plane as displayed in Figure 3.10. The typical components of the electric field along z-axis are in inverse bearings, so that they cancel each other. Be that as it may, the extraneous components, along the x-axis, have the same course which comes about within the combination of the coming about areas and makes a greatest emanating field from the fix typical to the surface of the structure.

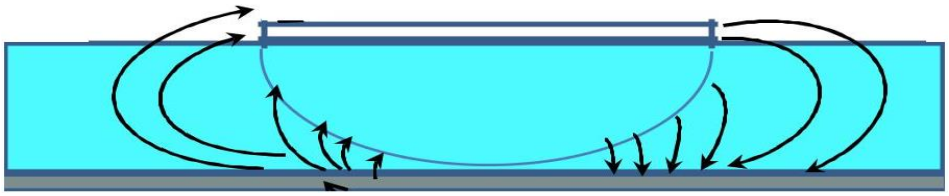

Figure 3.12: The physical and effective lengths of the rectangular micro-strip patch [26].

Moreover, the bordering areas along the width can be demonstrated as emanating spaces, and the length of the fix radio wire is shown as a small less than its electrical measurements since of the bordering impact. As such, the length of the fix, L, must be amplified by 2ΔL. This expansion of length L, is communicated as [46].

$$\frac{\Delta L}{h} = 0.412 \frac{\left(\varepsilon_{reff} + 0.3\right)\left(\frac{W}{h} + 0.264\right)}{\left(\varepsilon_{reff} - 0.258\right)\left(\frac{W}{h} + 0.8\right)}. \tag{3.3}$$

The resonant frequency of the rectangular patch antenna for the TM_{mn} mode is given by [46].

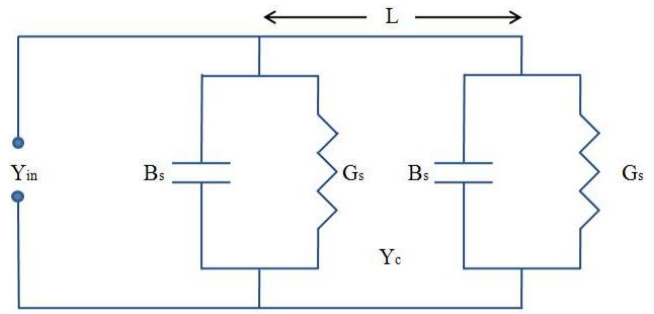

Figure 3.13: Equivalent admittance circuit model [26].

The areas at the two transmitting spaces are 180o out of stage at the radio wire reverberation. This happens when the length, L, could be a small littler than half a guided wavelength λg. The bordering areas at the transmitting edges can be accounted for by altering L, and so the susceptances cancel each other at the two openings coming about in a absolutely resistive input induction [26]. The E-plane is characterized as the plane containing the electric field vector and the heading of greatest radiation, and the H-plane is the plane containing the attractive field vector and the course of most extreme radiation.

3.9.2. Cavity Model

Depth show is another show accessible for examining micro-strip received wires. This demonstrate varies from the transmission line show because it offers distant better;a much better;a higher;a stronger;an improved">an improved way to show the radiation designs and is closer within the physical translation of the radio wire characteristics. An exertion is made to show a physical translation within the arrangement of the areas inside the depression and radiation through its side dividers depression demonstrate examination as appeared in Figure 3.13 [26].

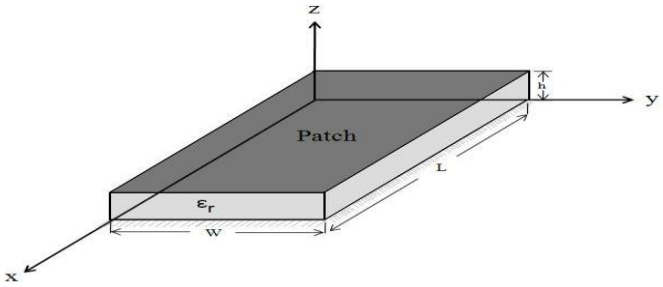

Figure 3.14: The cavity model geometry [26].

When the micro-strip fix is invigorated, there's a charge dispersion on the upper and lower surfaces of the fix, as well as on the surface of the ground plane. The charge dissemination is directed by two components: one appealing and the other terrible. The alluring component is between the comparing inverse charges on the foot side of the fix and the ground plane, which is likely to preserve the charge concentration on the foot of the fix. The terrible component is between the charges on the foot surface of the fix, which is likely to thrust a few charges around its edges, to its best surface [26]. The action of these charges makes coordinating current densities Jb and Jt , at the foot and beat surfaces of the fix, separately, As for most viable micro-strip the height-to-width proportion is exceptionally little, the alluring instrument rules and most of the charge concentration and current stream remain underneath the fix. A little sum of current streams around the edges of the fix to the best surface. On the other hand, this current stream decays as the height-to-width proportion decreases. At the constrain, the current stream to the best would be zero, which in a perfect world would not make any extraneous attractive field components to the edges of the fix. This would allow the four side dividers to be displayed as idealize attractive conducting surfaces which in a perfect world would not meddled with the attractive field and, in turn, the electric field conveyances underneath the fix. In hone, a little limited height-to-width proportion exists, and the digressive attractive areas at the edges would not be precisely zero, but since they will be little, a great estimation to the depth demonstrate is to consider the side dividers as culminate attractive conductors.. This show produces great standardized electric and attractive field conveyances (modes) underneath the fix.

For the essential mode, the electric field is zero within the center of the fix, most extreme (positive) on one edge, and least (negative) on the other edge. This variety between most extreme and least is the same as the stage of the RF flag. The field setups for rectangular microstrip fix received wire are appeared in Figure8.13 with the conveyance of the electric field along the side dividers of the depression. The electric areas of the TM010, TM001, TM020 and TM002 are shown in Figures 3.14 a – 3.14 d, separately. For the TM010 mode, the electric field changes in half wavelength along the length, L, of the fix. On the other hand, there are no varieties along the width, W, and the stature, h, as demonstrated in Figure 3.13a. So also, the electric field changes along the width within the TM001 mode. In TM002 mode the electric field shifts in one

wavelength along the width, without any variety along the length and tallness as seen in Figure3.14 d. The same applies to TM020, which is demonstrated in Figure3.14c.

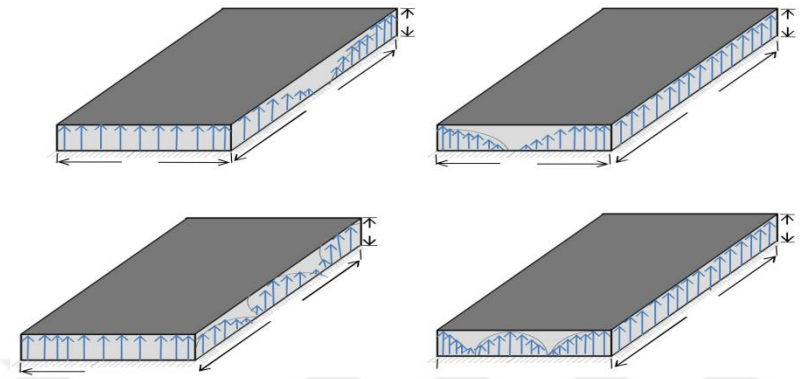

Figure 3.15: The field configurations for rectangular micro-strip patch in cavity model [26].

In light of the over, the depth demonstrate offers an amazing knowledge into the radiation instrument of the micro-strip received wire. On the other hand, this demonstrate does not consider the impacts of the nourish, the limited degree of the substrate and ground plane. These impediments can be overcome utilizing thorough numerical strategies spoken to by full- wave arrangements.

3.9.3 Numerical Techniques (Full wave Model)

A number of modern numerical procedures have been created and utilized for the examination of irregular-shaped structures. Explanatory arrangements of Maxwell's conditions are as it were realistic for the basic geometries. There are a few 3D commercial full-wave program bundles that give remarkable execution and knowledge into all of the 3D-EM issues. In this proposition, two distinctive commercial computer program bundles: Computer Reenactment Innovation Microwave Studio (CST MWS) and Tall Recurrence Structure Test system (HFSS) with its premise within the Limited Fundamentally Method (FIT) and Limited Component Show (FEM), separately, are utilized to upgrade the antenna's measurements and survey their execution in terms of radiation designs, diffusing parameters and scattering chart.

3.9.3.1 CST MWS software package

CST MWS can be a full-fledged electromagnetism software package for electromagnetic investigation and long-term planning for replication. CST MWS has a few FIT-dependent switches suitable for different types of applications, where no single strategy works equally well in all application areas such as time management, space repeater analyzer, indispensable condition analyzer, Eigen mode analyzer , To suit their needs specific applications [26] - [17]. Transitory provides the most significant adaptive adaptability to obtain the full bandwidth replication behavior of the simulation tool in terms of a single computation. In addition, the CST MWS provides a range of versatile standards that deliver accurate office performance, such as workflow modulation based on wavelength expansion and network settings reduction. Also, CST MWS provides boundary conditions, since the computer has the ability to calculate problems with limited extension. As such, the application of boundary condition may be a noteworthy task. Includes an open space in all titles with a connected Idealize Coordinate Layer (PML). The PML definition provides a four-tiered independent fabric retention requirement and a reflection coefficient number of 0.0001 as shown in Figs 3.13 and 3.14 individually.

