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THREE DIMENSIONAL CONTROL OF MACHINE
TOOLS BY USING MICROCOMPUTER

A MASTER'S THESIS

in

Electrical and Electronic Engineering
Gaziantep Engineering Faculty
Middle East Technical University

044168

By
İnan GÜLER
October 1985

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B. Kaftanoğlu
for Prof. Dr. Bilgin Kaftanoğlu

Director

I certify that this thesis satisfies all the requirements as a thesis for the degree of Master of Science in Electrical and Electronic Engineering.

M. Köksal
Assoc. Prof. Dr. Muhammet Köksal
Chairman of the Department

We certify that we have read this thesis and that in our opinion it is fully adequate, in scope and quality, as a thesis for the degree of Master of Science in Electrical and Electronic Engineering.

M. Köksal
Assoc. Prof. Dr. Muhammet Köksal
Supervisor

Examining Committee in Charge:

Assoc. Prof. Dr. Hüseyin FİLİZ (Chairman) *H.F. Filiz*

Assoc. Prof. Dr. Muhammet KÖKSAL *M. Köksal*

Assoc. Prof. Dr. Süleyman SARITAŞ *S. Saritaş*

Assoc. Prof. Dr. Bekir Sami YILBAŞ *B. Sami Yilbaş*

Asst. Prof. Dr. Erol AYRANCI *E. Ayrancı*

044168

ABSTRACT

THREE DIMENSIONAL CONTROL OF MACHINE TOOLS BY USING MICROCOMPUTER

GÜLER, İnan

M.S. in Electrical and Electronic Engineering Department

Supervisor: Assoc. Prof. Dr. Muhammet Köksal

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In this thesis, three dimensional control of a machine tool by using microprocessor is designed and constructed. In this system, an important part can be observed as drive circuits. These circuits are designed and fitted to other devices. The microcomputer that is used in this system is also designed and constructed. In a Computerized Numerical Control, (CNC) machine tool, tables should be able to follow any arbitrary path for shaping the workpiece in any form. In this work, linear motion is selected. Linear motion has a great importance in the path following ability of the cutting tool. This is the basic of the other motions. The main purpose here is to design a control unit which controls the tables of the machine tool. There are three tables, two of them are dependent each other, and one of them is independent from the others. After the testing of the system, satisfactory operation in open loop on no load conditions is obtained up to 1300 mm/min. feedrate in 0.05mm steps. In closed loop operation, due to damped oscillation of the motor's rotor the desired operation is not obtained.

Key words: Computerized Numerical Control, Linear motion, microprocessor, machine tool, drive circuit.

ÖZET
TAKIM TEZGAHLARININ ÜÇ BOYUTTA
MİKROBİLGİSAYAR İLE KONTROLU

GÜLER, İnan

Yüksek Lisans Tezi, Elektrik ve Elektronik Müh.Bölümü

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Bu tez çalışmasında, bir takım tezgahının üç boyutta mikroişlemci ile kontrolü dizayn edildi ve gerçekleştirildi. Sistemde önemli kısım olarak sürücü devreler göz önüne alınabilir. Bu devreler dizayn edilerek diğer cihazlarla uyumu sağlandı. Sistemde kullanılan mikrobilgisayarda bu çalışma içerisinde dizayn edildi ve gerçekleştirildi. Bilgisayarlı Nümerik Kontrollü bir takım tezgahında tablalar işlenecek parçayı herhangi bir şekilde işleyebilmek için değişik hareket yolları takip ederler. Bu çalışmada, lineer hareket seçilmiştir. Lineer hareket kesici tezgahlarda, önemli bir özelliğe sahiptir. Bu hareket diğer hareketlerin temelidir. Burada temel gaye takım tezgahının tablalarını kontrol eden kontrol birimini dizayn etmektir. İkisi birbirine bağımlı, diğeri ise bağımsız olmak üzere üç tabla vardır. Sistemin test edilmesi neticesinde açık çevrim yüksüz şartlarda 0,05 mm lik adımlarla 1300 mm/dak. besleme hızına kadar arzu edilen çalışma gerçekleştirildi. Kapalı çevrim çalıştırmada motorların roterlerindeki salınımlardan dolayı arzu edilen işlem elde edilemedi.

Anahtar sözcükler: Bilgisayarlı Nümerik Kontrol, Lineer hareket, mikroişlemci, takım tezgahı, sürücü devre.

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CHAPTER I

INTRODUCTION

1.1. General Information About Computerized Numerical Control (CNC) Systems

There is no doubt that Numerical Control (NC) of Machine Tools has been the mostly spoken objective in the Production Field, in the last 35 years. A gradual, disciplined approach to the design of NC of machine tools, coupled with intensified support from organized research, particularly with the rapid progress in integrated circuit technology has improved the quality of design in this field. The concept of numerical control ultimately extends far beyond the individual machine tool; it is a comprehensive form of organisation of internal data transmission and processing. The application of numerical control to machine tools has increased the speed and repetitive accuracy in production considerably, and the potential consequences of this to the world's productivity and economy are enormous [I] .

The rapidly progressing development of electronic components has, as in other areas of automation, strongly influenced the design of control systems. Thus new developments in CNC are determined extensively by the availability of new, very large scale integrated components such as microprocessors and semiconductor memories. The influence of the new technology is reflected in the scope of functions available, as well as in volume and price of

the controls. Between the years 1966 and 1976 the scope of functions in NC's was approximately doubled, the volume has been reduced to, a twentieth and the price to a quarter[SP].

1.2. Historical Background of CNC Systems

The first attempts to control machine tools numerically were initiated in 1950's and since then, it has been considerably developed. Today, there is a more general use of numerically controlled machine tool and a rapid increase can be expected as many more alternatives, in both price and facilities, are offered.

In a machine tool, when the relative positions of the workpiece and cutting tool are determined by the information derived from numerical values, it is said to be under numerical control. The information which is called program is suitably coded for storing on cards, paper or magnetic tape and once provided to the system, it is carried out automatically, in a predetermined sequence.

While making a workpiece on a machine tool, several judgements of speeds, feeds, mathematics and sometimes, even tool configuration, were machine tool operator's own responsibility. Moreover he had to manufacture the part by making intermittent measurements to ensure that he did not remove too much metal. With the introduction of numerical control, all or part of these features are performed, by the coded instructions, much more accurately, quickly and less scrap and rejects[I].

As with most technological advances, the sizes of control units decrease due to the evolutions in semiconductor field. Also tapes and tape readers are gradually phased out as interfaces connected directly from computers by pass them.

Since Computer Aided Manufacturing (CAM) has gained considerable importance, the manufacturing control system makes use of a

special NC machine (CNC) which receives its instructions directly from a computer. Because a conventional NC equipment is incapable of performing these tasks, a modified NC system is required. A CNC system uses a digital computer to replace some or all of the logical functions performed by the conventional machine control unit.

Since the development of the first numerically controlled machine tools in the early fifties, the inner structure and the outside appearance of numerical controls have changed decisively. Whereas, the first numerical controls were based on relays as the logic components, soon afterwards transistors, and some years later integrated circuits were employed to cope with the ever extending control functions. From 1972 onwards minicomputers were available which were suitable for use in NC. Today, new developments of numerical controls are being determined by the availability of inexpensive highly integrated components, such as microprocessors [HU].

1.2.1 Computer Control Types of Machine Tools

There are mainly two types of computer control of machine tools.

The first one is the Computerized Numerical Control, CNC, which is an NC system where the dedicated computer is used to perform some or all of the basic NC functions for a single machine tool (or a number of machine tools if the total number of controlled axes is small) in accordance with control programs stored in the read-write memory of the computer.

Generally, the control and operating programs are read into the computer via the tape reader and held in the memory. The operating program is the logic of the control system which defines the features of the control, whereas the NC program is the set of

cycle instructions which define the movements of the machine tool [RA], [WE].

First CNC Systems were equipped with minicomputers and core memories. Today's CNC systems employ a microcomputer and semiconductor memories as the central controlling unit. This CNC, which is often termed as a "microprocessor CNC", offers the manufacturer the opportunity to produce the hardware at a reasonable price independently the machine application such as turning, boring, milling or punching. When the functions of the control are divided over several microprocessors, the term "multi-microprocessor control" is employed [SP].

The second type of computer control of machine tools is Direct Numerical Control, DNC, which is the use of a computer to distribute data to machine tool controllers in a multiple machine tool arrangement. The important difference between DNC and CNC is the replacement of a dedicated computer by a larger computer that manages many machines on a time-shared basis [SP], [MI].

1.3. Design Problem of CNC Systems

The purpose in the design is the microcomputer control of a machine tool table in a closed-loop system. In a CNC machine tool, tables should be able to follow any arbitrary path for shaping the workpiece in any form. Generally in practical situations, these paths consist of some combinations of linear, circular, elliptical, hyperbolic, segments or they are approximated to these forms [BR], [I].

Linear motion has a great importance in the path following ability of the cutting tool. This is the basic of the circular, elliptical, hyperbolic, etc. motions. It is possible to obtain these motions with a predetermined accuracy using piecewise linear

motion approximations $[DAN]$, $[SO]$, $[CH]$.

In this work, the main purpose is to design a control unit which controls the tables of a machine tool. There are three tables, two of them are dependent each other, and one of them is independent from the others. The general block diagram of the system is shown in Fig 1.1.

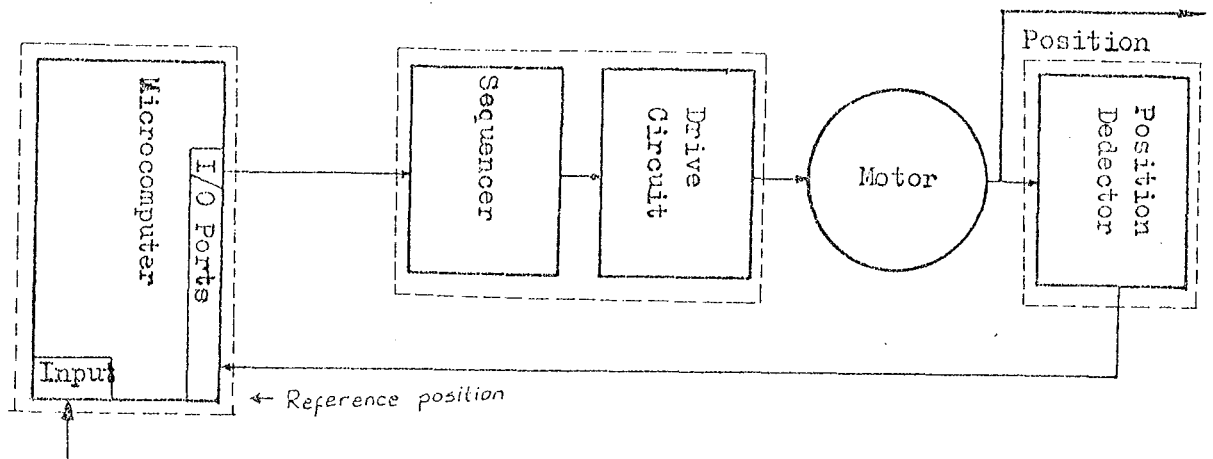


Fig 1.1. General block diagram of computer control

Microcomputer, that is used in this work is also designed and constructed. More detailed information about the microcomputer is given in Chapter II. The informations about hardware of the designed system are given in Chapter III. In Chapter IV, the software of the system is discussed. In Chapter V, results, discussions and recommendations are expressed.

This work was initiated in the Electrical and Electronic Engineering Department of METU, it is also supported by Electrical and Electronic Engineering Department of Erciyes University.

CHAPTER II

MICROCOMPUTER -- ALP 85

2.1. General

Microcomputer is making an impact on every aspect of our lives; and soon it will play a significant role in the daily functioning of all industrialized societies. The basic structure of the microcomputer is no different from any other computer. In the 1960's, computers were accessible and affordable only to large corporations, big universities, and government agencies. Because of advances in semiconductor technology, the million-dollar computing capacity of the 1960's is now available for less than ten dollars in an integrated circuit called microprocessor. A computer which is designed by using a microprocessor is called a microcomputer [GA]. The special microcomputer designed in this thesis work is called ALP 85.

2.1.1 Computer Technology

In the last twenty-five years, semiconductor technology has undergone unprecedented changes. Integrated circuits (ICs) appeared on the scene at the end of the 1950's following the invention of the transistor. In an integrated circuit, an entire circuit consisting of several transistors, diodes, and resistors is contained on a single chip. In the early 1960's, logic gates known as the 7400 series were commonly available as ICs, and the technology of integrating the circuits of a logic gate on a single chip became known as Small-Scale Integration (SSI). As semiconductor technology advanced, more than a hundred gates were

fabricated on one chip; this was called Medium-Scale Integration (MSI). A typical example of MSI is a decade counter (7490). Within a few years, it was possible to fabricate more than a thousand gates on a single chip; this came to be known as Large-Scale Integration (LSI). Now we are in the era of Very-Large-Scale Integration (VLSI) and Super-Large-Scale Integration (SLSI). The lines of demarcation between these different scales of integration are ill-defined and arbitrary.

As the technology moved from SSI to SLSI, the face of computer changed. Initially, computers were built with discrete logic gates (SSI). As more and more logic circuits were built on one chip using LSI technology, it became possible to build the whole central processing unit, (CPU), with its related timing function on a single chip. This came to be known as the microprocessor, and a computer built with a microprocessor is known as a microcomputer. This distinction may soon disappear, however, as the computing power of the microprocessor approaches that of the CPUs of the traditional large computers. Early microcomputers were built with a 4-bit microprocessor. Now a 64-bit microprocessor is being used in some prototype computers. Even if they are built with a microprocessor, it is meaningless to classify them as microcomputers [GA].

2.1.2 Microcomputer Organization

Figure 2.1 shows a simplified but formal structure of a microcomputer. It includes four components: microprocessor, input, output, and memory (read/write memory and read only memory) units. These components are organized around a common communication path called a bus. The entire group of components is called a system or a microcomputer system, while the components are called subsystems.

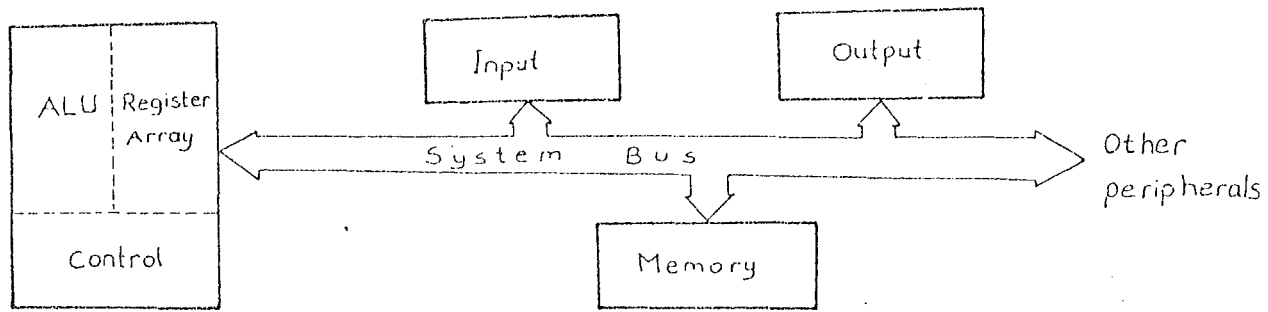


Fig. 2.1. Block diagram of a microcomputer.

At the outset, we will differentiate between the terms microprocessor and microcomputer because of the common misuse of these terms in popular literature. The microprocessor is one component of the microcomputer. On the other hand, the microcomputer is a complete computer similar to any other computer, except that the CPU functions of the microcomputer are performed by the microprocessor. Similarly, the term peripheral is used for input/output devices; however, occasionally memory is also included in this term [GA], [BR].

2.2. Designed Microcomputer - ALP 85

The designed microcomputer is constructed on Euro-card format in which the dimensions are 100mmx160mm. It is an 8-bit microcomputer based on the Intel 8085 microprocessor chip. It has 2K bytes monitor stored in EPROM. 4K bytes RAM and 2K bytes EPROM are available for the user programs. The microcomputer uses Intel 8155 as input/output ports. Intel 8155 has 3 ports: 2 programmable 8-bit I/O ports and 1 programmable 6-bit I/O port. The microcomputer has also a keyboard and 8 seven-segment displays. The communications between keyboard and displays are achieved by using Intel 8279 keyboard and display controller chip. An EPROM programmer and a magnetic tape interface are available on CPU card.

The block diagram of the designed microcomputer is given in Figure 2.2 [ECB].

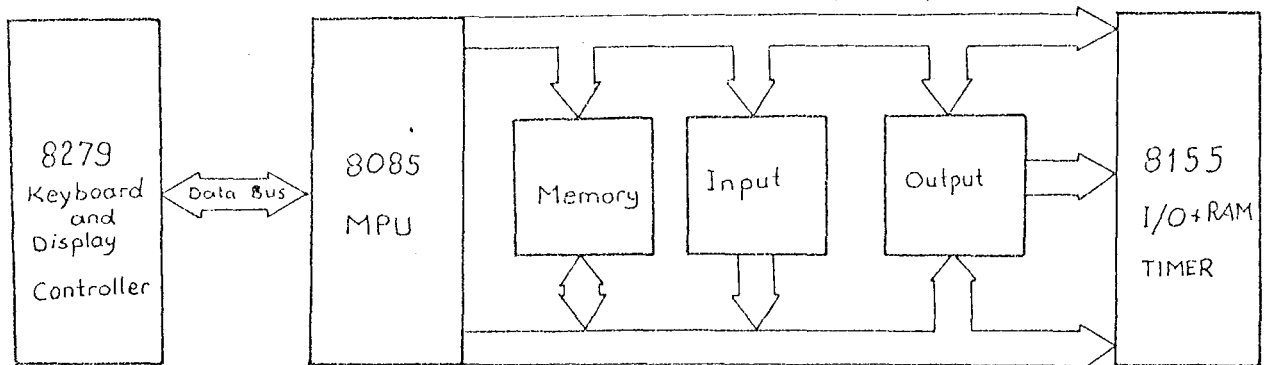


Fig. 2.2. Block diagram of the ALP 85 microcomputer.

2.3. Components of ALP 85

The various components of the microcomputer shown in Figure 2.2 and their functions are described in the following paragraphs.

2.3.1. 8085 MPU

The term Micro-Processing Unit (MPU) is similar to the term Central Processing Unit (CPU) used in traditional computers. We define the MPU as a device or a group of devices that can communicate with peripherals, provide timing signals, direct data flow, and perform computing tasks as specified by the instructions in memory. The unit will have the necessary lines for the address bus, the data bus, and perform computing tasks as specified by the instructions in the memory. The unit will have the necessary lines for the address bus, data bus, and the control signals, and would require only a power supply and a crystal to be completely functional.

Using this description, the 8085 microprocessor can almost qualify as an MPU, but with the following two limitations:

1. The low-order address bus of the 8085 microprocessor is

multiplexed (time-shared) with the data bus. The buses need to be demultiplexed.

- Appropriate control signals needed to be generated to interface memory and I/O with the 8085 [IN],[LE].

The 8085 is an 8-bit general purpose microprocessor capable of addressing 64Kbytes of memory. The device has forty pins, requires a +5V single power supply, and can operate with a 3-MHz single-phase clock. The 8085 is an enhanced version of its predecessor, the 8080A; its instruction set is upward-compatible with that of the 8080A, meaning that the 8085 instruction set includes all the 8080A instructions plus some additional ones. Programs written for the 8080A will be executed by the 8085, but the 8085 and the 8080A are not pin compatible.

Figure 2.3 shows the logic pinout of the 8085 microprocessor. All the signals or signal ports can be classified into six groups: (1) address bus, (2) data bus, (3) control and status signals, (4) power supply and frequency signal, (5) interrupts and peripheral initiated signals, and (6) serial I/O ports.

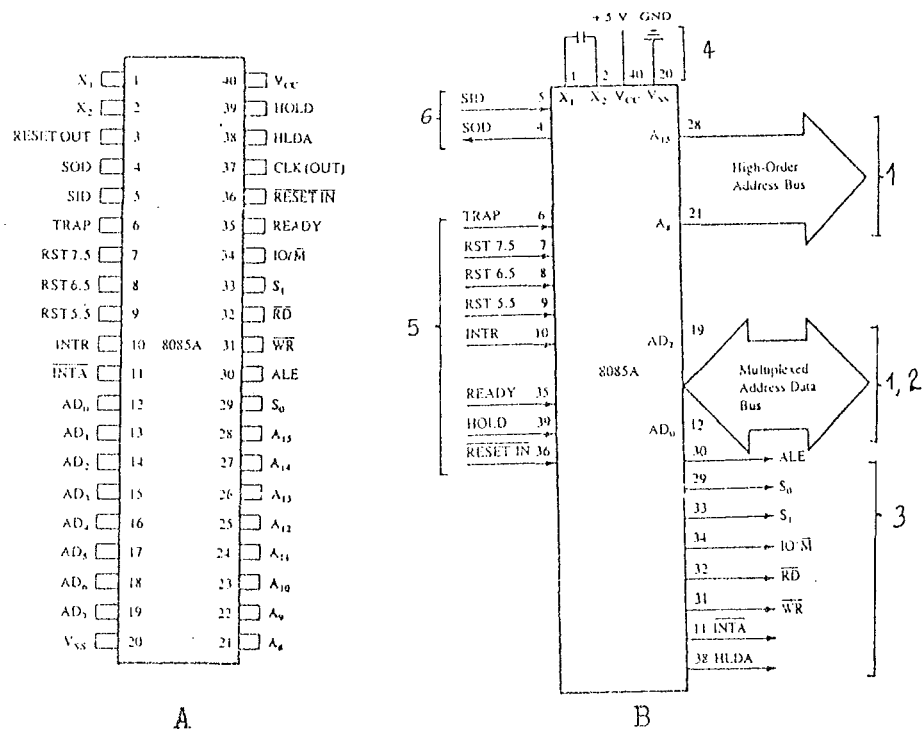


Figure 2.3. Microprocessor pinout and signals.

2.3.2. 8155 I/O Ports--RAM--TIMER

The Intel 8155H are RAM and I/O chips implemented in N-channel, depletion load, silicon gate technology (HMOS), to be used in the 8085A microprocessor system. The RAM portion is designed with 2048 static cells organized as 256x8. They have a maximum access time of 400 ns to permit use with states in 8085A CPU. The 8155A has maximum access times of 330 ns for use with the 8085A CPU.

The I/O portion consists of three general purpose I/O ports. One of the three ports can be programmed to be status pins, thus allowing the other two ports to operate in handshake mode.

A 14-bit programmable counter/timer is also included on the chip to provide either a square wave or terminal count pulse for the CPU system depending on timer mode. The functional block diagram and pin configuration of 8155 is shown in Figure 2.4.

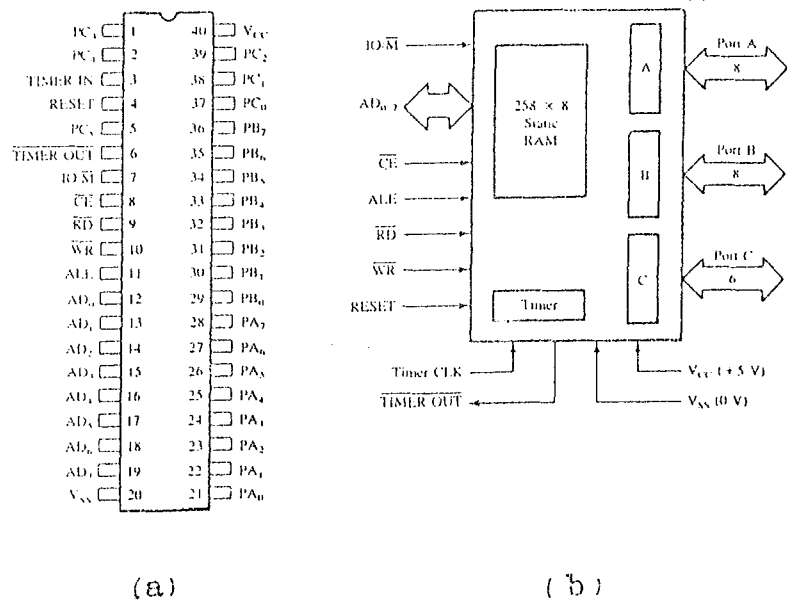


Figure 2.4. Pin configuration and functional block diagram of 8155

- a) Pin configuration,
- b) Block diagram.

The 8155 contains the followings:

- 2K Bit Static RAM organized as 256x8,
- Two 8-bit I/O ports (PA and PB) and one 6-bit I/O port (PC),
- 14-bit timer-counter

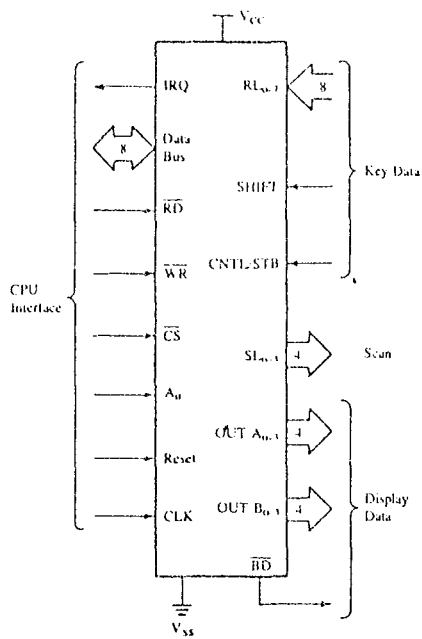
The IO/ \bar{M} (IO/Memory Select) pin selects either the five registers (Command, Status, PA₀₋₇, PB₀₋₇, PC₀₋₅) or the memory (RAM) portion.

The 8-bit address on the Address/Data lines, Chip Enable input \bar{CE} , and IO/ \bar{M} are all latched on-chip at the falling edge of ALE [IN].

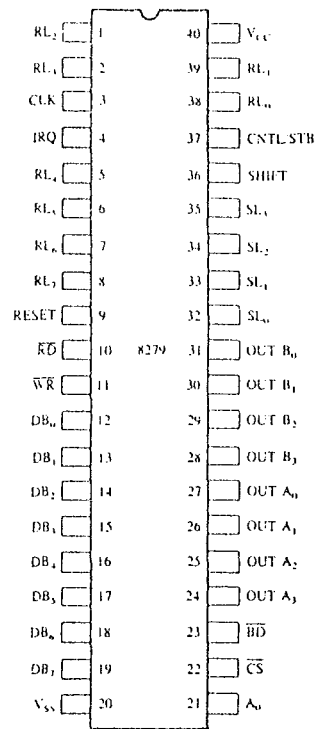
2.3.3. 8279 Programmable Keyboard/Display Controller

The Intel 8279 is a general programmable keyboard and display I/O interface device designed for use with Intel micro-processors. The keyboard portion can provide a scanned interface to a 64-contact key matrix. The keyboard portion will also interface to an array of sensors or a strobed interface keyboard, such as the hall effect and ferrite variety. Key depressions can be 2-key lockout or N-key rollover. Keyboard entries are debounced and strobed in an 8-character FIFO. If more than 8 characters are entered, overrun status is set. Key entries set the interrupt output line to the CPU.

The display portion provides a scanned display interface for LED, incandescent, and other popular display technologies. Both numeric and alphanumeric segment displays may be used as well as simple indicators. The 8279 has 16x8 display RAM which can be organized into dual 16x4. The RAM can be loaded or interrogated by the CPU. Both right entry, calculator and left entry typewriter display formats are possible. Both read and write of the display RAM can be done with auto-increment of the display RAM address. Logic symbol and pin configuration of 8279 is shown in Figure 2.5.



(a)



(b)

Figure 2.5. a) Logic symbol and b) Pin configuration of 8279

Since the main purpose is not to design and the construction of microcomputer, these explanations about the components are enough. More detailed information about these components can be obtained in the references [IN], [BR], [GA].

The printed circuit board, components, and a photograph of CPU card are shown in Appendix A.

CHAPTER III

HARDWARE OF THE DESIGNED SYSTEM

3.1. Purpose and Design Problem

Purpose of the design problem is the closed loop control of machine tool tables in three dimensional linear motion with a microcomputer. There are three axis in the machine tool. In the scope of controlling the machine tool, the x and y axes are dependend on each other while z axis is independent from x and y axis. If the system is started to work, firstly z-axis value is given to the microcomputer and then the other values are given to the microcomputer.

In the designed system, the maximum dimensions of the workpiece is taken as 100mmx200mmx1500mm in view of the available machine tool. The software can easily be modified for other dimensions.

Machining the workpieces made of different materials requires different feedrates. It is reported and found that practically possible feedrate range is between 0 and 1500mm/min. Therefore, in the designed system, a performance in this range is sought (OR).

In the designed system, it is assumed that the smallest linear movement (one step) will be equal to the accuracy tolerance. This accuracy tolerance is taken as 0.05 mm. In the system, stepping motors are used to give the movements to tables, a lead screw should be used to translate the rotational movement of the motors to linear movement of the tables. Stepping motors have quantized angular movement characteristics. The smallest rotation

is the step angle of this motor. Gear boxes will be employed to obtain smaller step sizes. It is decided that the gear ratio must be such that the tables make a 0.05 mm linear motion (accuracy tolerance) corresponding to each step angle of the motors.

CNC machine tools can be controlled into two ways; (i) Point to Point Control and (ii) Continuous Path Control.

In point to point control, there is no functional relationship between the movements along the different coordinates. This group includes drilling machines, boring machines, milling machines etc. It is further, divided into two sub-groups: Positioning control and Straight-line control. In positioning control, the tool is never in contact with the workpiece while the control moves the slide to individual discrete points such as from A to B in Figure 3.1. A most typical example is the coordinate drilling machine. In straight-line control, the tool is in contact with the workpiece while the slide is moved along the coordinates. The motion is however always parallel to one of the axes of the machine tool. An example to this type of control is milling machine with traverse tables which is shown in Figure 3.2.

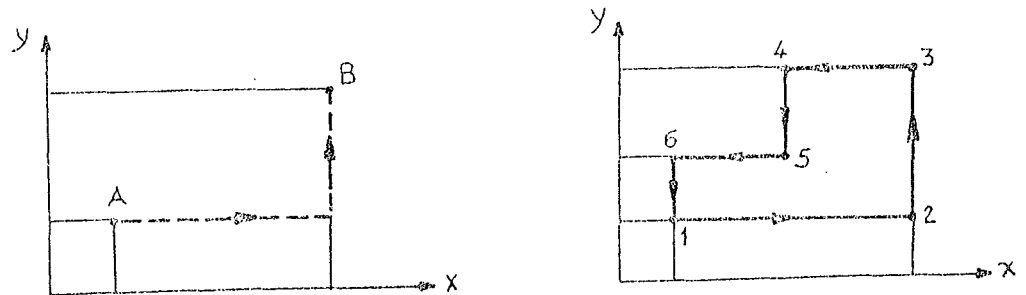


Figure 3.1. Positioning control Figure 3.2. Straight-line control

In continuous path control, there is a functional relationship between the movements along different coordinates. Hence the tool can be guided along a path of any desired shape as shown in Figure 3.3. A typical example is the profile milling machine [1].

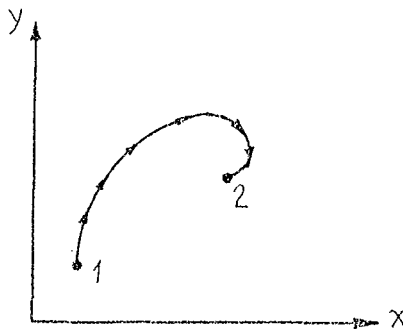


Figure 3.3. Continuous path control

To achieve this type of motion, continuous control of the slide positions is required. Moreover an interpolator or a curve generator is included, which has two main functions to perform:

1. To interpolate between fixed data points or generate a particular type of curve,
2. To resolve the feedrate into the component velocities of the slides.

3.2. Principles of the Solution and Available Choices

3.2.1. Stepping Motor

The characteristic property of stepping motors is the stepwise rotation of the motor shaft. One revolution of the shaft is composed of an exactly defined number of angle steps, which is dependent upon the design of the motor. The step is the process involving a rotation of the motor shaft by the step angle. Stepping motors are used chiefly in the area of control technology which includes e.g. also office machines. For this reason, it only makes sense to compare motors designed for the same tasks. This comprises small rotating field motors, asynchronous motors, and DC motors.

In the course of the development of digital systems, step motors came into increasing use. They are preferred because they exhibit the following combination of the following characteristics:

1. Positioning accurate to one step without feedback, by imposition of a definite number of control pulses,

2. High torque at small angular velocities,
3. In the stationary state with excitation, a high holding torque which causes self brake action.