4. ANTENNA ARRAY AND UNIT CELL GEOMETRICAL DETAILS

This segment talks about the proposed radio wire cluster based on dismiss band channel. The radio wire component is based on a wind line monopole energized with a quarter wavelength transformer as seen in Fig. 1. The transformer is outlined to function at 2.45GHz. By the by, the radio wire cluster is built from two components are isolated with a MTM unit cell. In any case, the received wire cluster is supported with fractional ground plane and abandoned with an EBG unit cell. The partition remove between the two received wires is approximately 20mm when mounted on an FR-4 substrate of 1mm thickness. The MTM unit cell structure is outlined from the 3rd arrange Hilbert-shaped fractal mounted on the back side of the substrate between the radio wire components. Whereas, the EBG deformity is carved from ground plane between the nourishing quarter wavelength transformers on the back. The feeding point may be a 50 Ω SMA port. The FR-4 substrate permittivity ε_r is 4.2 with a loss tangent $tan\delta$=0.00167. By the equation analytical estimate we can calculate the resonance frequency:

$$f_r = \frac{2c}{3S_e\sqrt{\varepsilon_r}} \tag{4.1}$$

where, c is the light velocity and S_e is the patch side length [5]. The proposed EBG defect is the complementary geometry of the proposed MTM with a larger size of 1.6 factor.

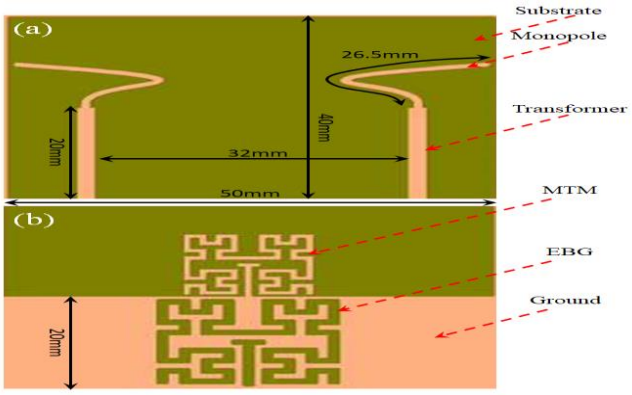

Figure 4.1: Antenna designed geometry: (a) in the Front view and (b)in the Back view.

Presently, the proposed channels properties are inferred based on an expository circuit show. The circuit demonstrate is displayed in Figure4.2.

In table 4.1. the assessed circuit demonstrate values are recorded. These values are assessed concurring to model of giving a dismissal around 2.45GHz with at slightest 25% transmission capacity.

Table 4.1: Lumped elements values.

Parameters	Value
G_F	0.2S
L_F	2.7nH
C_F	9.1nF
R_F	20Ω
G_T	0.4S
L_T	0.7nH
C_T	10nF
R_T	23Ω
G_{EBG}	5S
L_{EBG}	0.7nH
C_{EBG}	0.1nF
R_{EBG}	40Ω

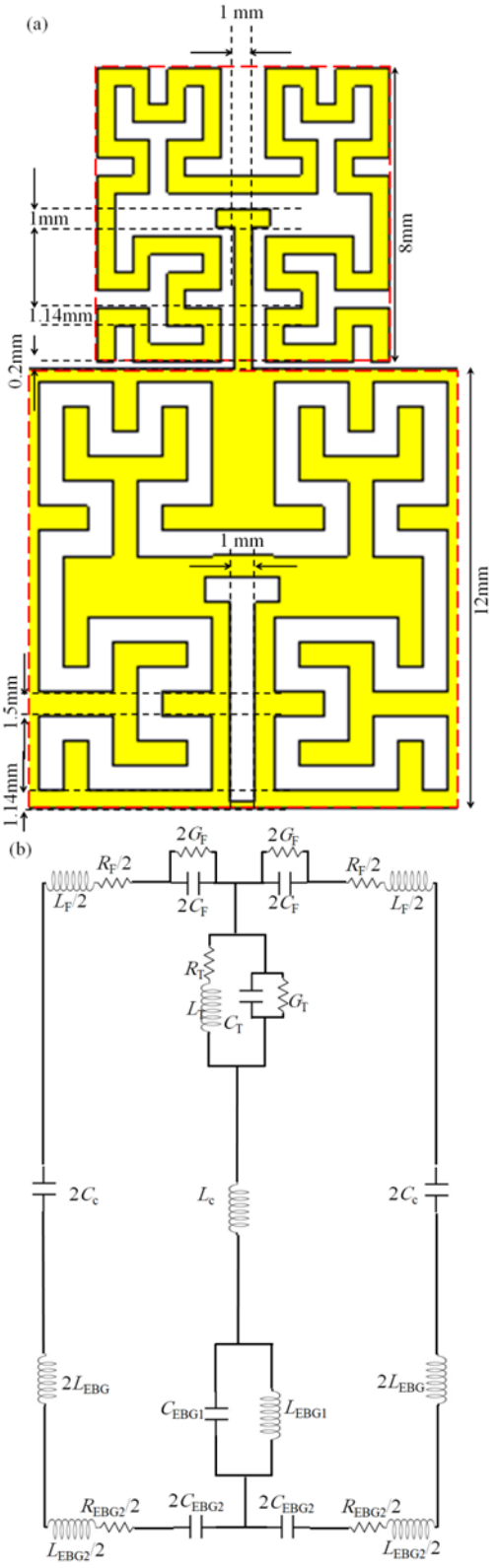

Figure 4.2: Designed Antenna Unit cell: (a)the Geometrical details and (b)the Equivalent circuit .

4.1 UNIT CELL PROPERTIES

The scatter diagram of the proposed unit cell is evaluated in terms of TE and TM using CST MWS [20]. Subsequently, the HFSS computer software packages [21] and advertisements [22] are used for approval. From the assessment reached, the designed unit cell shows a bounce frequency of 2.25 GHz to 2.75 GHz - you can see this in Figure 4. 3. With respect to the re-enactment of legislation in CST MWS, two criteria were implemented to ensure accurate meeting [6]. The basic metric is to estimate the three-dimensional structure and the field field in the space appropriately. Thus, a mixed-use selection is connected starting from the beginning of Nx = 40, Ny = 40, and Nz = 20 with the line ratio limiting by 50. Four sheets of PML with reflection - From 10-4 in normal cases with a separate of λ / 3 at 2.45GHz [7]. The work of the hexagon is used with a 30 refining account for the metal edges [8]. The refresh is performed when SS11 reaches less than 2% between any two consecutive passes in the frequency range of curiosity. This condition ends in a dynamic decrease of -40 dB under the most extreme estimate. Within the HFSS re-enactment, where the basic requirement was required where more than 11% S starting from 2% was used starting from the start of the 12,188 tetrahedral work. The quick, clear selection is related to the straight-through frequency step from 1 GHz to 4.5 GHz with a 0.1 GHz step estimate. To begin the order of the work hypothesis, it is adjusted with the maximurevisio per pass by 30% [8].

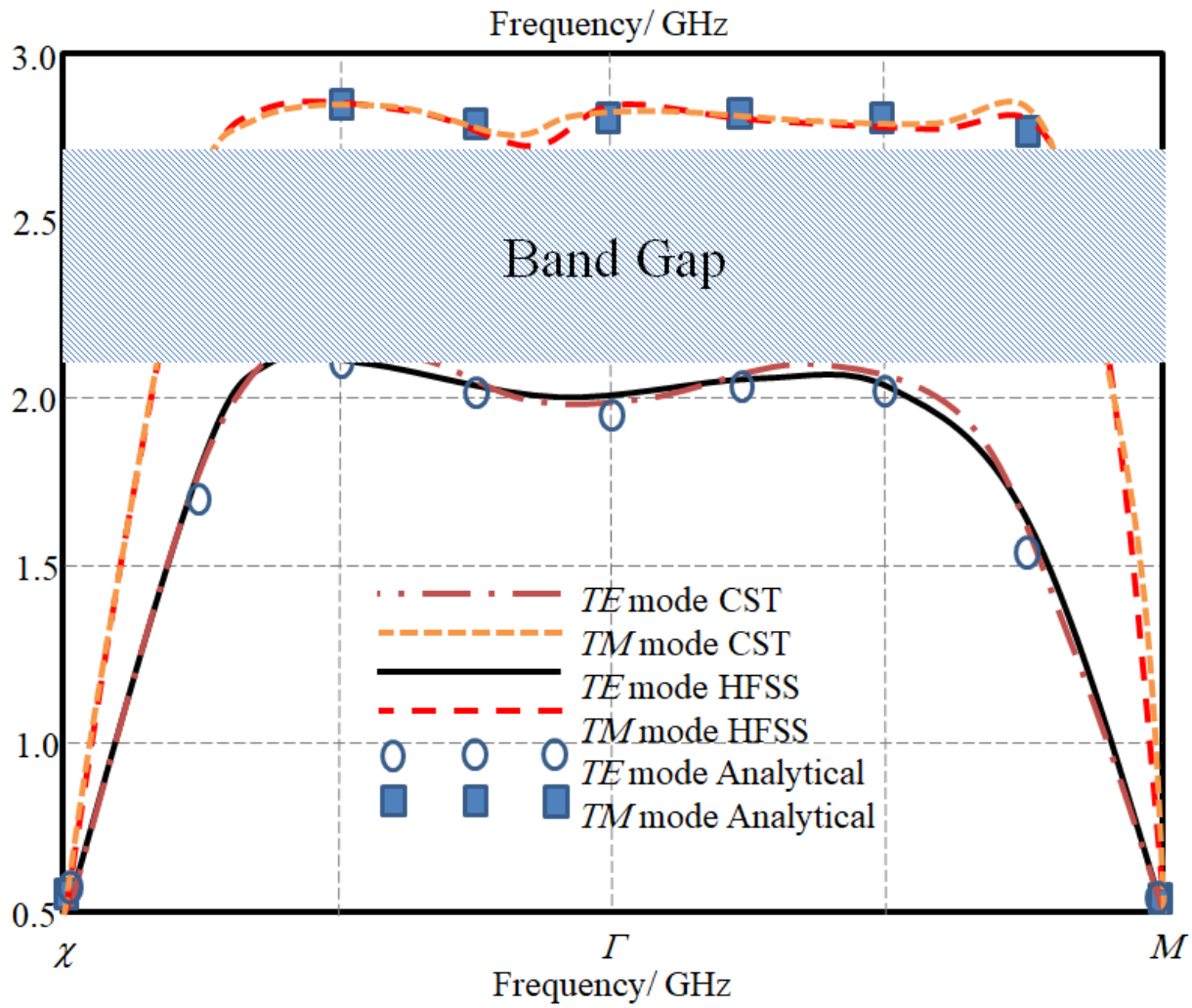

Figure 4. 3: The Evaluated dispersion Figure for the designed (simulated) unit cell.