In most cases, problems incurred in the use of step motors have been of dynamic nature. A step motor may be described with good approximation as a linear system of second order with oscillatory behavior. This system is modified by the coupling of external inertial and load moments. As for all drive types, this gives rise to three essential problem areas [BER]:

1. Start problems,
2. Brake problems,
3. Oscillation problems.

The stepping motors can be classified in three categories: Permanent Magnet Stepping Motors, Variable Reluctance Stepping Motors and Hybrid Stepping Motors. In general, hybrid stepping motors are more efficient and are available in higher power models. Also they have longer detent torque (the maximum steady torque that can be applied to the shaft of an unenergized stepping motor without causing continuous rotation) than the others.

A hybrid stepping motor has a permanent magnet mounted on the rotor. The main flux path for the magnet flux, shown in Figure 3.4., lies from the magnet N-pole, into a soft-iron end cap, radially through the end-cap, across the air-gap, through the stator poles of section X, axially along the stator back-iron, through the stator poles of section Y, across the air-gap and back to the magnet S-pole via the end-cap.

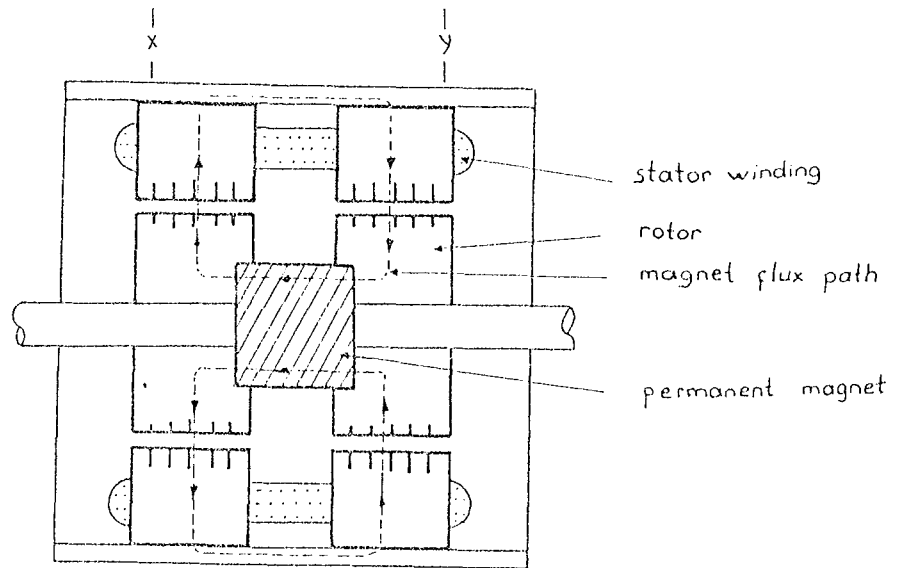


Figure 3.4. Cross-section of a hybrid motor parallel to the shaft.

There are typically eight stator poles, as in Figure 3.5; and each pole has between two and six teeth. The stator poles are also provided with windings which are used to encourage or discourage the flow of magnet flux through certain poles according to the rotor position required. Two windings are provided and

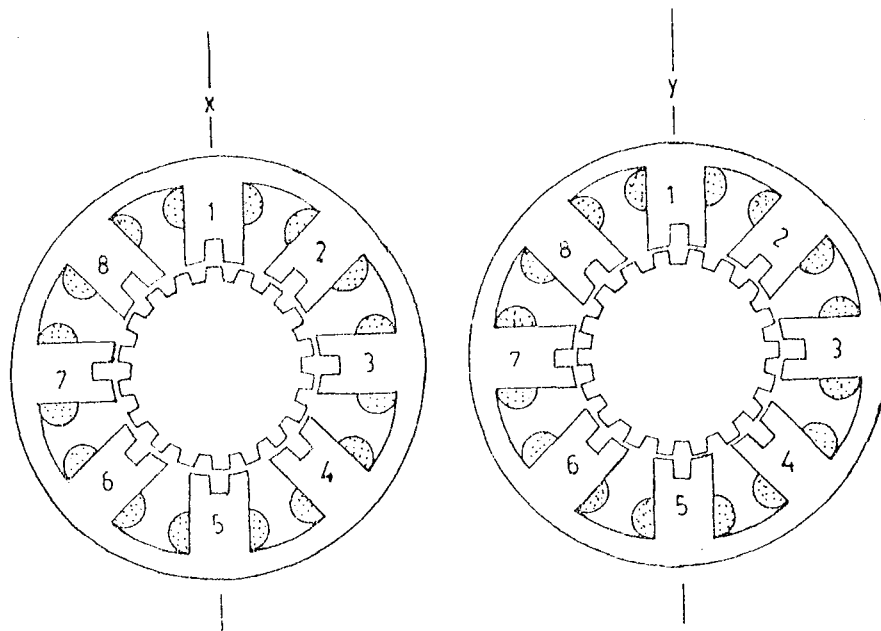


Figure 3.5. Cross-section of a hybrid motor perpendicular to the shaft.

each winding (phase) is situated on four of the eight stator poles; winding A is placed on poles 1,3,5,7 and winding B is on poles 2,4,6,8. Successive poles of each phase are wound in the opposite sense, e.g. if winding A is excited by positive current the resultant magnetic field is directed radially outward in poles 3 and 7, but radially inward in poles 1 and 5. A similar scheme is used for phase B and the situation for the whole machine is summarized in Table 3.1.

Winding	Current Direction	Direction of pole Radially outward	magnetic field Radially inward
A	Positive	3,7	1,5
A	Negative	1,5	3,7
B	Positive	4,8	2,6
B	Negative	2,6	4,8

TABLE 3.1. Position pole magnetic field.

Hybrid motor is a natural choice for applications requiring a small step length and high torque in a restricted working space. Therefore, in the designed system, hybrid stepping motor is selected. The step angle of the selected hybrid motor is 1.8° and it is 2-phase full-step hybrid motor. Characteristics curves of the hybrid motor are shown in Figure 3.6, and some specifications are given in Table 3.2 [AC], [OR].

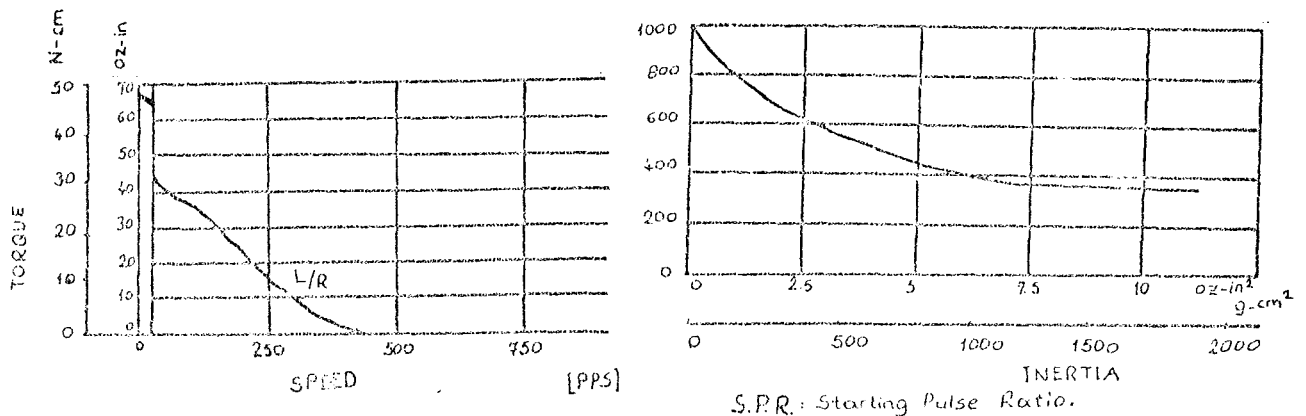


Figure 3.6. Characteristics curves of PH 265-03 hybrid motor;
a) Speed v.s. Torque char. b) Inertia v.s. S.P.R. char.

Motor Type	Voltage	Current per phase	Holding Torque		Resistance per phase	Inductance per phase
			oz-in	N-cm		
Single shaft	V	A/phase	oz-in	N-cm	ohm/phase	mH/phase
PH265-03	24	0.21	58.3	41.2	114	14.4
Rotor inertia : 0.6 oz-in ² (110 g-cm ²), Weight : 1.2 lbs (0.54 kg)						

Table 3.2. Specifications of PH 265-03 stepping motor.

3.2.2. Drive Circuit

The control signals for a stepping motor system are invariably low power, e.g. TTL digital integrated circuits provide 5V at 18 mA, whereas a typical variable-reluctance stepping motor giving a torque of 1.2 Nm has a rated winding excitation of 5V and 3A. Therefore if the drive circuit is based on conventional bipolar junction transistor switches, the controller must be interfaced to the motor via several stages of switching amplification [AC].

There are two types of drive circuits:

(i) Unipolar Drive Circuit and (ii) Bipolar Drive Circuit.

3.2.2.1 Unipolar Drive Circuit

A simple unipolar drive circuit suitable for use with a three-phase variable-reluctance stepping motor is shown in Figure 3.7. Each phase winding is excited by a separate drive circuit, which is controlled by a low-power phase control signal.

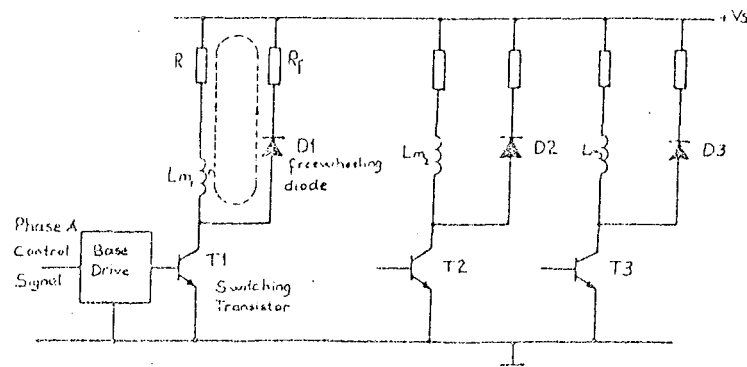


Figure 3.7. Three-phase unipolar drive circuit.
(----- freewheeling current path).

This control signal may require several stages of switched amplification before it attains the power level required at the base of the phase transistor.

The phase winding is excited whenever its switching transistor is saturated by a sufficiently high base current. Under these conditions the full d.c. supply voltage is applied across the series combination of phase winding and forcing resistance, since the voltage drop across the saturated transistor is small (typically 0.1 V). The d.c. supply voltage (V_s) is chosen so that it produces the rated winding current (I) when applied to the total phase circuit resistance, which is equal to the sum of the phase winding (r) and forcing (R) resistances:

$$V_s = I(r+R) \quad (3.1)$$

In general the phase winding has a considerable inductance, so its natural electrical time constant (L/R) is long. The build-up of phase current to its rated value would be too slow for satisfactory operation of the motor at high speeds. By adding the forcing resistance, with a proportional increase in supply voltage, the phase electrical time constant can be reduced, enabling operation over a wider speed range.

Another consequence of the finite phase winding inductance is that the phase current cannot be switched off instantaneously. If the base drive of the switching transistor was suddenly removed a large induced voltage would appear between the transistor collector and emitter, causing permanent damage to the drive circuit. This possibility is avoided by providing an alternative current path-known as the freewheeling circuit-for the phase current. When the switching transistor is turned off the phase current can continue to flow through the path provided by the freewheeling diode and freewheeling resistance.

delivered to the load. Stepping motor drive circuit accepts sequence signals from sequencer and converts them to the proper format for driving the motor windings.

Since the motor rotates in response to a changing pattern of interactions between the rotor magnetic field and the stator magnetic field, a sequencer which creates the proper sequence should be designed. The outputs of the sequencer will be the inputs of the drive circuit. The sequencer is explained in the following section.

In the designed system, a maximum of 1500 mm/min feedrate is desired. This corresponds to 500 steps/sec or 150 rev/min or 2.5 rev/sec. The stepping motors used in the system have 1.8° step angle. This means 200 steps per revolution. That is, one revolution of the motor will create 200×0.05 mm (=10 mm) linear motion. The designed drive circuit is shown in Figure 3.9 satisfy the system torque and speed requirements [BRE].

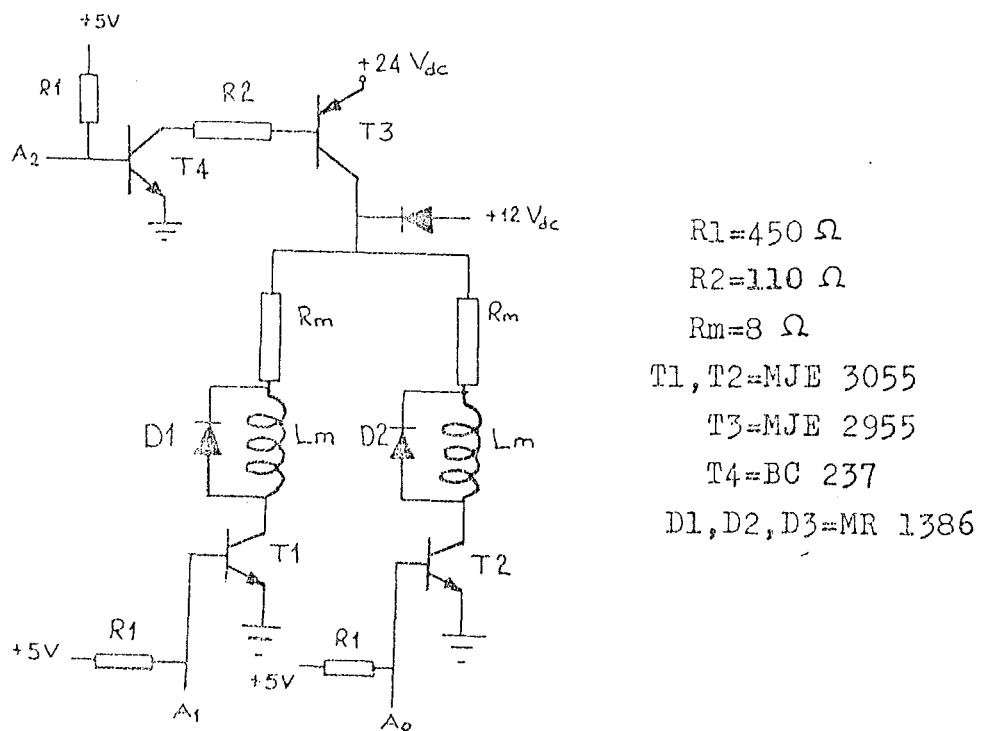


Figure 3.9. Designed drive circuit for stepping motor

The phase current if is established at its rated value then the maximum voltage ($V_{ce\ max}$) across the switching transistor occurs in the instant after the transistor switch is opened. The current (I) has not started to decay and flows through the freewheeling resistance (R_f), so the maximum collector-emitter voltage (neglecting the forward voltage drop across the freewheeling diode) is:

$$V_{ce\ max} = V_s + R_f I \quad (3.2)$$

The phase current therefore decays in the freewheeling circuit and the magnetic energy stored in the phase inductance at turn-off is dissipated in the freewheeling circuit resistance (winding+forcing+freewheeling) resistances [AC], [OZ], [OR].

3.2.2.2. Bipolar Drive Circuit

One phase of a transistor bridge bipolar drive circuit, suitable for use with a hybrid or permanent-magnet stepping motor, is shown in Figure 3.8. The transistors are switched in pairs according to the current polarity required. For positive excitation of the phase winding transistors T1 and T4 are turned on, so that the current path is from the supply, through transistor T1 to the phase winding and forcing resistance, then through transistor T4 back to the supply. In the opposite case the transistors T2 and T3 are turned on so that the current direction in the phase winding is reversed.

The four switching transistors in the bridge require separate base drives to amplify the two (positive and negative) phase control signals. In the case of the 'upper' transistors (T1 and T2) the base drive must be referred to the positive supply rail, which may be at a variable potential. For this reason the phase control signals to these upper base drives are often transmitted via a stage of optical isolation.

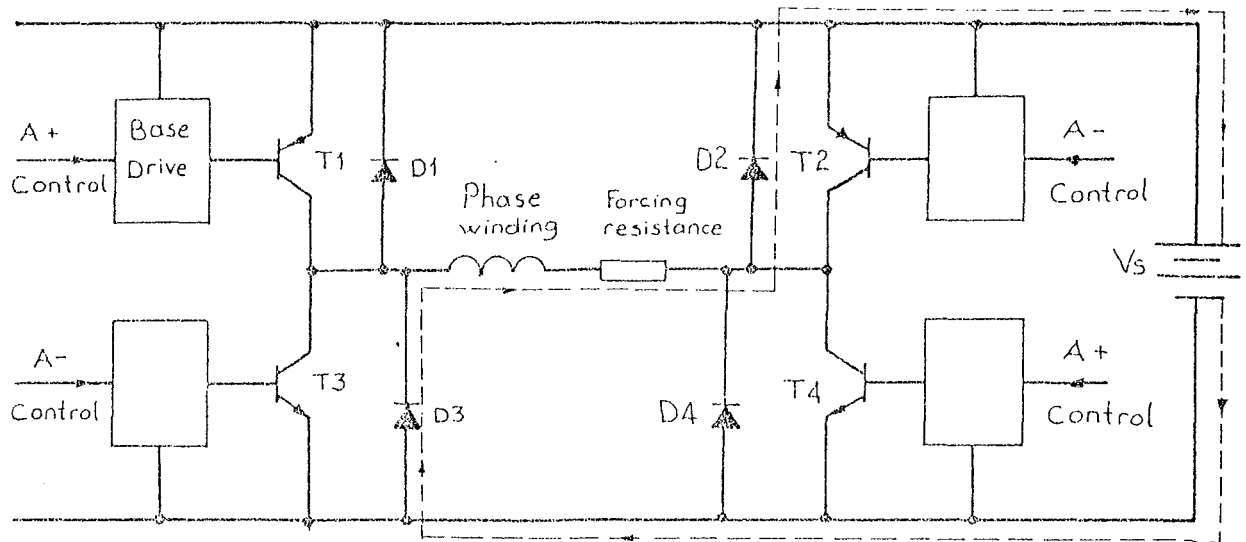


Figure 3.8. One phase of a transistor bridge bipolar drive circuit (----- freewheeling current path after T1 and T4 turn off).

A bridge of four diodes, connected in reverse parallel with the switching transistors, provides the path for freewheeling currents. In the illustration of Figure 3.8, the freewheeling current path, via diodes D2 and D3, corresponds to the situation immediately after turn-off of transistors T1 and T4. The freewheeling path includes the d.c. supply and therefore some of the energy stored in the phase winding inductance at turn-off is returned to the supply. The consequent improvement in overall system efficiency represents a significant advantage of the bipolar bridge drive over the unipolar drive and for this reason most large (greater than 1KW) stepping motors, including variable-reluctance types, are operated from bipolar drive [NG] [AC].

In the designed system, unipolar voltage drive circuit is selected. This is a simple method of driving stepping motors. However, this type of driver with selected component values can give the required torque and the speed of the designed system.

Drive circuit design is one of the most important aspects of a stepping motor system. The overall system performance is heavily dependent upon the drive system, not only on the power

3.2.3. Sequencer

The sequencer creates the proper sequence and pattern of states in response to a serial pulse train. Microcomputer generates this serial pulse train and this created proper sequence (by the sequencer) is fed to the drive circuit to make the motor rotate.

A sequencer has two inputs and four outputs. Its inputs are the direction and the pulse inputs. 0V or 5V is given to the direction input to determine the direction of the motor rotation. Each pulse given to the pulse input makes the motor rotate one step in the determined direction. The outputs of the sequencer are used in switching of the drive circuit to energize the windings in proper sequence. The designed sequencer circuit and its output waveform are shown in Figure 3.10 [MA].

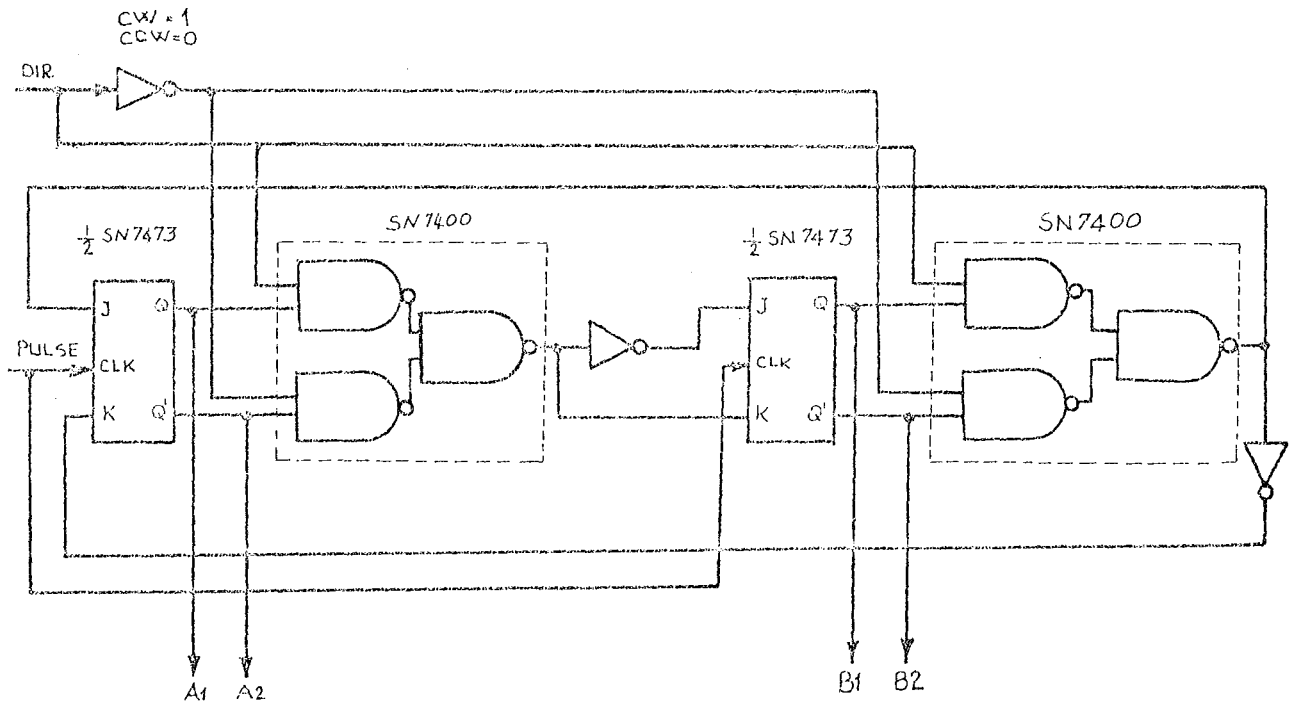


Figure 3.10. a) Sequencer.

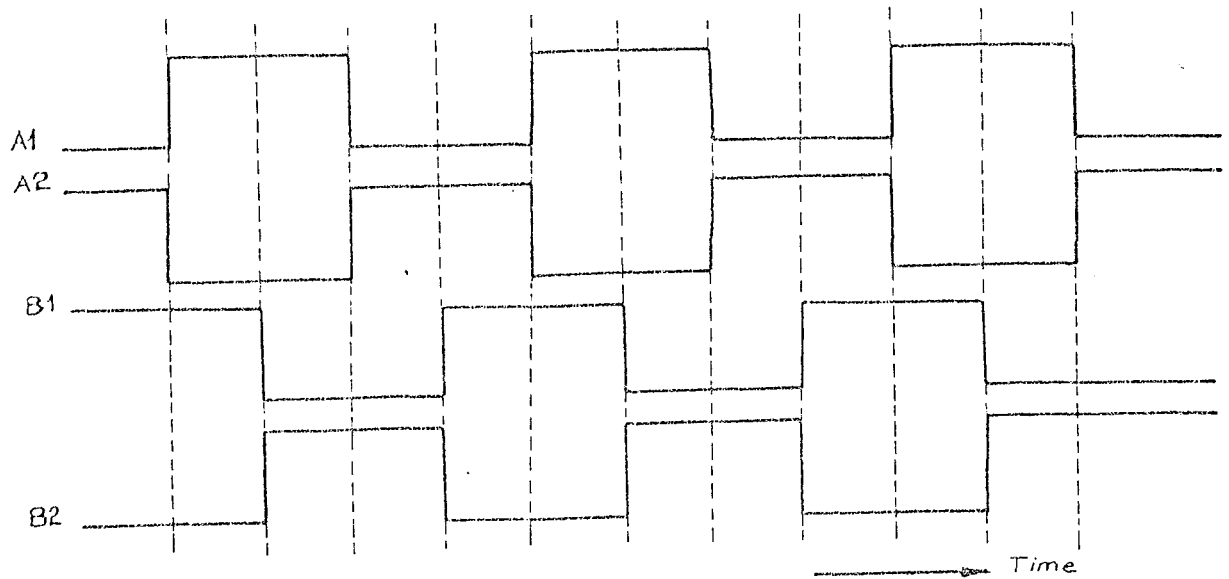


Figure 3.10. b) Output of sequencer

3.2.4 Position Transducer

Position transducers are explained as incremental and absolute devices separately.

3.2.4.1 Digital Incremental Position Transducers

The digital methods of controlling machine tools depend upon the generation of a series of pulses by the motion of the table. These pulses are generated either by rotary or linear position transducers.

In Incremental Rotary Devices, electromechanical methods, using springs or brushes to sweep the conducting and nonconducting areas on a disc, for scanning are rarely used nowadays because of their tendency to faulty operation. The most common method of generating electrical pulses is by the variation of the intensity of light falling on a photocell. A disc called digitizer, having equally spaced, alternate opaque and transparent segments suits for this purpose as shown in Figure 3.11a. The digitizer is placed between a light source and a photocell in which a flow of current is induced, as shown in Figure 3.11b.

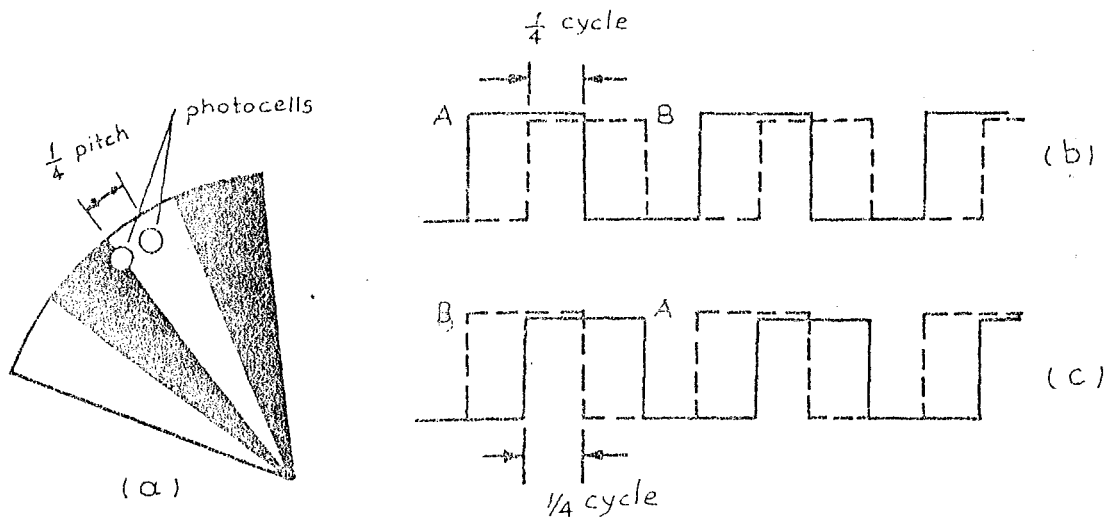


Figure 3.12. Photocells and their outputs;

- a) Positioning of two photocells,
- b) Outputs of photocells A and B in clockwise rotation.
- c) Outputs of A and B in counter clock wise rotation.

Another way to determine the direction of rotation would be to code the digitizer specially and place the photocells as shown in Figure 3.13. In clockwise rotation after they are both illuminated, A will close while B receives light, whereas in opposite direction both will be dark, as illustrated in Table 3.3.

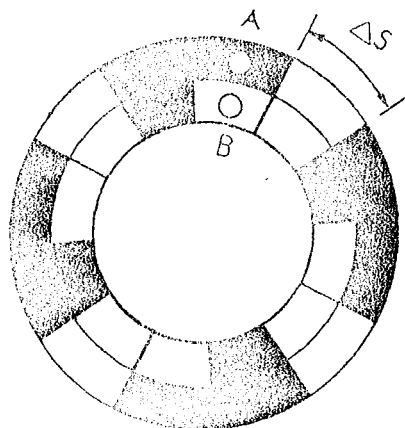


Figure 3.13. Coded digitizer.

Thus a logic circuit may easily detect the direction of rotation.

<u>Clockwise rotation</u>		<u>Counterclockwise rotation</u>		
<u>A</u>	<u>B</u>	<u>A</u>	<u>B</u>	
1	1	1	1	
0	1	0	0	
0	0	0	1	1: Illuminated
1	1	1	1	0: dark

Table 3.3. Photocells according to direction of rotation.

In incremental Linear Devices, it is again not practical to use electromechanical (by brushes and segments) or inductive scanning for small displacement elements " ΔS ", which is the case in machine tool practice. Therefore uncoded linear scales are best scanned by photoelectric means either by transmitted light method or reflected light method.

In the first method a linear scale, consisting of a glass plate upon which a number of equispaced lines are engraved, is fixed to the slide with a light source and a photocell on either side. The number of lines determines the accuracy obtainable and the process is similar to that of a rotary system. Another possibility is to introduce an auxiliary grating and making use of Moire Fringes to determine the passage of lines more easily.

As illustrated in Figure 3.14, the lines on both gratings are spaced by ΔS , and rotation of the auxiliary grating by " S " produces a Moire Fringe Pattern. One of the gratings is fixed while the other is connected to the slide. Depending on the direction of motion of the machine slide, the fringes will appear

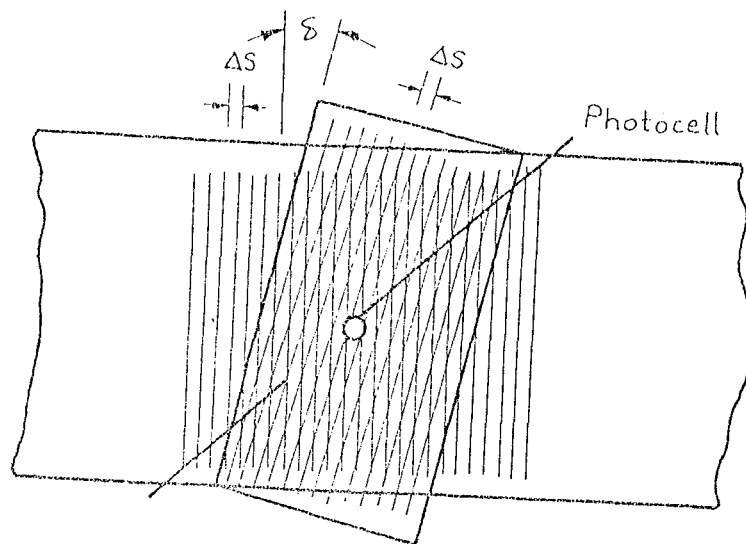


Figure 3.14. Moire Fringes with an auxiliary grating.

to move either upward or downward. The intensity of light seen by the photocell will vary sinusoidally, as the gratings are moved relative to each other. The sinusoidal output of the photocell is either counted directly or a phase comparison method may be used to determine the distance travelled.

The reflected light method bears the same principle, with the difference that instead of a glass scale it utilizes polished stainless steel scale on which the light is reflected. Though this method has the advantage that the easily broken, dirt attracting glass scale is avoided, it requires a complicated optical system. In addition to that it has a less optical resolution, due to the increased spacing of the graduations for reliable operation[I].

3.2.4.2 Digital Absolute Position Transducers

To measure the position absolutely relative to a predetermined datum, in addition to the incremental grating track with the $2^0 \times \Delta S$ grating, further tracks with binary gratings $2^1 \times \Delta S$, $2^2 \times \Delta S$, $2^3 \times \Delta S$ etc. are introduced. However, each track is considered separately and scanned by different scanners. Again

photoelectric scanning is found to be more suitable due to the previously stated reasons.