The proposed unit cell electromagnetic properties in terms of the refractive file (n) from the relative permittivity (εr) and relative penetrability (μr) are assessed as outlined in Figure. 4.3 In this manner, CST MWS is conjured to assess the constitutive electromagnetic parameters spectra inside the recurrence run of intrigued. In this think about, the proposed unit cell is embedded interior an air-filed square cross-sectional range waveguide to imitate the free space environment. For this, electrical conductive and attractive dividers are doled out around the unit cell edges [6]. In any case, two open waveguide ports are found at the closes of the waveguide for excitation [7]. The proposed person unit cell appears refractive record (n) approach to zero agreeing to the connection of =√εμ , where, ε≈μ≈0 at 2.4GHz. This can be since the impact of the capacitive parts, given by the Hilbert structure, release within the inductor parts, given by T-stub [8] , that

balances the radiation spillage on the unit cell surface. Subsequently, the overall refractive file (n) for such structure approaches zero [29]. From the received diffusing lessening examination [32], the unit cell must encompass the question to delay the scrambling from the protest and to return the electromagnetic waves back to its beginning. Subsequently, in the event that the intelligent record of the proposed unit cell is less than solidarity, the stage speed would be more prominent than the speed of light within the free space [33]. It is found that the proposed unit cell based on a single unit cell shows a zero refractive index at 4GHz. However, when two unit cells are coupled, as presented in Fig. 1, the proposed unit cell shows zero refractive index (*n*) at about 2.5GHz.

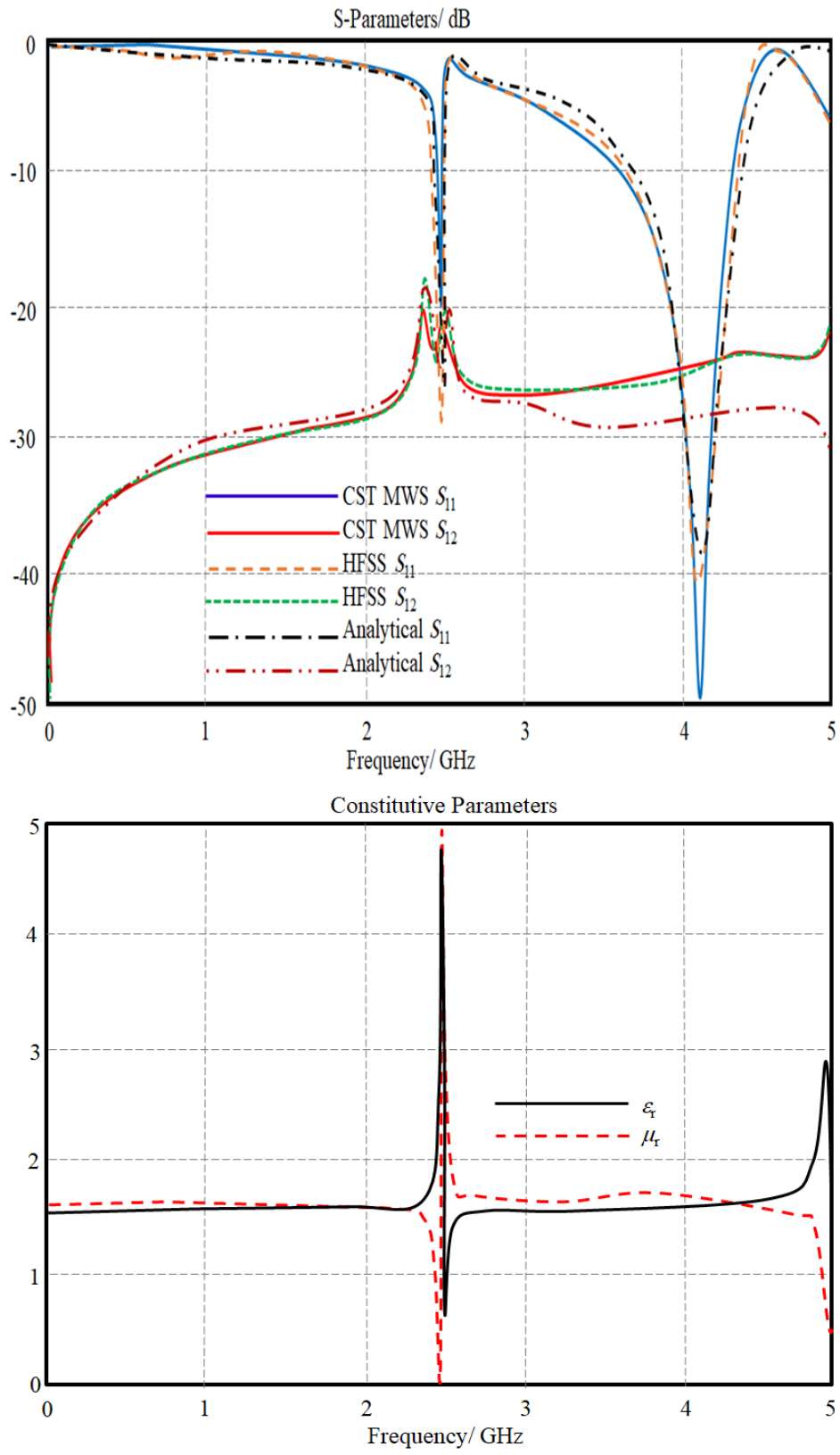

Figure 4.4: The designed Antenna parameters with unit cell

4.2 ANTENNA ARRAY RESULTS

To consider the impacts of presenting the proposed unit cells on the shared coupling diminishment between the proposed radio wire components, two benchmarks are considered. To begin with, the division remove between the two received wires must not increment over λ/2 at the recurrence band of intrigued. Moment, the coupling basis, characterized in terms of S12, must not surpass -20dB. Subsequently, the radio wires shared coupling between the radio wire components without unit cells is assessed in show disdain toward of the λ/2 basis by checking S12 range. In any case, the other related execution comes about without unit cells are assessed and compared in like manner as taking after:

4.2.1 Effect of the Unit Cell Introduction

A numerical parametric think about is conducted to realize the impact of presenting the proposed unit cell on the receivig wire shared coupling. To assess the common coupling lessening with diverse number of unit cell presentation, a numerical parametric ponder is conducted. In such think about, S12 range is observed with changing the unit cell structure as: Without unit cell, MTM as it were, and the proposed unit cell as MTM and EBG structures. As seen from S12 spectra in Fig. 5, it is found that the shared coupling at 2.45GHz is approximately -9dB, without unit cell presentation. Be that as it may, it is decreased essentially to -13dB when MTM is presented. At EBG, the S12 is decreased to -16dB. At last, at utilizing the proposed unit cell, the coupling is decreased to -31dB as delineated in Figure 4.5. Such common coupling diminishment is due to the capacity of the proposed unit cell to supply ZRI reaction, in which offers an electromagnetic cloaking at 2.45GHz. This is often accomplished by the taking after:

$$\eta = \sqrt{\frac{\mu}{\varepsilon}} = \frac{\mu}{\sqrt{\varepsilon\mu}} = \frac{\mu}{n} \tag{4.3}$$

$$\beta = \frac{\omega\sqrt{\varepsilon\mu}}{c} = \frac{\omega}{c}n \tag{4.4}$$

where, η is the electro-magnetic wave impedance, β is the electro-magnetic wave phase change, and c is the speed of the light [40]. Therefore, when (n) in the limit of zero as shown below [6]:

$$\lim_{n \to zero} (\eta) = \infty \tag{4.5}$$

$$\lim_{n \to zero} (\beta) = 0 \tag{4.6}$$

Therefore, according to equations [7] (4.5) and (4.6), the resulted wave through such structure would be a vanished mode at which the phase change is zero with a very high impedance medium[6].