In absolute rotary devices, the digitizer disc may be constructed to give a binary output. It is then said to be encoded. Figure 3.15 shows segments that will give the decimal digits 0 to 9 in the binary code. The subdivision ΔS of the disc determines the resolution obtainable. Errors in the scanning of the

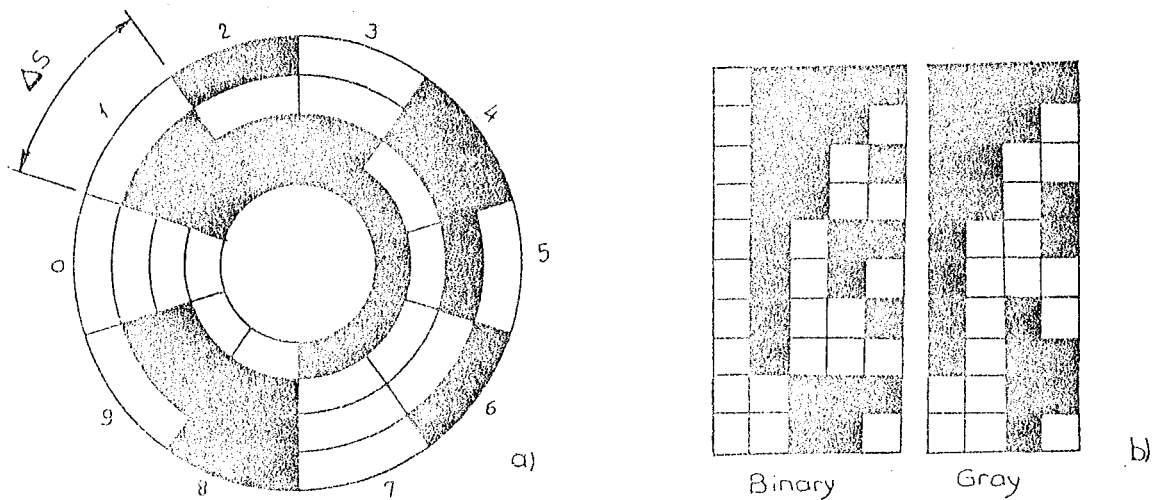


Figure 3.15. a) Absolute digitizer and b) Codes.

encoded disc are possible due to slight misalignment or to the ambiguity of signal when the disc comes to rest, with the scanner on an edge position. An alternative scan may be used with two sets of photocells displaced on the disc, the two outputs being sorted out by logic circuitry. Several alternative codes may also be used, of which the Gray Code is popular as shown in Figure 3.15b. In this code there is only one transition point for each change of digit.

Rotary discs are either made by increasing the number of tracks and employing precision gearing in their connection to the appropriate elements (however scanning becomes more complex and expensive) or connected to each other by gearing as shown in Figure 3.16 and zonal scanning is used [1].

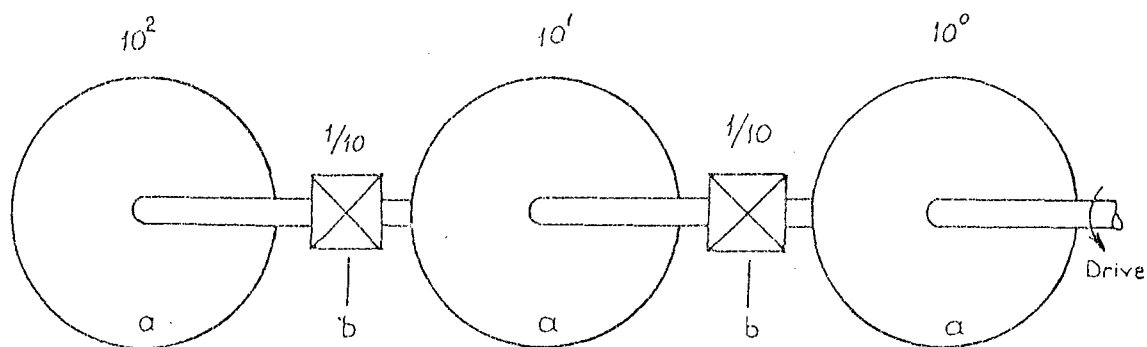


Figure 3.16. Zonal scanning of binary coded discs;
 a: Code discs as in figure 3.15a,
 b: Gearing with 1:10 ratio.

In the absolute linear devices, to determine the position of the machine slide directly, coded linear scales are desirable. However the cost of producing these scales is very high due to the following reasons:

- a) displacement elements "AS" are very small,
- b) large number of tracks are required,
- c) precautions are needed to ensure accurate scanning, etc.

In the designed system, because of accuracy the linear incremental digital position transducers are used.

3.3. Interfacing of the system with the Microcomputer.

As previously explained, the microcomputer will generate two signals (one direction and one pulse signal) for each motor sequencer. It should also take the positional feedback information from position transducers. There are three positional information, they are taken from x,y, and z axes. Microcomputer should be informed about whether the table is going in the positive or in the negative direction in any axis. It takes totally 2 bits feedback signal.

The position transducers which convert the positional information into voltage form using only the logic levels 0 and

1(0V and 5V). It is an incremental transducer. It also gives the direction of the movement. The physical position values are determined by software counts of taken incremental values.

In general, the winding of stepping motors may be adjusted to different impedance levels to trade voltage for current. Rising rate of current in a winding is proportional to V/I_m , so that higher voltages will yield better high-speed performance generally. It is not unusual for the source voltage for pulse initiation and termination to be 10, 20 or 30 times the steady state motor voltage. There are some methods for control of this 'overdrive' voltage so as to limit steady state current to correct value[0Z].

As mentioned earlier, the control system is designed for linear motion in three dimensions. The system programs are written for this purpose. Motion along any counter can be traced by suitably dividing it into linear sections and a master program can be written to do this. Note that such a master program will require additional memory space and care must be taken so that the additional calculations can be done in the available interval.

The linear motion between the given two points can be traced in different manners. The point to point control method is selected. This is one of the reliable and simple methods.

In this method, at the beginning point of the motion, one step is taken in the x direction, then the calculation is done to find the corresponding y coordinate and this y coordinate is reached by taken steps in y direction. When this y coordinate for the physical (actual) x coordinate is obtained, one more step is taken in x direction. This continues until the destination point is reached. As it is noticed, only one directional motion is done at a time.

Movement in x direction is always one step. This corres-

ponds to the accuracy tolerance as explained earlier. If the calculated y coordinate for the physical x coordinate is not a multiple of 0.05 mm(one step), then the largest of the multiples of 0.05 mm which is smaller than this y value is taken.

CHAPTER IV

SOFTWARE OF THE SYSTEM

4.1. Software Specifications

Logic of the system software is as follows:

The coordinates and feedrate are entered from keyboard by the operator in the decimal form which are converted to the hexadecimal form and so, all the operations and calculations are done in this form. The entered data, coordinate or feedrate, which is not permissible for the system is not accepted by the software. The coordinates of the line between the given two points is determined, its parameters are calculated. The delay routines are adjusted by the entered feedrate to make it possible to achieve the required pulse frequency which is given to the control circuits of the motors. This pulse frequency determines the speed of the motors and, indirectly, the feedrate. Therefore, this pulse frequency is obtained by adjusting the delay duration through which microcomputer passes in the loop. The microcomputer takes the physical (actual) positional data from position transducers and compare them with the ideal ones which are calculated from the line equation and determines the control signals for the rotations of the motors [DA].

4.2. Basic Functions

The four basic arithmetic operations, addition, subtraction, multiplication and division are used in the system software.

Since they are used many times in the system program, they are developed and organized as subroutines. Three subroutines are written for these four basic arithmetic operations. These are:

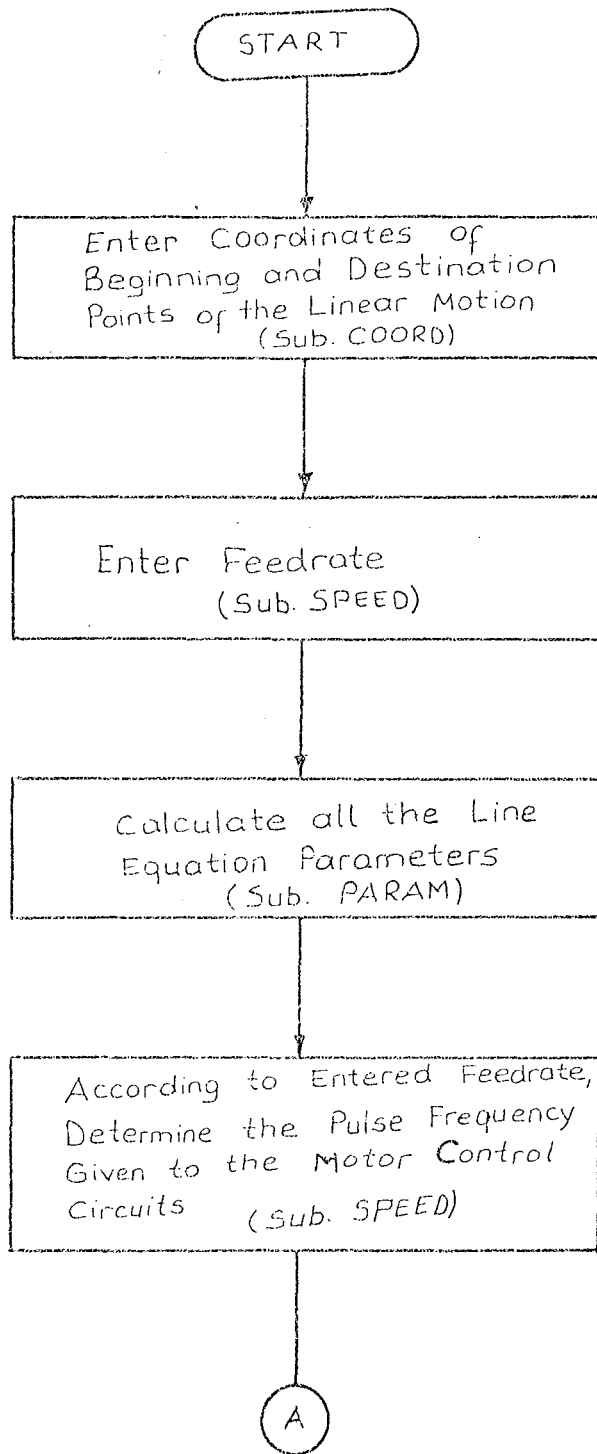
ADDSUB: This subroutine performs the addition of two 3 byte length signed binary numbers to give a result of 3 byte length signed binary number. If the second number has a minus sign then the operation will be the subtraction.

MULT: This subroutine multiplies the two 4 byte length signed binary numbers. As a result it gives a 4 byte length signed binary number.

DIVI : This subroutine divides a 2 byte length binary number to another 2 byte length binary number. *The result of the division* consists of 4 bytes. Right most 2 bytes give the fractional part of the result[F].

4.3. Main Program

Since the programs related to some system functions are written as subroutines, the main program generally is constituted of calling some subroutines. As it is seen from Figure 4.1, at the beginning, the main program calls COORD subroutine for accepting the beginning and destination point coordinates. Calling the subroutine SPEED makes the entrance of feedrate possible. The subroutine PARAM calculates the line equation parameters. Ideal Y- coordinate is calculated by calling the subroutine YID using the line equation parameters. The output from microcomputer to the control circuits of the motors is done by calling the subroutine OUT. Transferring the positional information from position transducers to microcomputer is done by the subroutine INPUT. The determination of whether the entered destination point is reached or not is done by the subroutine OKEY.



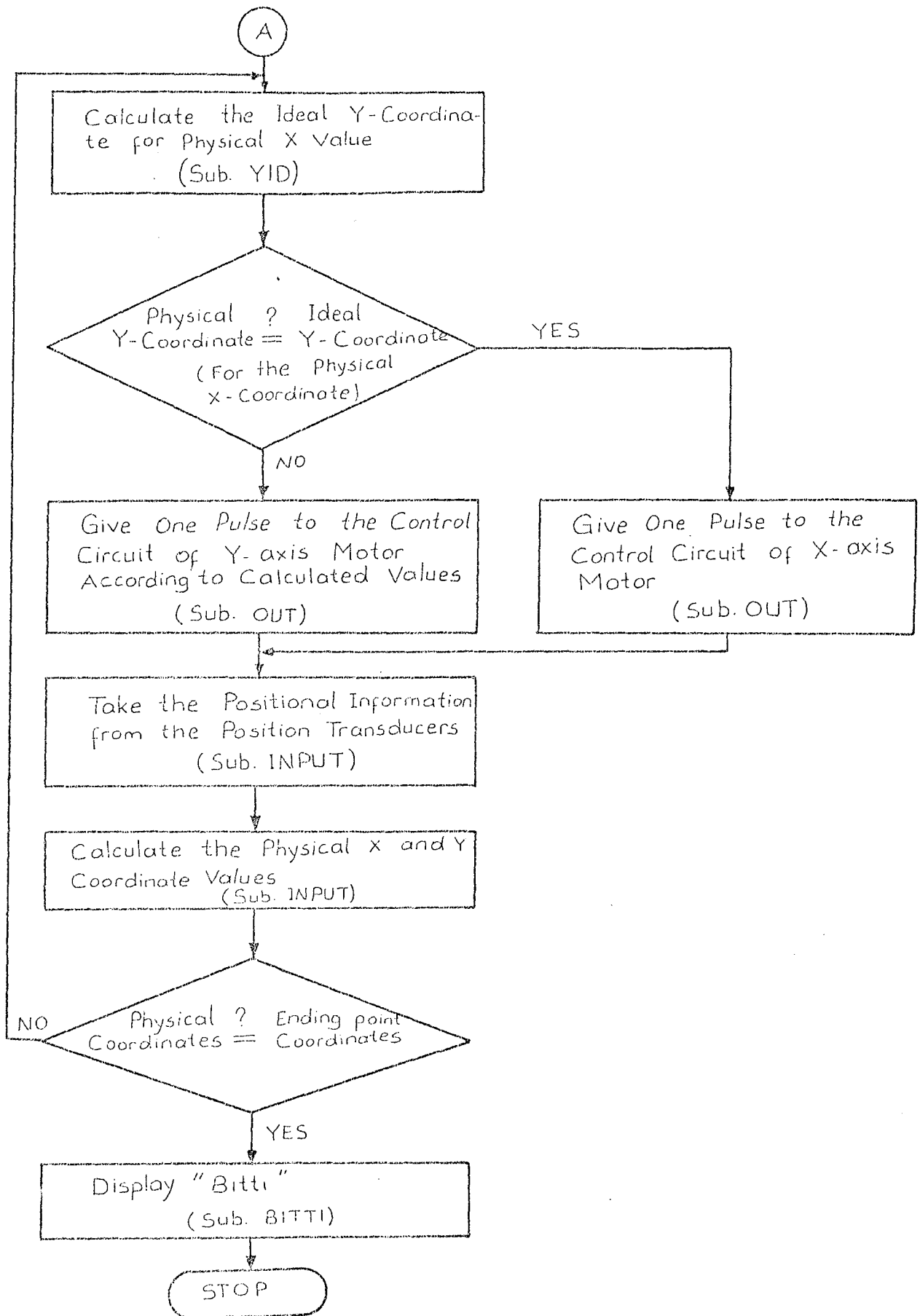


Figure 4.1. Flowchart of main program.

If the destination point is reached then the subroutine BITTI indicates this by displaying "Bitti" on seven segment displays. Otherwise, the main program calls the subroutine YID once more and goes on as it is seen from the flowchart in Figure 4.1.

In the following sections, brief flowcharts and explanations of these subroutines are given. The main program and these subroutines written in Assembly Language are given in Appendix B.

4.4. Subroutines

In this section the subroutines called by the main program will be considered one by one. Their logic and flowcharts will be given in the simplified form [DA].

These subroutines are: COORD, SPEED, PARAM, YID, OUT, INPUT, OKEY, BITTI.

4.4.1. COORD Subroutine

This subroutine makes possible the entrance of the beginning and destination point coordinates of the linear motion, $(X1, Y1)$ and $(X2, Y2)$, respectively, from the keyboard. First "HI=" is displayed on the seven segment displays which is corresponded about Z axis. Z-axis is independent of X and Y axes. The data is entered to microcomputer in decimal form. After entrance of data in decimal form, microcomputer converts this into hexadecimal form. The flowchart about this subroutine is shown in Figure 4.2.

4.4.2 SPEED Subroutine

Like the subroutine COORD, this subroutine makes the entrance of feedrate from the keyboard possible. As explained before, it does not accept the feedrate values which are out of the range of the system. If the entered feedrate value is in the permissible range, i.e. valid data, then this entered decimal data value is

converted into the hexadecimal form as shown in Figure 4.3. This subroutine also adjusts the variables of delay routines. These routines are adjusted according to the entered feedrate value. They adjust time delays between the pulses given to the motor control circuits to get the desired speed of the motor.

4.4.3 PARAM Subroutine

This subroutine calculates and finds the slope and constant term of the line equation on which the ideal motion takes place. These two calculated values are used in the subroutine YID. Its flowchart is given in Figure 4.4.

4.4.4. YID Subroutine

Since the system is a closed loop control system, a feedback is taken from the position of the table and according to this feedback new decisions are done. In system, for the physical (actual) x coordinate, the desired y coordinate should be known. Subroutine YID calculates ideal y coordinate (YID) for any x coordinate value. It uses the slope (M) and constant term ($YI-M \times XI$) of the line, on which the ideal motion takes place, which is calculated in subroutine PARAM. The flowchart of YID is given in Figure 4.5.

4.4.5. OUT Subroutine

At the control of the motion, after the comparison between the physical and ideal coordinates, some pulses should be produced to the motor control circuits. The main aim of this subroutine is to produce these pulses. Each pulse given to the pulse input of the sequencer makes the motor take one

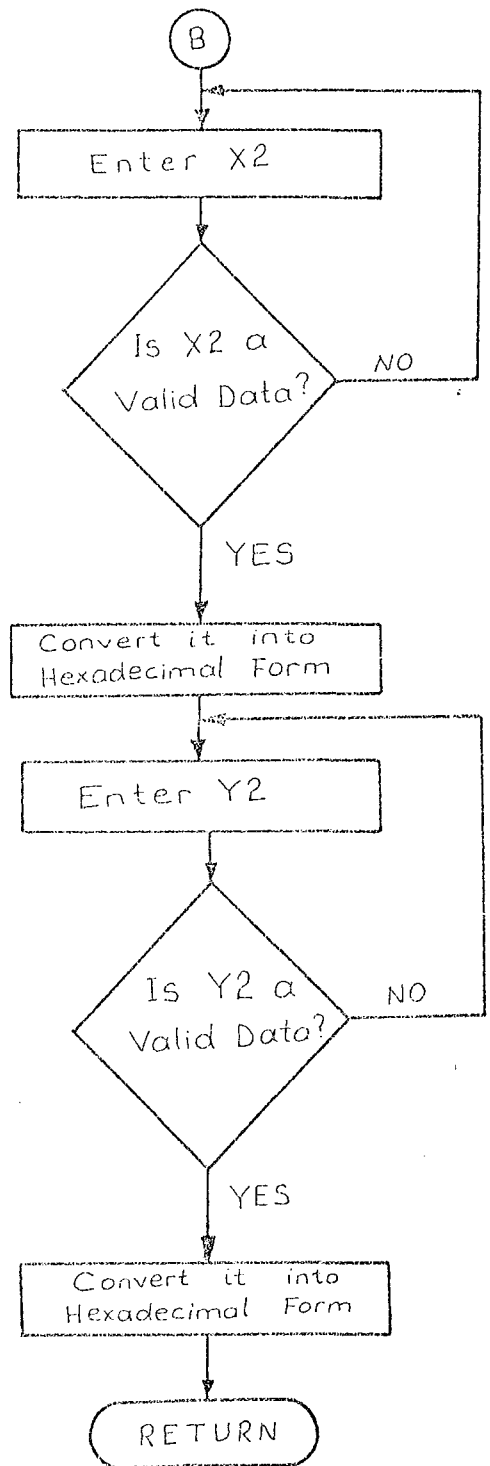
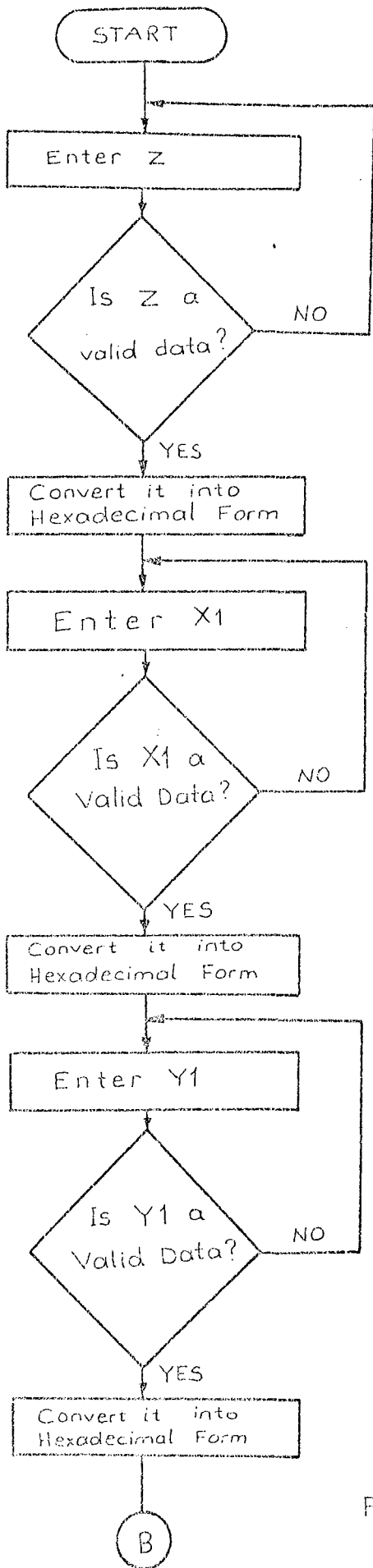


Figure 4.2. Flowchart of COORD subroutine.

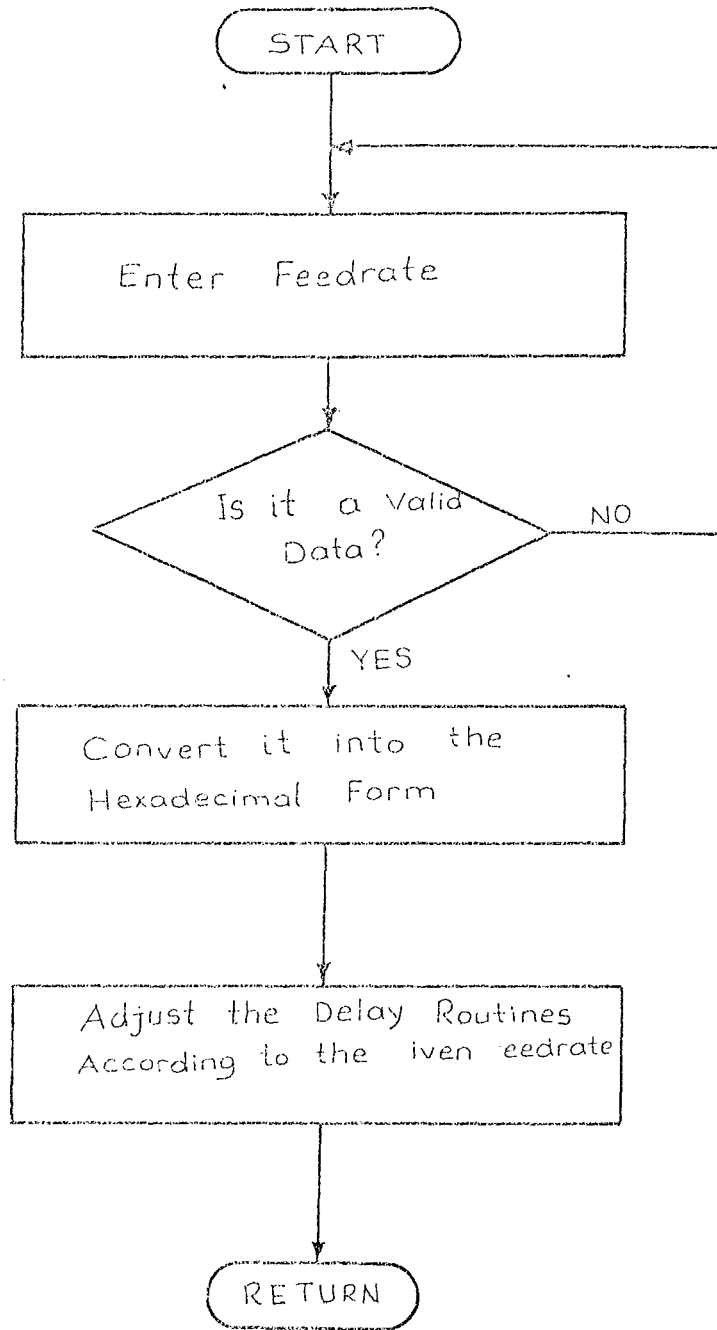


Figure 4.4. Flowchart of SPEED subroutine.

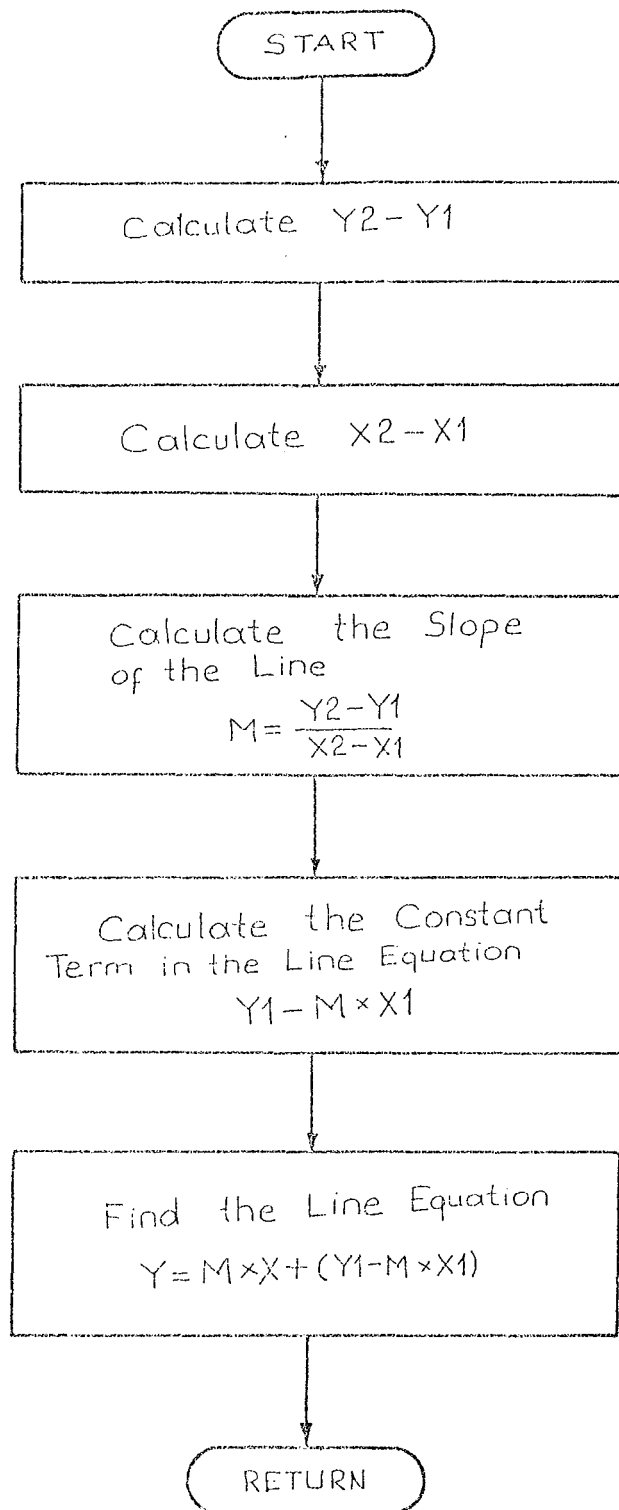


Figure 4.3. Flowchart of PARAM subroutine.

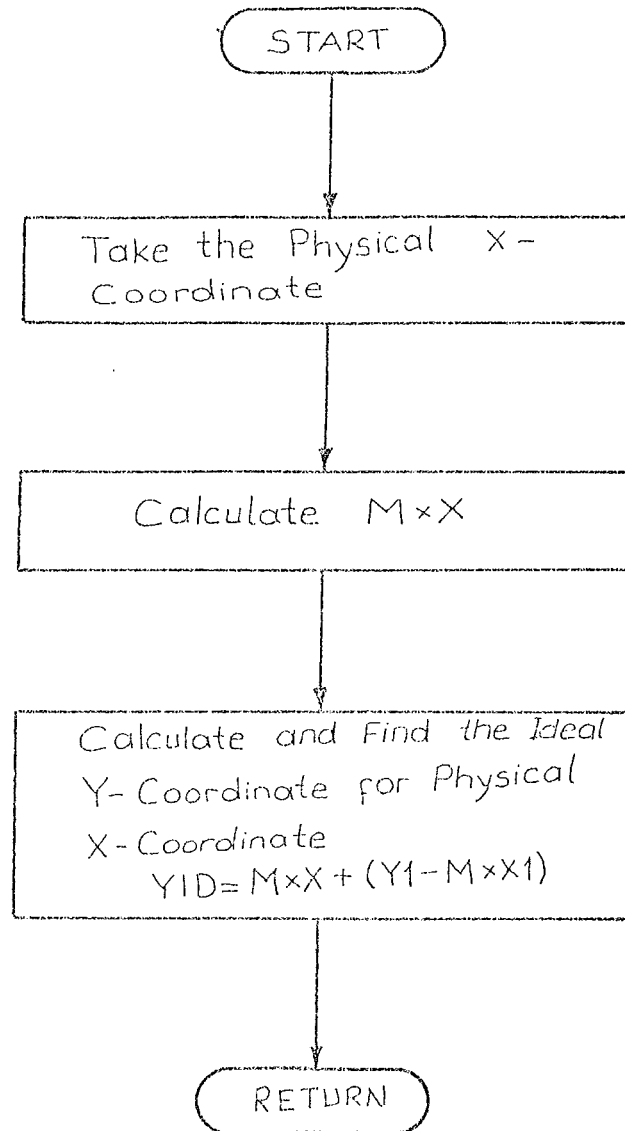


Figure 4.5. Flowchart of YID subroutine.

step in the determined direction by the sequencer direction input. For a given x coordinate, this subroutine compares physical (YAC) and ideal (YID) y coordinates. As a result of this comparison, it generates the required pulses to the motor control circuit to make the physical y coordinate equals to the ideal y coordinate. If they are equal, then it brings the x coordinate to its next value. The flowchart of this subroutine is shown in Figure 4.5.

4.4.6 INPUT Subroutine

In the closed loop control system, a feedback always exists. The purpose of this subroutine is to produce this feedback so that the ideal position of the table could be continuously checked with the current position defined by subroutine OUT. It takes positional data input from position transducers and determines which data is from which transducer. By the use of this obtained incremental positional data, absolute physical x-coordinate (XAC) and y coordinate (YAC) are obtained as shown in Figure 4.7.

4.4.7 OKEY Subroutine

This subroutine determines whether the destination point of the motion is reached or not. It sets the variable OKEY to 1 if it is reached, to 0 if not reached. This variable is checked in the main program to see whether the destination point is reached or not. Its flowchart is given Figure 4.8.

4.4.8. BITTI Subroutine

This subroutine displays "Bitti" on the seven segment displays to show that destination point of the motion is reached. Its flowchart is shown in Figure 4.9.