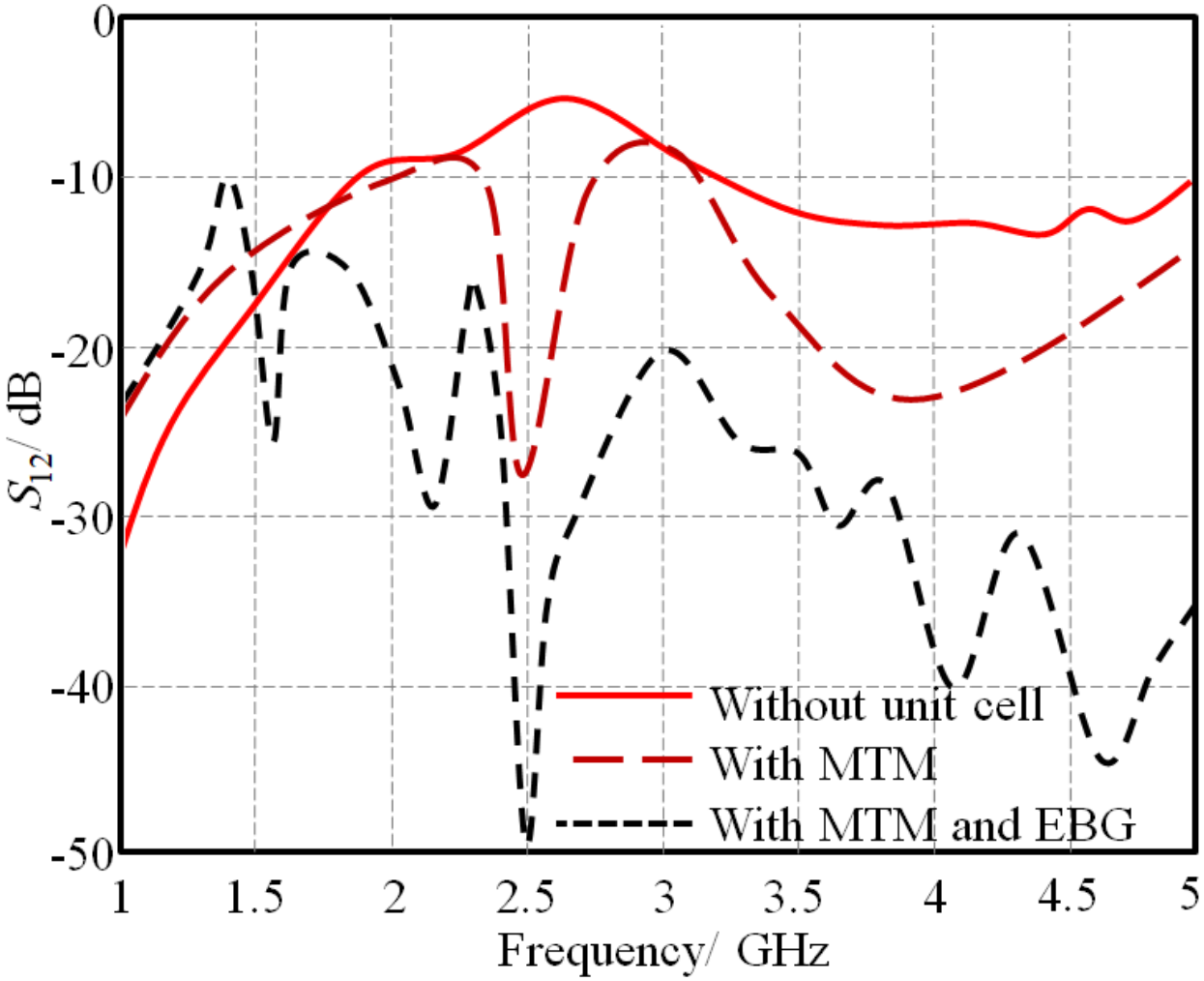

Figure 4.5: The changes in the mutual coupling spectra for different numbers of the proposed designed Antenna.

4.2.1.1 Separation distance

The partition impacts between the MTM structure and the EBG abandons on the common coupling are assessed at three diverse separations at 1.7mm, 2.7mm, and 3.7mm as seen in Fig. 6. It is found the coupling impacts decreased to -38dB at 2.7mm, be that as it may, it is expanded to -30dB when the remove gets to be 3.7mm, whereas, when it is settled to 1.7mm is found to be -

23dB. From this ponder, it is watched that such coupling diminishment changes conversely with division separate increment; typically since the alter within the capacitive impacts between the MTM and EBG structures [41]. Thus, the cross item of the initiated current (HA) on the radio wire surface and the surface waves (HS) gets to be more noteworthy than zero that increments the coupling impacts [42] as given bellow:

$$H_A \times H_S > |H_A||H_S|\sin(\theta) > 0 \quad (4.7)$$

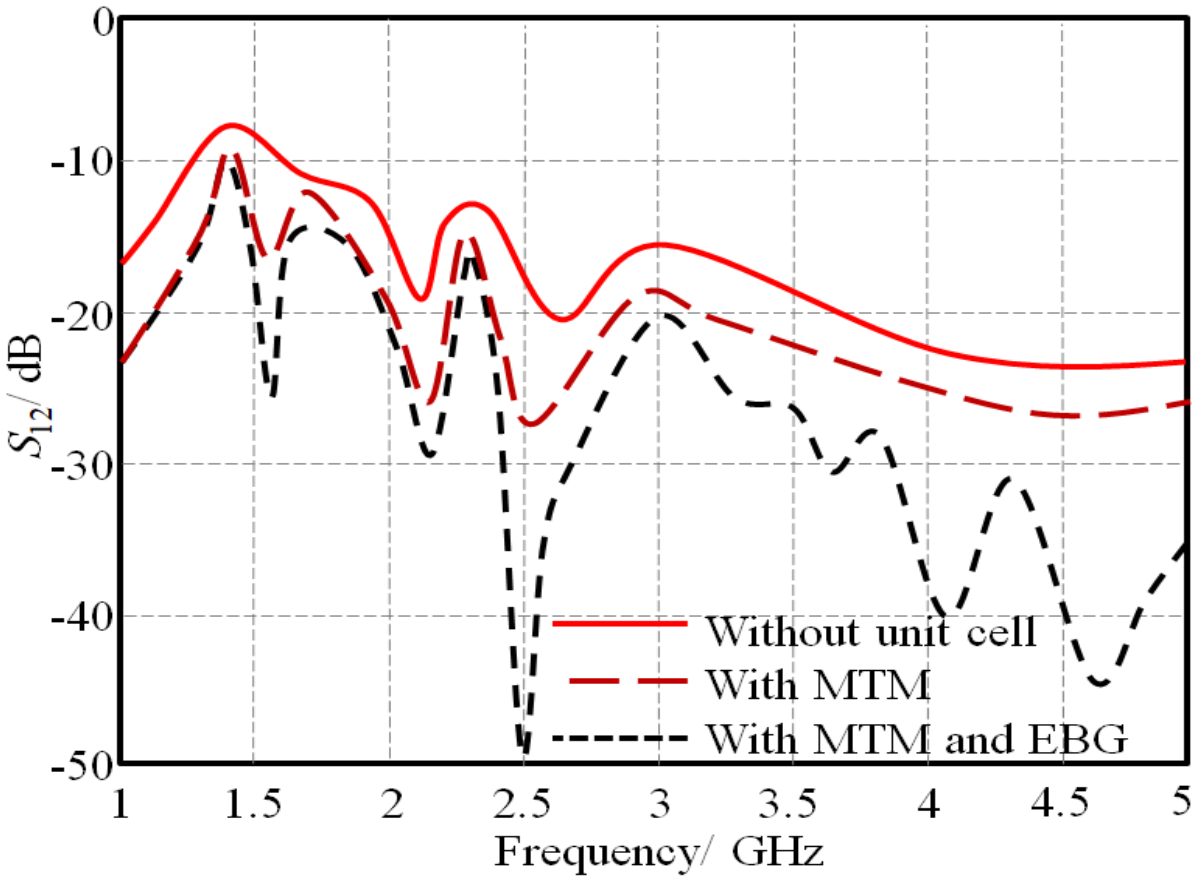

Figure 4.6: The effects of Adding and changing the separation's distance on the mutual coupling reduction in the antenna .

4.2.1.2 Effect of the unit cell orientation

Situating the proposed unit cell with 90o around the center and the impact on the common coupling lessening is examined in this area. It is found that the proposed unit cell appears totally distinctive reaction in terms of S12 after the introduction. As seen in Fig. 7, the shared coupling is compared some time recently and after the introduction. After the introduction, it is found that

the greatest |S12| lessening is moved from 2.45GHz up to 2.55GHz with |S12|=-20dB, be that as it may; |S12| is found almost -4dB at 2.45GHz. Hence, it is watched by pivoting the proposed unit cell by Θ=90o, the heading of the surface wave gotten to be in parallel to the initiated current on the radio wire components that increments the coupling impacts [36].