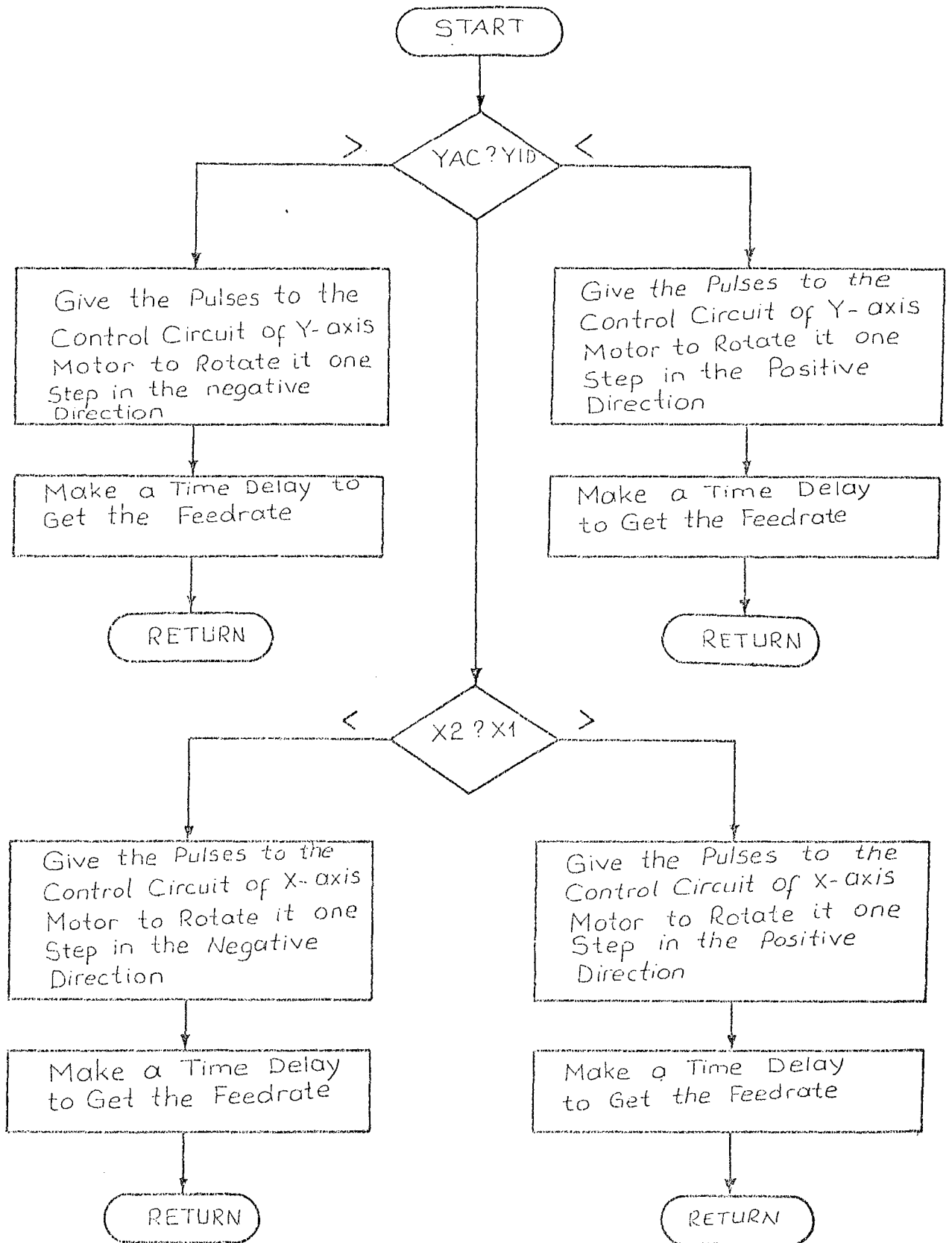


Figure 4.6 Flowchart of OUT subroutine.

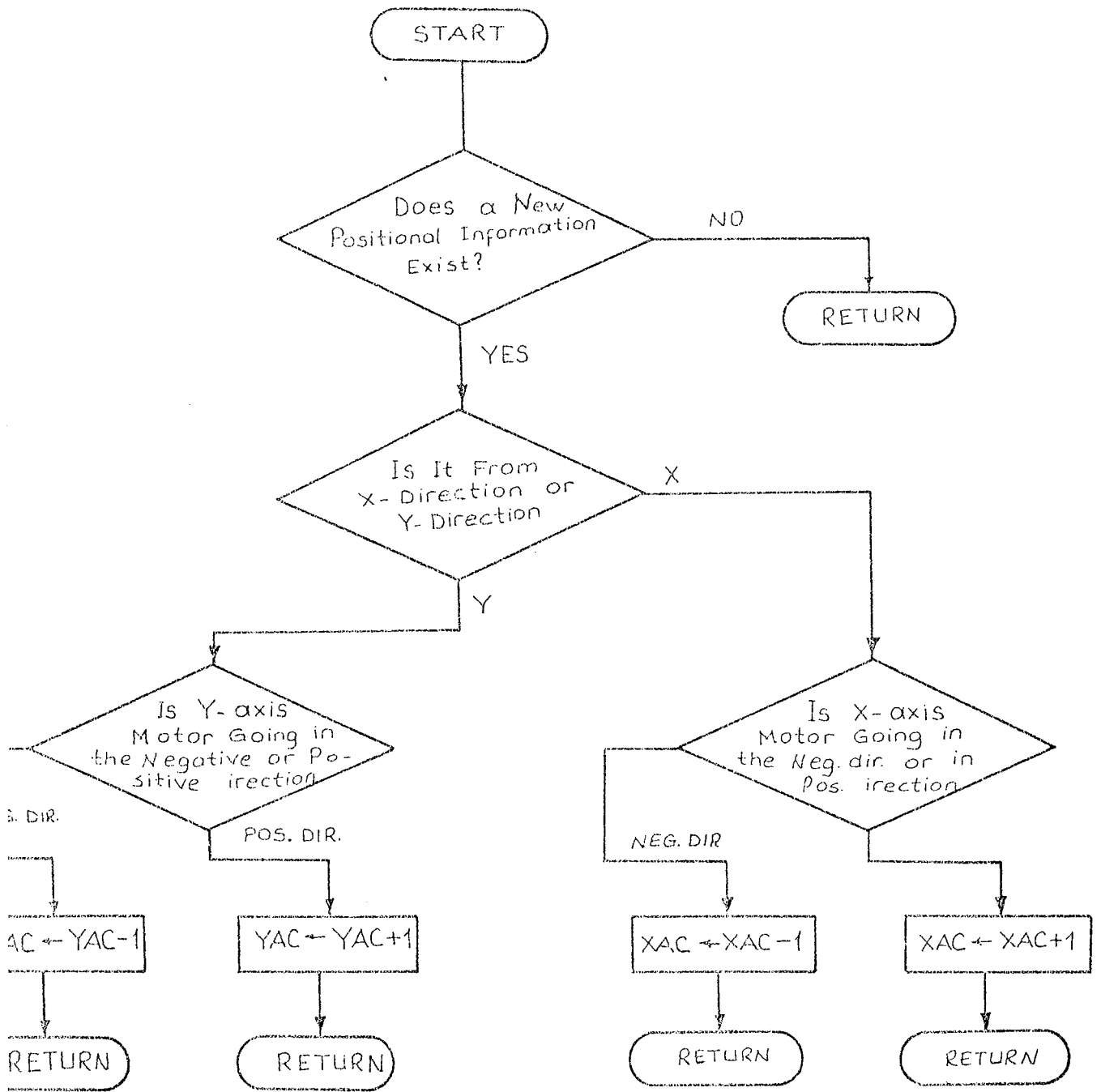


Figure 4.7. Flowchart of INPUT subroutine.

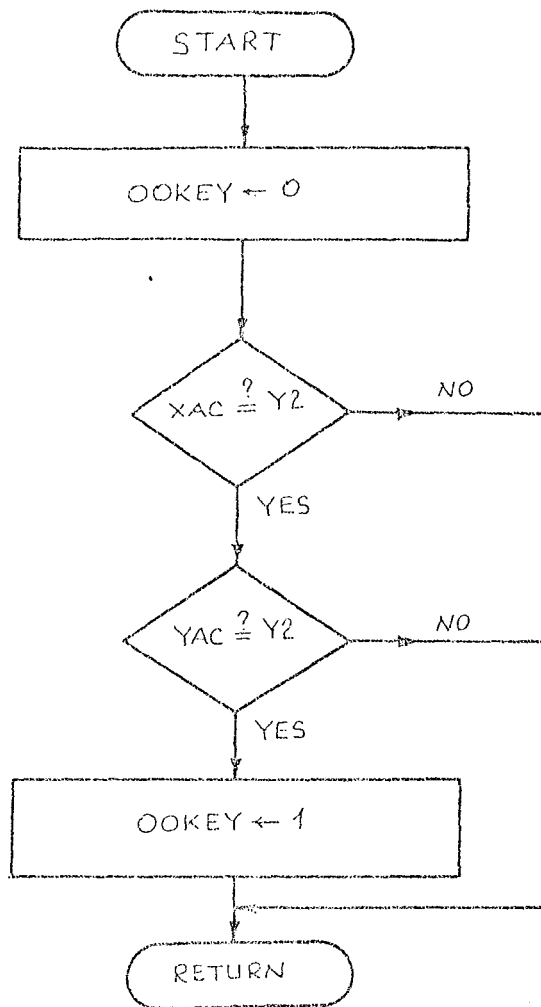


Figure 4.8 Flowchart of OKEY subroutine.

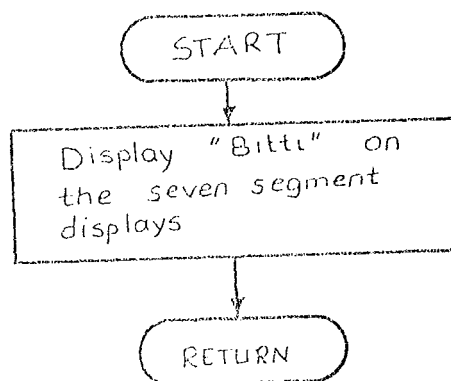


Figure 4.9. Flowchart of BITTI subroutine.

CHAPTER V

RESULTS, CONCLUSIONS AND RECOMMENDATIONS

5.1. Testing of Motor Operations

Sequencer and drive circuits are connected to the stepping motor to test the motor operation. A square wave generator is then used to get different rotational speeds. This is achieved by changing the frequency of the applied signal. It is observed that any rotational speed between 0 and 500 steps/sec can be reached. After this, starting and stopping performances of the motor at these speed values are also observed as shown in Figure 5.1a. In order to find out whether the motor loses any step or not, the sequencer is excited by different number of pulses at a predetermined frequency and corresponding rotation of the motor is observed. From these tests, it is seen that for the speed values up to 400 steps/sec the motor does not lose any step as shown in Figure 5.1b.

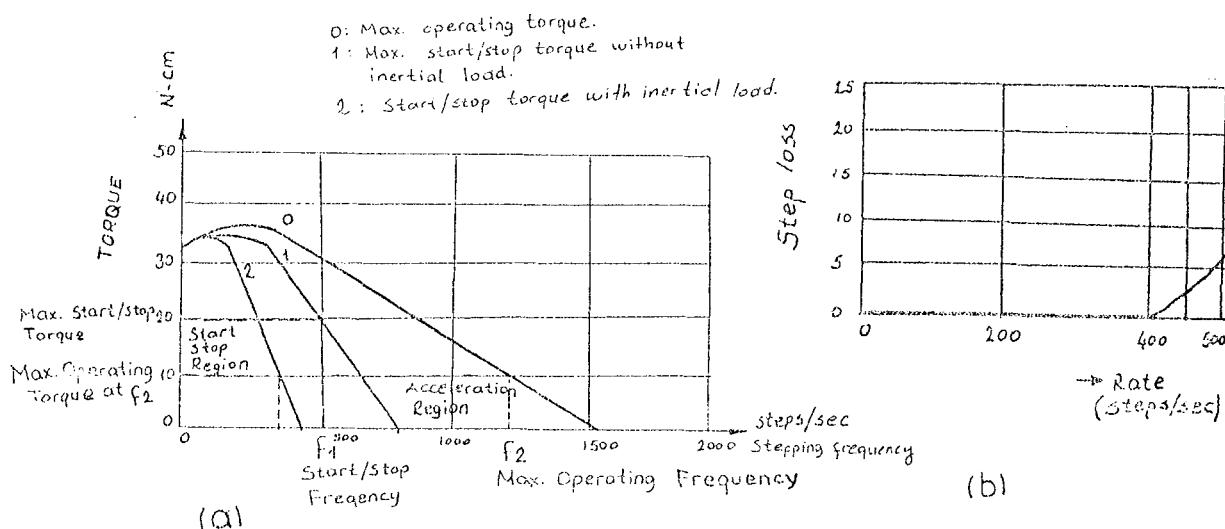


Figure 5.1 Starting and Stopping Performance of Motor

However, for higher rotational speed values the motor may lose some steps while starting or stopping it. It should also be pointed out that these tests are carried out with an insufficient voltage and current values so that the drive circuit is operated under its nominal current and voltage ratings. Its reason was the nonexistence of a well regulated dc power supply.

It is expected that the higher start-stop speed rates can be attained without losing any step if the drive circuit could be fed at its nominal ratings. On the other hand, it would be better to use higher voltage and current rating switching transistors to improve the performance, or the VMOS power field-effect transistors.

During the operation of the motor, its winding voltage and current waveforms are observed as shown in Figure 5.2.

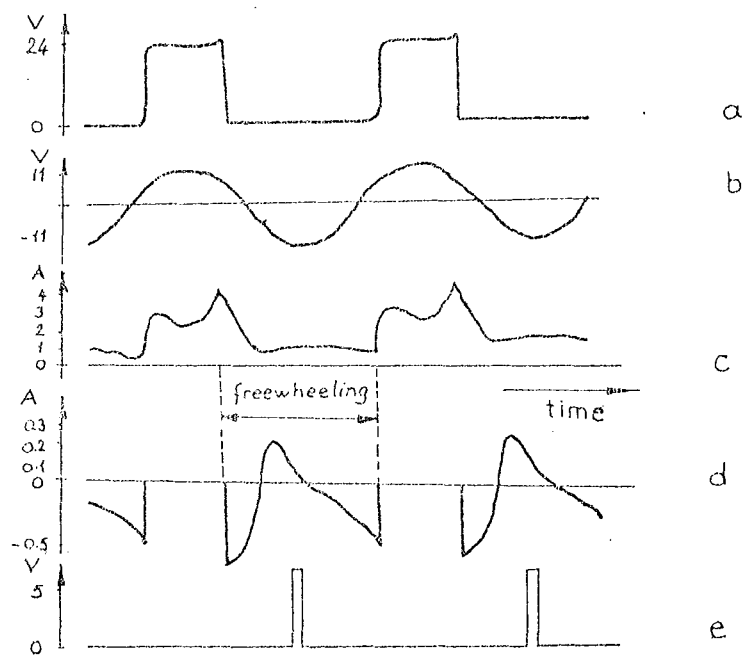


Figure 5.2. Waveforms of Phase-A Winding;

- a) applied phase voltage,
- b) motional voltage,
- c) current waveform,
- d) rate of change of freewheeling current
- e) position detector pulse.

At the beginning of each pulse, there is a damped oscillation in the winding voltage. This shows that, at these points there is a mechanical damped oscillation in settling of the rotors, which is a known problem of stepping motors [AC], [FI]. These tests are done on no load conditions. When it is operated on load conditions, these oscillations would be more dominant. In future work, to reach higher speed values this fact should be taken into account. This will require a more sophisticated drive circuit.

5.2. Testing of the Linear Motion

When the beginning and destination point coordinates and feedrate are entered to the microcomputer, it generates control pulses to the sequencers for x,y,z directions. In the first stage of the test procedure, the motor is connected to the system to control the motion in x direction and x component of the linear motion is observed. For the second stage, the above procedure is repeated for y direction and y component of the linear motion is obtained.

5.3. Conclusions and Recommendations

In this thesis work, microcomputer control of linear motion of the table of a machine tool is designed, constructed and tested with the available facilities.

In tests, satisfactory operation in open loop on no load conditions is obtained up to 1300 mm/min feedrate in 0.05 mm. steps. In closed loop operation, due to damped oscillation of the motor's rotor the desired operation is not obtained.

From the above discussions it is clear that by carrying out some modifications, the closed loop performance of the system can be improved.

This work shows that such a control system can be designed and implemented to satisfy the design specifications. However, improvements in the drive circuit is essential. The software of the system, on the other hand is quite flexible therefore is suitable to be used as a building block in a more sophisticated control environments.

A master program which converts any given path into linear segments can also be written. Once the path is subdivided into linear segments, the software developed here can be used to follow these linear contour segments and thereby permitting complex contours to be followed. ALP 85 microcomputer EPROM memory region can be expanded with ease to store this master program.

The followings are recommended for future work on this topic:

- Oscillations of the rotor should be either electronically or mechanically damped for a reliable operation.
- A master program which converts any given motion path into linear segments should be written.
- EPROM memory region of the ALP 85 microcomputer should be expanded to store master program.
- Switching transistors can be replaced with VMOS field-effect transistor to improve the reliability of drive circuit.
- This work should be completed by fitting the system developed here to a proper machine tool. Some electrical and mechanical problems probably will arise when the overall system is operated with closed loop control strategy. These problems may be solved within the scope of another research work.

Operating instructions of the system are listed in Appendix C. A photograph of the system set is given in Appendix D.