$$H_A \times H_S = |H_A||H_S| \sin(90) = |H_A||H_S| \tag{4.8}$$

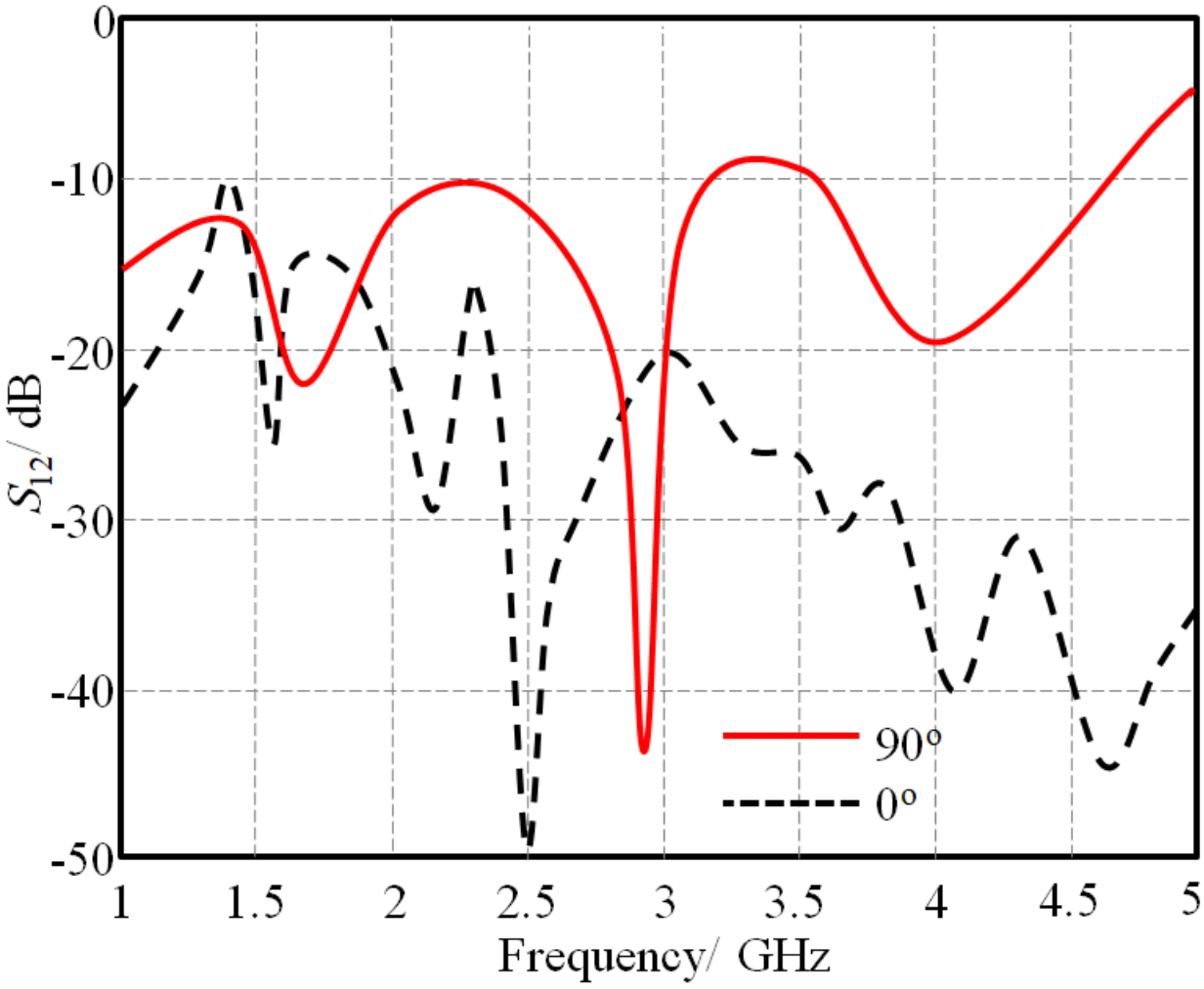

Figure 4.7: The effects of the designed Antenna on the resulting mutual coupling reduction.

4.2.1.3 Ground plane effects

It is watched a critical alter within the spectra of S12 with regard to changing the ground-plane width. It is chosen in this area to screen the coupling impact in terms of S12 by changing the ground-plane width as 40mm, 45mm, 50mm. From such consider, it is found the least coupling

happens when the ground plane width is 30mm. This might be credited to the impact of the surface wave hindrance from the ground plane edges, in which by expanding the ground plane width the coupling diminishment is accomplished [27]. In addition, the capacitive coupling between the ground plane edges may spill the radiation along the ground length [43].

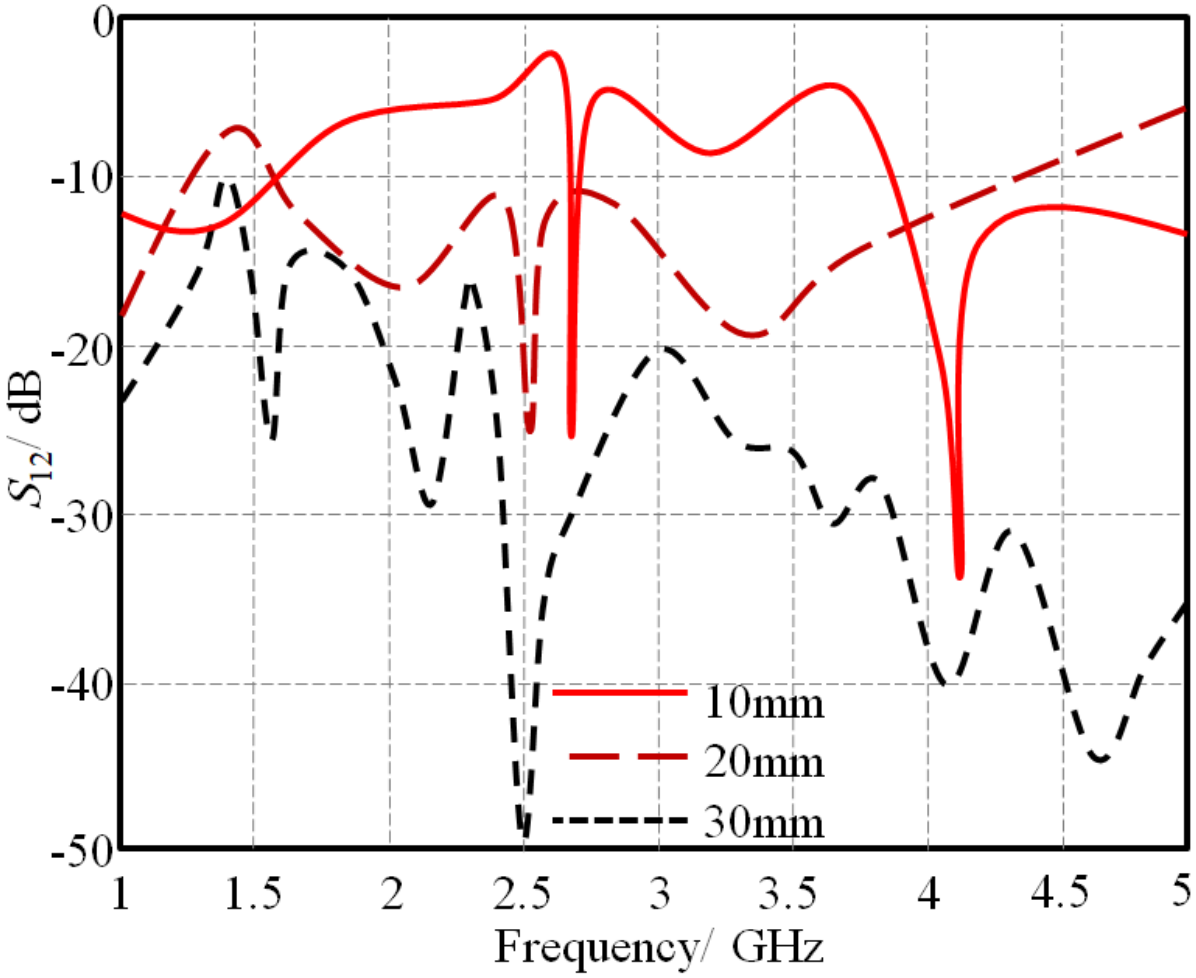

Figure 4.8: The effects for the ground-plane width on the mutual coupling reduction.

4.2.1.4 Space filling comparison

Figure 4.9 appears the gotten common coupling, S12, between the two received wire components. The space between the proposed radio wires is recorded with discuss (without incorporations), unit cell, and, PEC in arrange to realize the impact of the unit cell presentation on the shared coupling. It ought to be famous, the same separate between the proposed received wires are settled agreeing to the ideal division remove that was talked about already. It is

watched that the proposed unit cell gives the most extreme dismissal at 2.45GHz, in which, the discuss crevice and PEC presentation cannot give any diminishment at the recurrence band of intrigued. Note, the displayed case with discuss fill is considered to speak to a comparison of the common coupling without presentation of the proposed unit cell. In this manner, it is sealed that the impacts of including the proposed unit cell acts as culminate attractive dividers to dismiss the surface wave spillage [42]. Subsequently, the accomplished dismissal of S12≈0 due to the gotten introduction point between the H_S and H_A is 0^o as seen below:

$$H_A \times H_S = |H_A||H_S| \sin(0) = \text{zero} \tag{4.9}$$

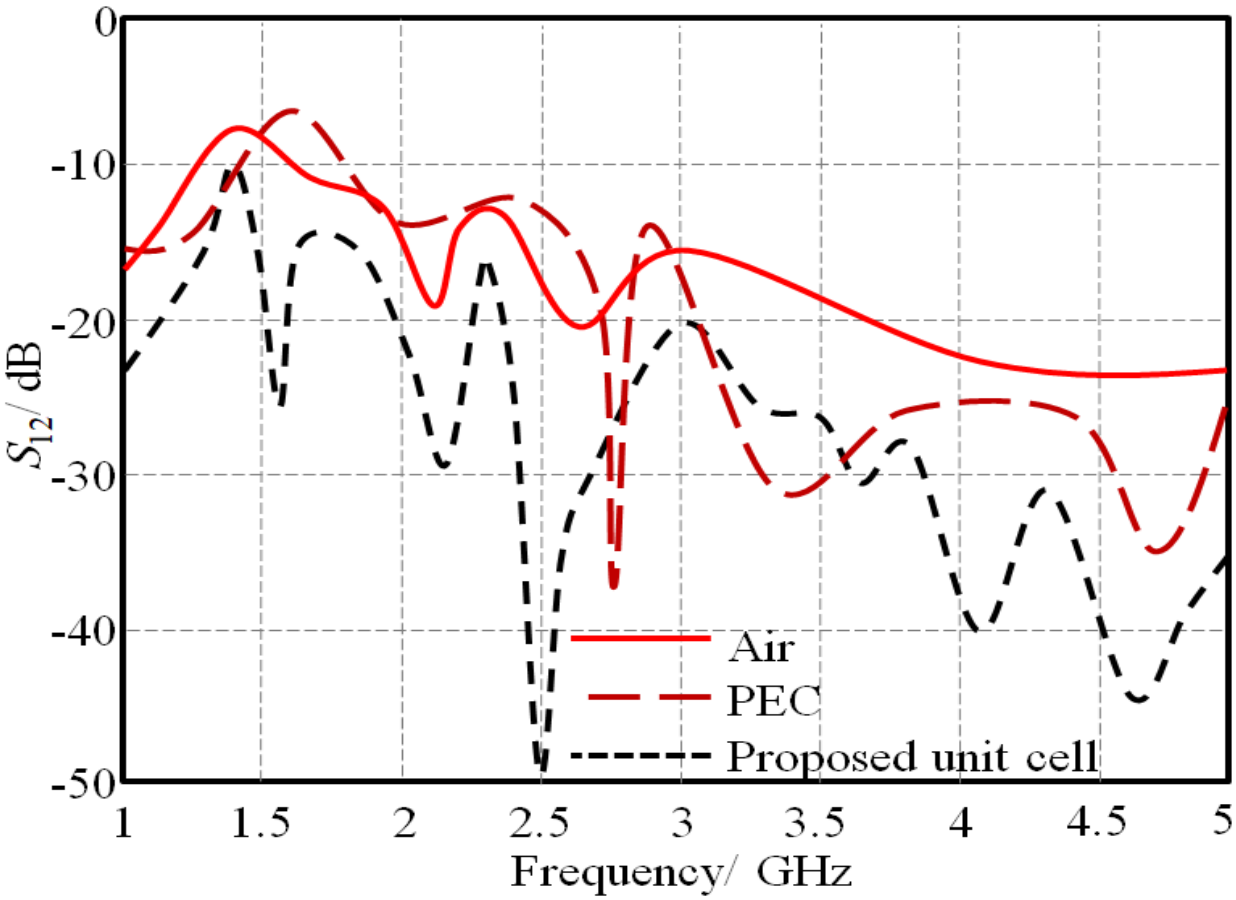

Figure 4.9: the mutual coupling when we are changing and filling the space .

4.2.1.5 Antenna return loss and radiation patterns

In Fig. 10, the radio wires return misfortune in terms of and S11 and/or S22 spectra are portrayed with and without unit cell presentation. The proposed receiving wire without unit cell appears

great coordinating, |S11|=-13.2dB, in any case, the radio wire coordinating is improved essentially to -46dB at 2.45GHz. Such upgrade is due to the impact of capacitive coupling between the radio wire structure and the proposed unit cell [36].

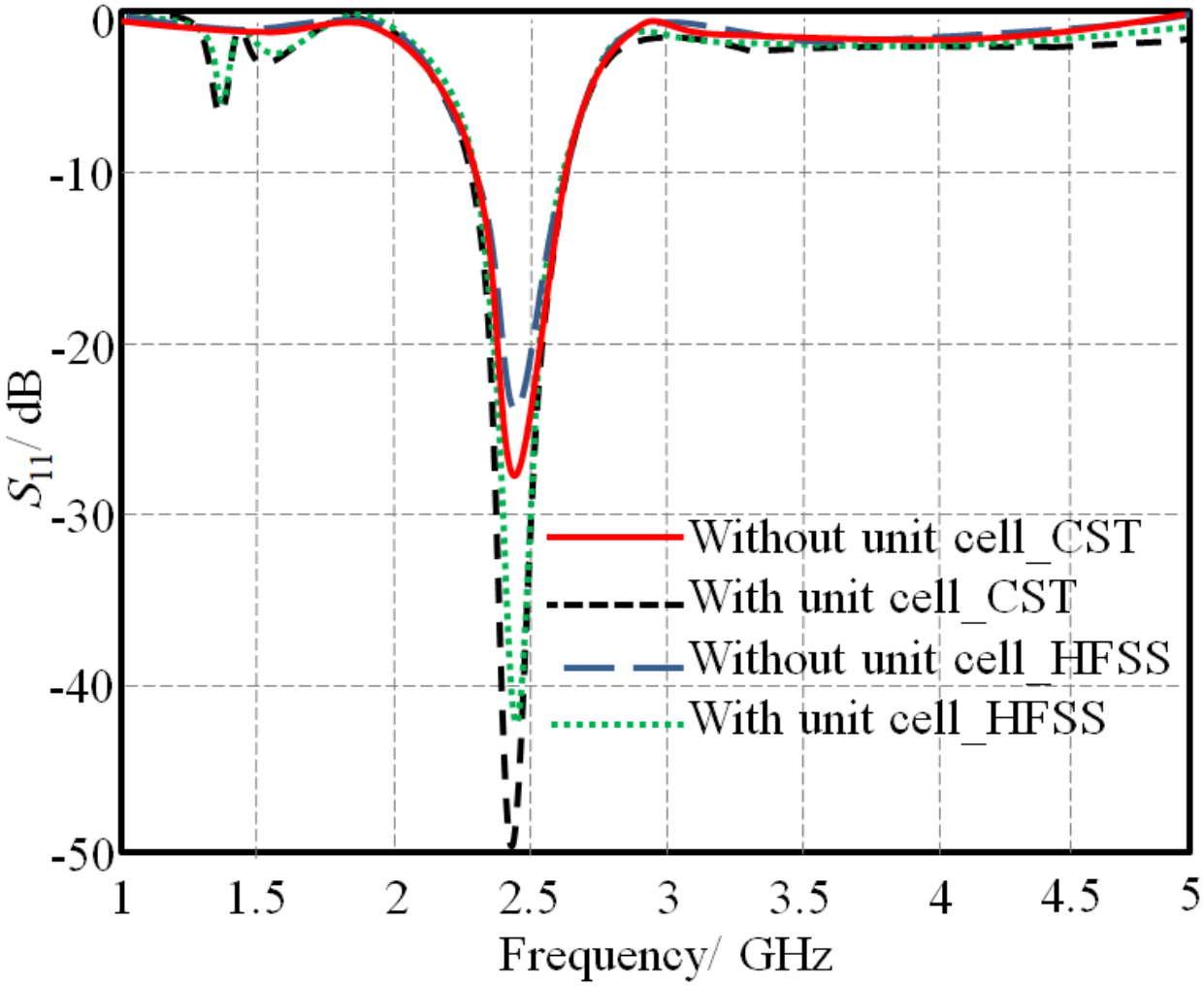

Figure 4.10: S_{11} spectra for the designed antenna array with and with-out unit cell.

The received wire radiation designs, within the far-field, are displayed in Fig4. 11 with and without unit cell. The radiation designs are assessed along the φ=90◦ and φ=0◦ planes. It is found a noteworthy alter within the radio wire radiation designs at both planes where most of the radiation is coordinated distant absent from the cluster center due to the unit cell presentation. On the other hand, the radio wire radiations without unit cell presentation are found generally coordinated toward to the other radio wire components to lead to a significant coupling. Hence, it is affirmed that the most received wire coupling decrease is credited to the proposed unit cell

cluster. It is critical to specify, that it is inconceivable to keep the same radiation designs some time recently including the unit cell considerations since their tall impedance effects. Subsequently, such alter within the radiation patters is exceptionally alluring within the MIMO applications to keep the tall radiation differing qualities within the far-field.

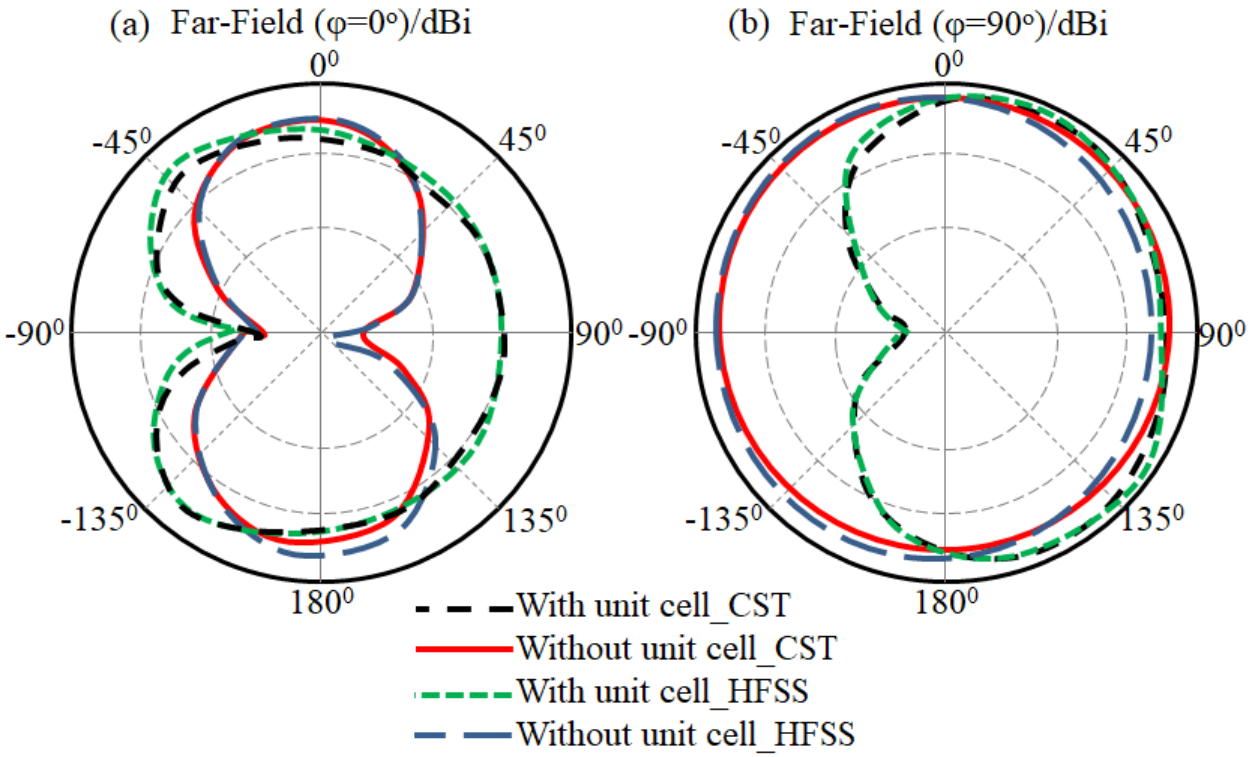

Figure 4.11: the designed antenna array's Radiation patterns with and with-out unit cell.

4.3 RESULTS VALIDATION AND DISCUSSIONS

The proposed radio wire cluster is manufactured and measured in terms of S-parameters and radiation designs tentatively. A chemical carving prepare is utilized to manufacture the proposed model as seen in Fig4. 12. The radio wire creation is conducted beneath the air situations. The radio wires are bolstered by 50Ω SMA ports through fastening prepare. The radio wire cluster is set on conductive surface of 1mm thickness as a ground plane. The external conductor of the SMA harbour is patched to the ground plane surface. The inward conductor of the SMA harbour is patch to the radio wire structure through the ground plane with-out conducting the conductor of the ground plane.

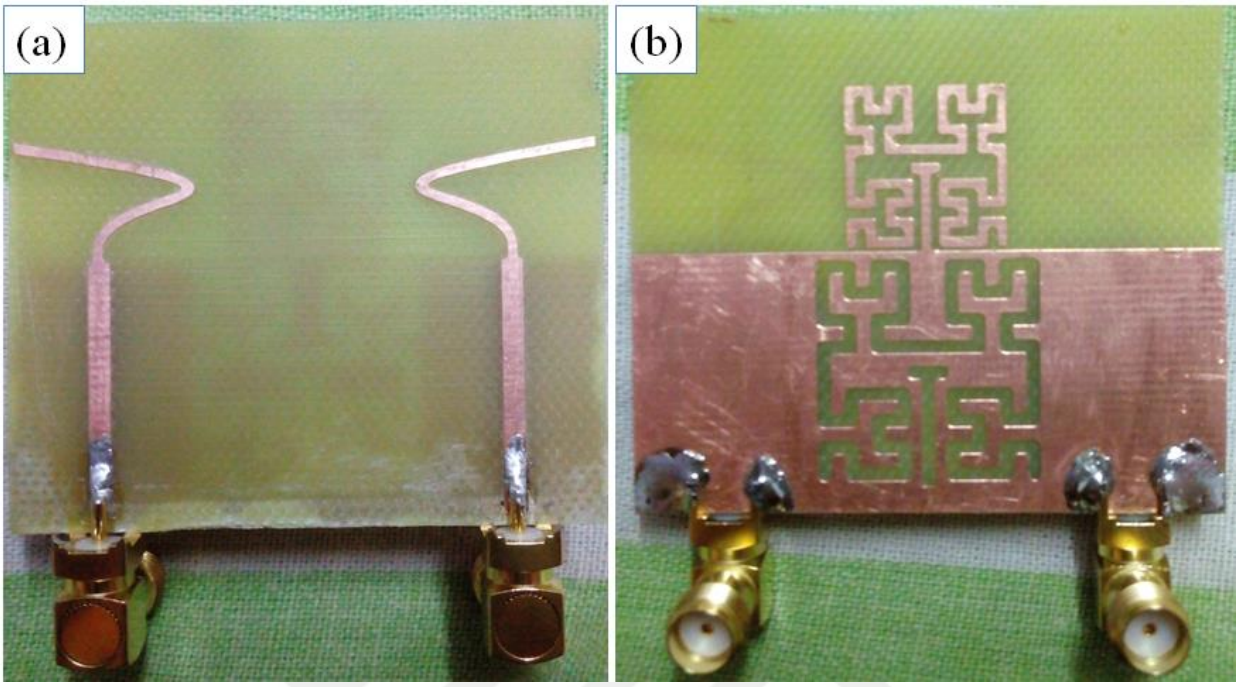

Figure 4.12: Designed Antenna; (a) in the front view and (b) in the back view.

From test comes about, it is found that the proposed radio wire cluster appears inconsequential coupling at 2.45GHz as approved in Fig4. 13. The gotten S-parameters from CST, HFSS, and estimation in terms of S11 and S12 spectra are found in amazing assention. The received wire appears an amazing coordinating |S11|<-10db at 2.45GHz. The coupling coefficient in terms of |S12| between the two radio wires is found to be -41dB at the same recurrence band of intrigued. Such coupling decrease is credited to the presentation of the proposed unit cell. The surface waves on the cluster surface are vanished due to the tall surface impedance of the proposed unit cell that gives great dismissal on both sides.

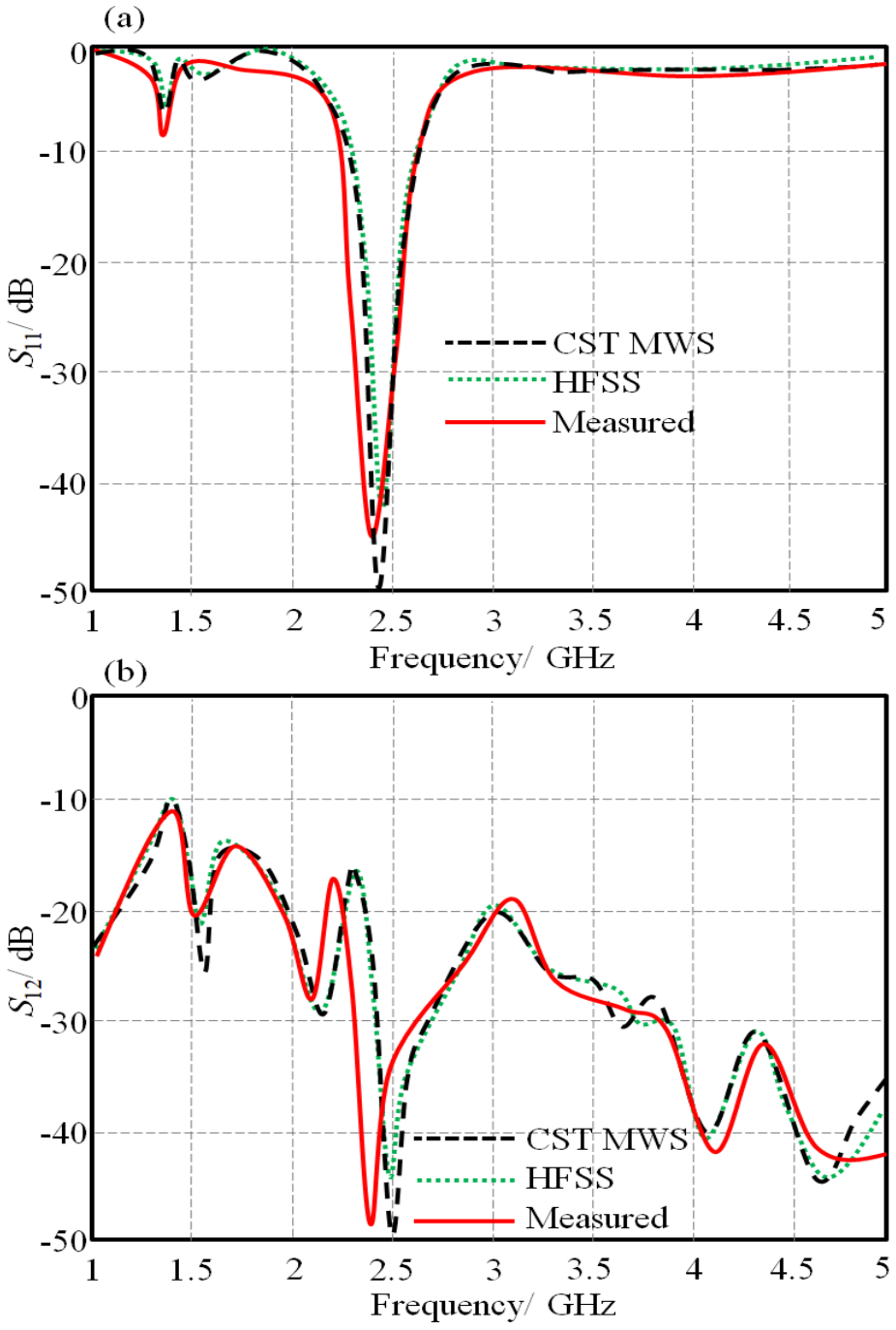

Figure 4.13: S-parameters spectra for the designed antenna array; (a) S_{11} and (b) S_{12}.

Presently, the received wire radiation designs are displayed in Fig.4. 14 from reenactment and estimation comes about. It is found that the designed radio wire cluster radiation designs are coordinated absent from each other against the cluster center. By the by, the proposed received

wire components give a bore-sight pick up of 2dBi at 2.45GHz. Such accomplishment is ascribed to the presentation of the unit cell that give tall impedance to dispose of the back hurls impact which in turns expanded the bore-sight hurls.

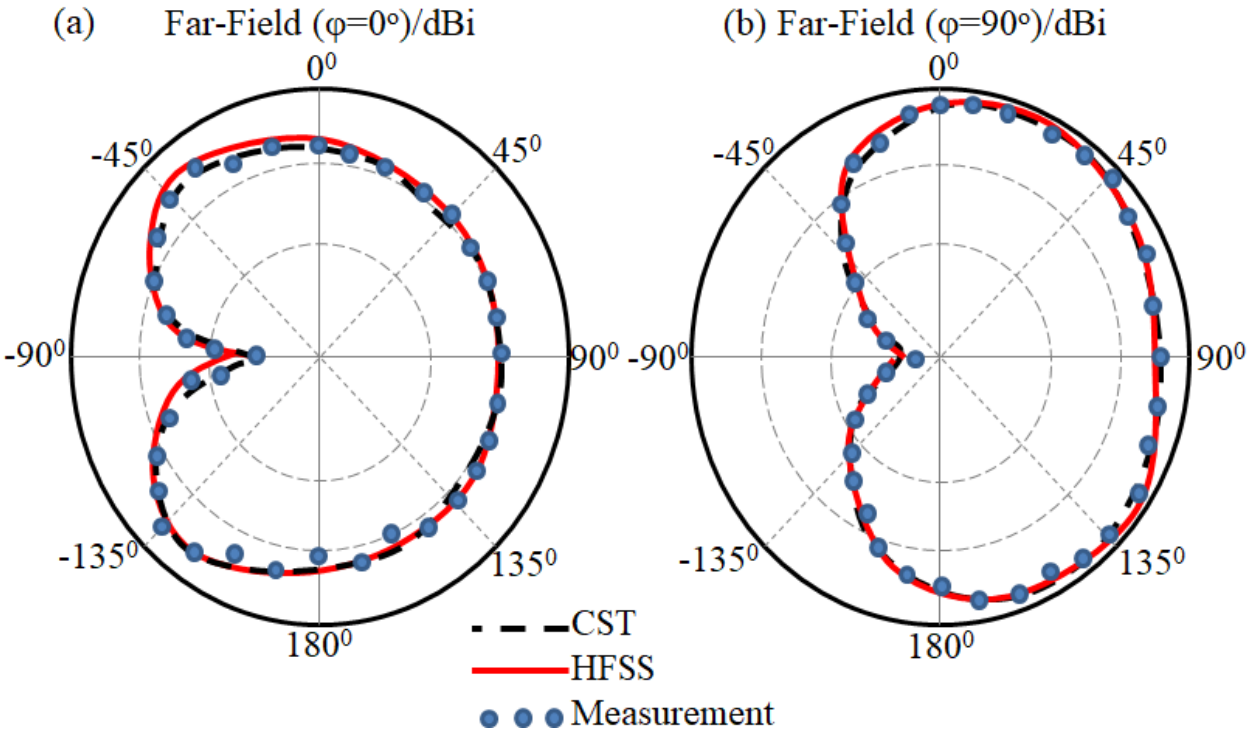

Figure 4.14: The measured radiation patterns for the simulated results: (a)for the E-plane and (b)for the H-plane.

The reenacted and measured relationship envelopes are appeared in Fig.4. 15(a). From these comes about, the proposed received wire cluster appears a moo relationship figure from the S-parameters around 0.02 at 2.45GHz that suits for using in the Wi-Fi communication frameworks [10]. The recreated and measured gather delay spectra are appeared in Fig.4. 15(b). It can be watched that the mutilation at 2.45GHz is less than 1ns. The gather delay variety is found to be within the normal of 1ns which appears great stage linearity.

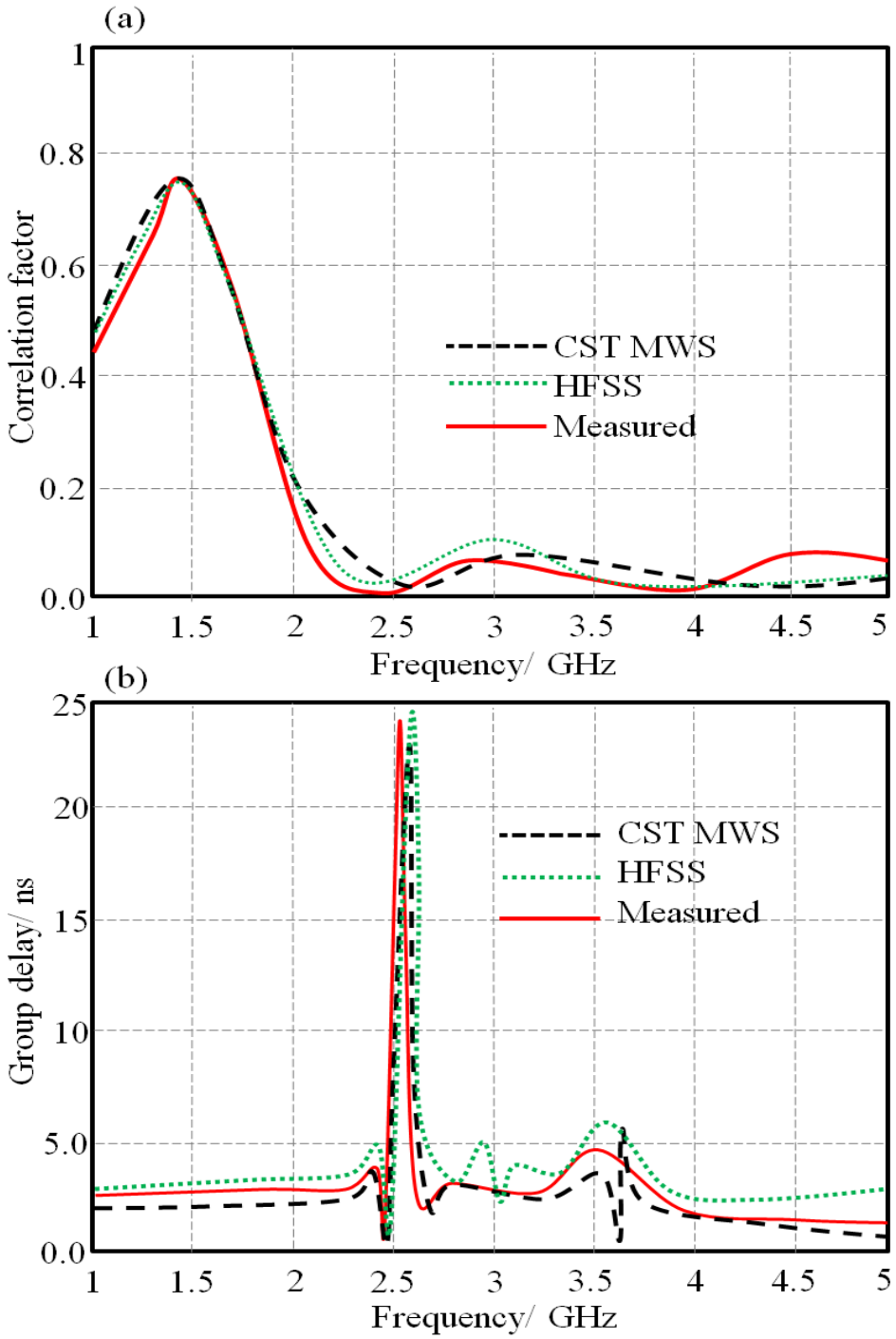

Figure 4.9: The difference between the measured and the simulated results in the design: (a) the Correlation spectra and (b)the Group delay spectra.

5. CONCLUSION AND FUTURE WORKS

5.1 CONCLUSION

The execution of a miniaturized MIMO received wire cluster based on two monopoles isolated with unit cell is explored hypothetically and tentatively. The proposed received wire components are printed on an FR-4 substrate and mounted ordinarily on copper ground plane. The plan strategy based on explanatory and numerical reenactments of the proposed unit cell and the radio wire cluster structure are examined in subtle elements utilizing CST, HFSS, and Advertisements details. The unit cell is built in this work to have tall surface impedance at 2.45GHz to suite the Wi-Fi applications. The impacts of presenting unit cell on the radio wire cluster execution are realized in terms of S-parameters and received wire radiation designs. It is found a critical shared coupling lessening, |S12|=-40dB, after presenting the proposed considerations to the received wire cluster. By the by, the received wire addition misfortunes, |S11|=-30dB, isn't influenced due to the unit cell structures. In addition, the received wire radiation designs are more coordinated within the inverse headings from the proposed received wire cluster center. The radio wire gives a bore-sight pick up approximately 2dBi at 2.45GHz. In this manner, the proposed unit cell is found to be an proficient strategy for protecting the electromagnetic back spillage in such MIMO clusters. The radio wire cluster execution in terms of relationship, differences, and gather delay spectra are characterized. It is found that the proposed cluster gives a moo relationship calculate at 2.45GHz with a differing qualities coefficient of 1. The gather delay is found to be less than 1ns. At long last, the recreated and measured comes about are concurred superbly.

5.2 FUTURE WORKS

In this work the author is applied a study on designing a printed circuit antenna array on FR-4 substrate for MIMO applications at 2.45GHz, however, there are many future work can be subjected for the future research as following:

1- The proposed design is conducted based on an FR-4 substrate. However; as a future study, the author is attended to attempt the study on a flexible substrate to provide a dual resources of diversity and size reduction.

2- The antenna is fabricated using copper layer etched from an F4 substrate; however, a subject for a future research the use of printing technology based nanoscale conductive ink to fabricate the proposed antenna.
3- The author used the third order fractal structure, therefore, the subject for this research in the future to apply further orders to reach further bands including 5.8GHz for Wi-Fi applications.
4- The antenna is fed with a quarter wavelength transformer to provide excellent matching impedance, therefore, in the future a further transformer is subjected to be used for this design.
5- The antenna design will be fabricated on flexible substrate to suit the wearable applications.
6- The antenna is designed to provide one frequency band; however, the author wishes in the future to extend the antenna design to provide an ultra-wide band matching bandwidth over further frequency bands.

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