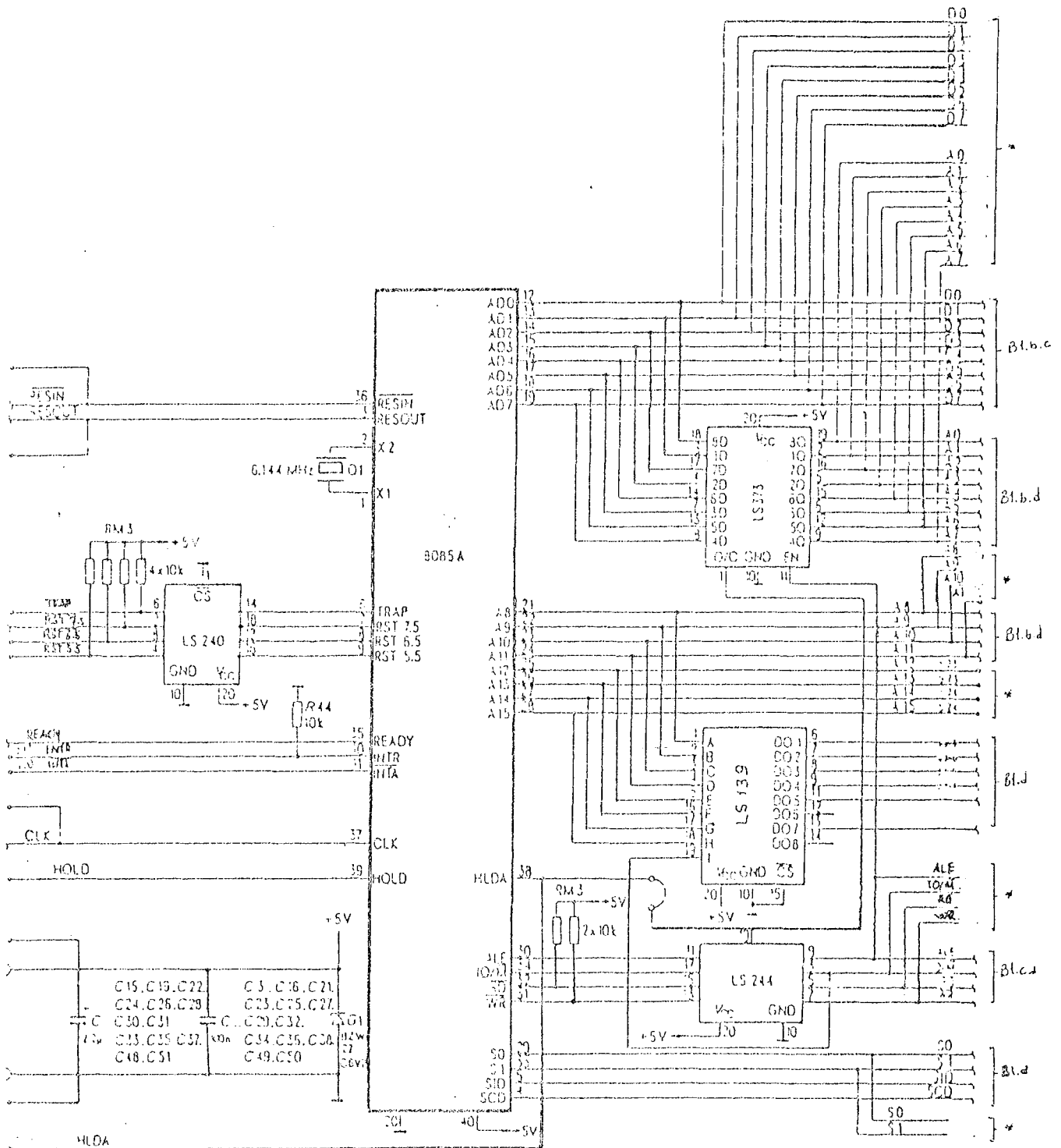
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APPENDIX A

MICROCOMPUTER CIRCUITS



* connections of the user interface.

Figure A-1. Circuit diagram of ALP 85 microcomputer

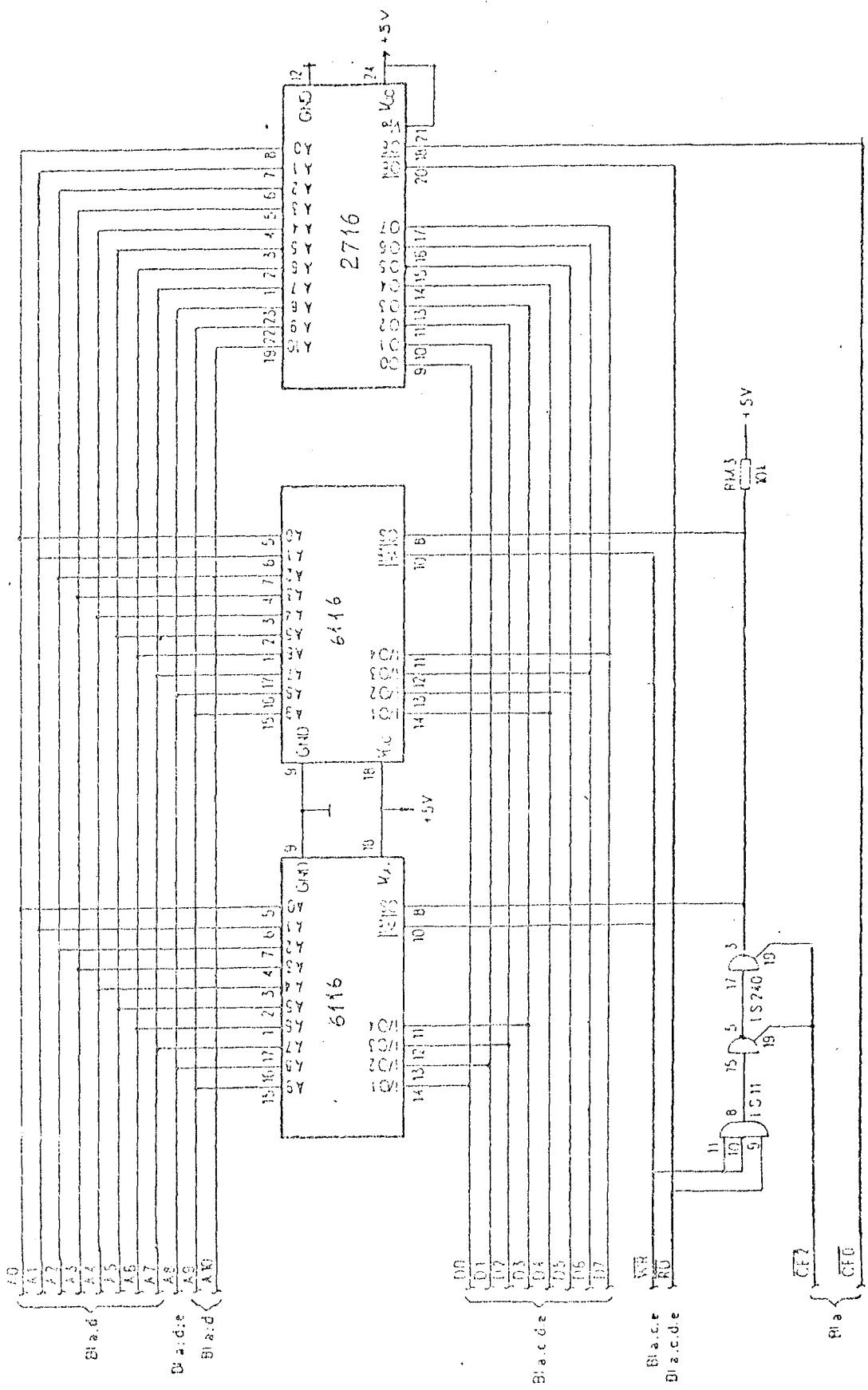


Figure A-1. Continued.

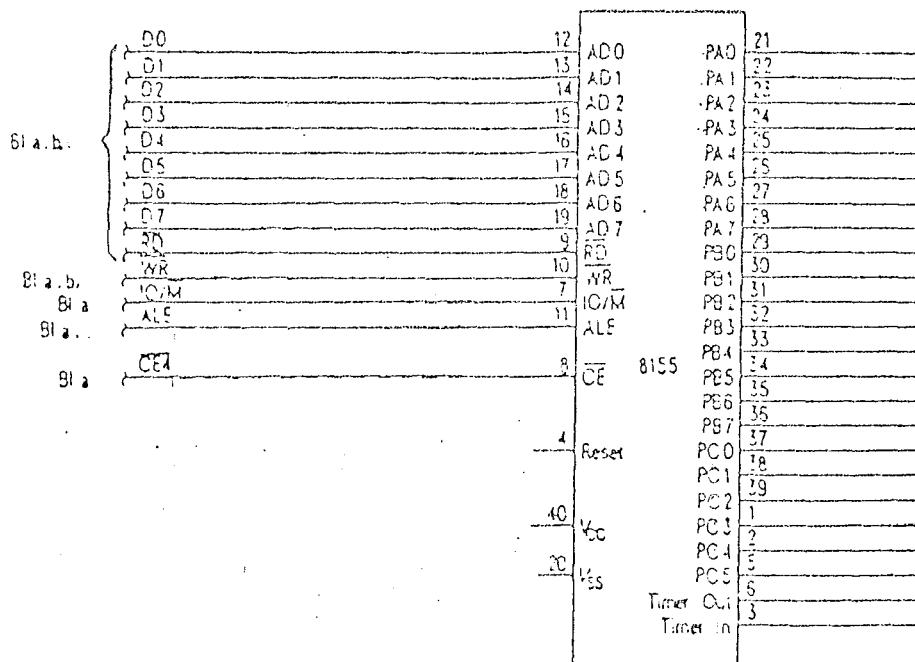
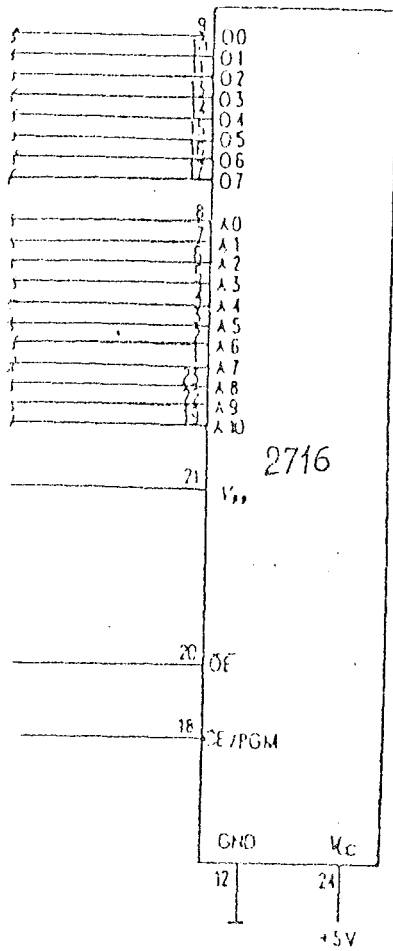


Figure A-1. Continued.

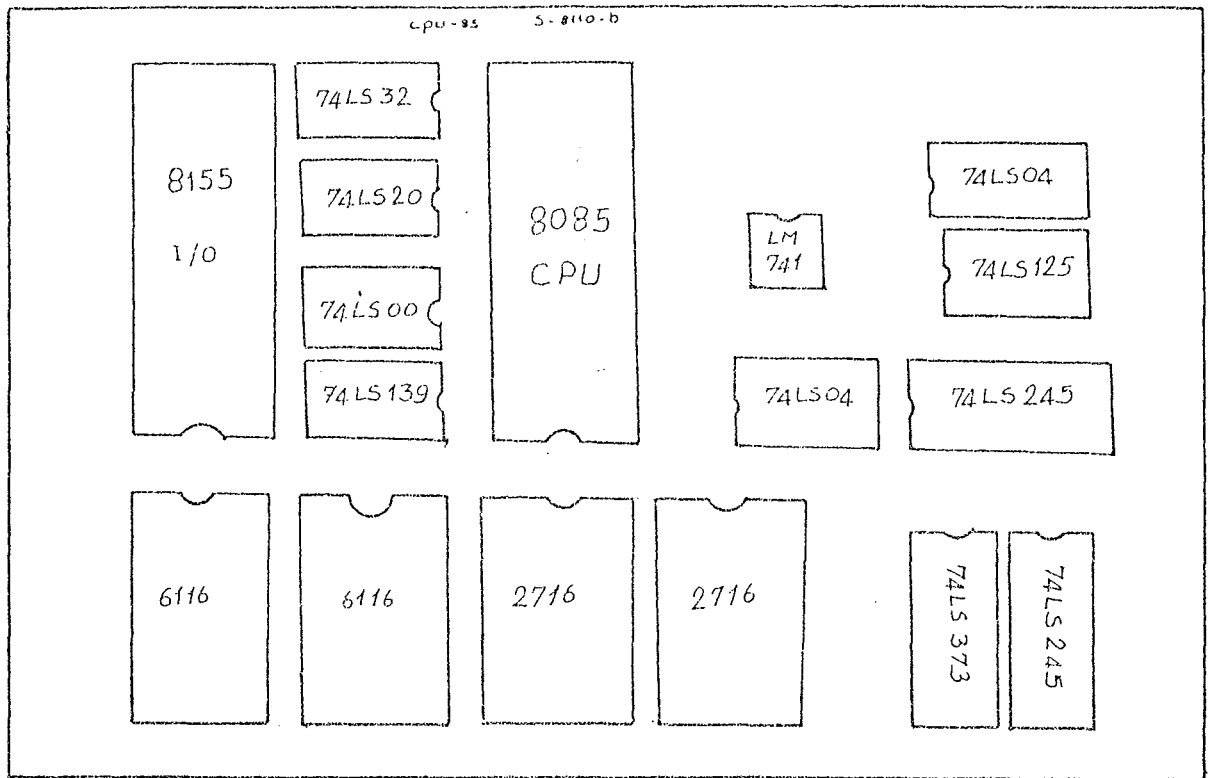


Figure A-2. Placement of Components on B- Side.

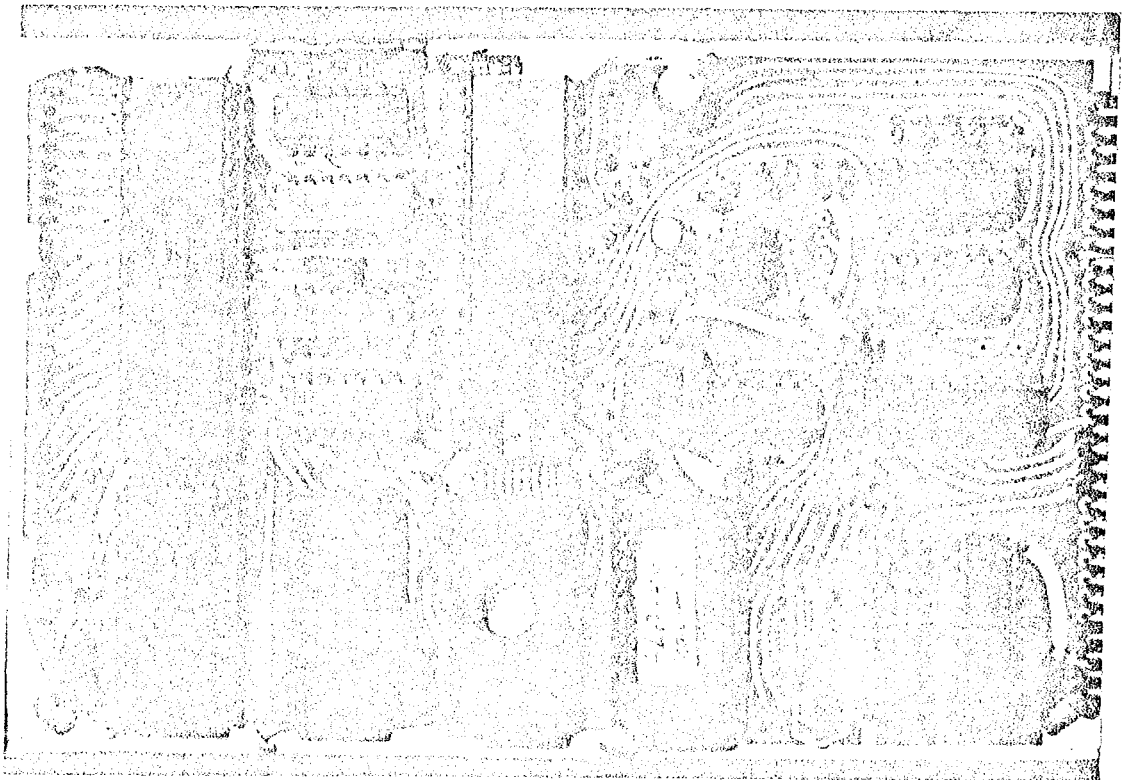
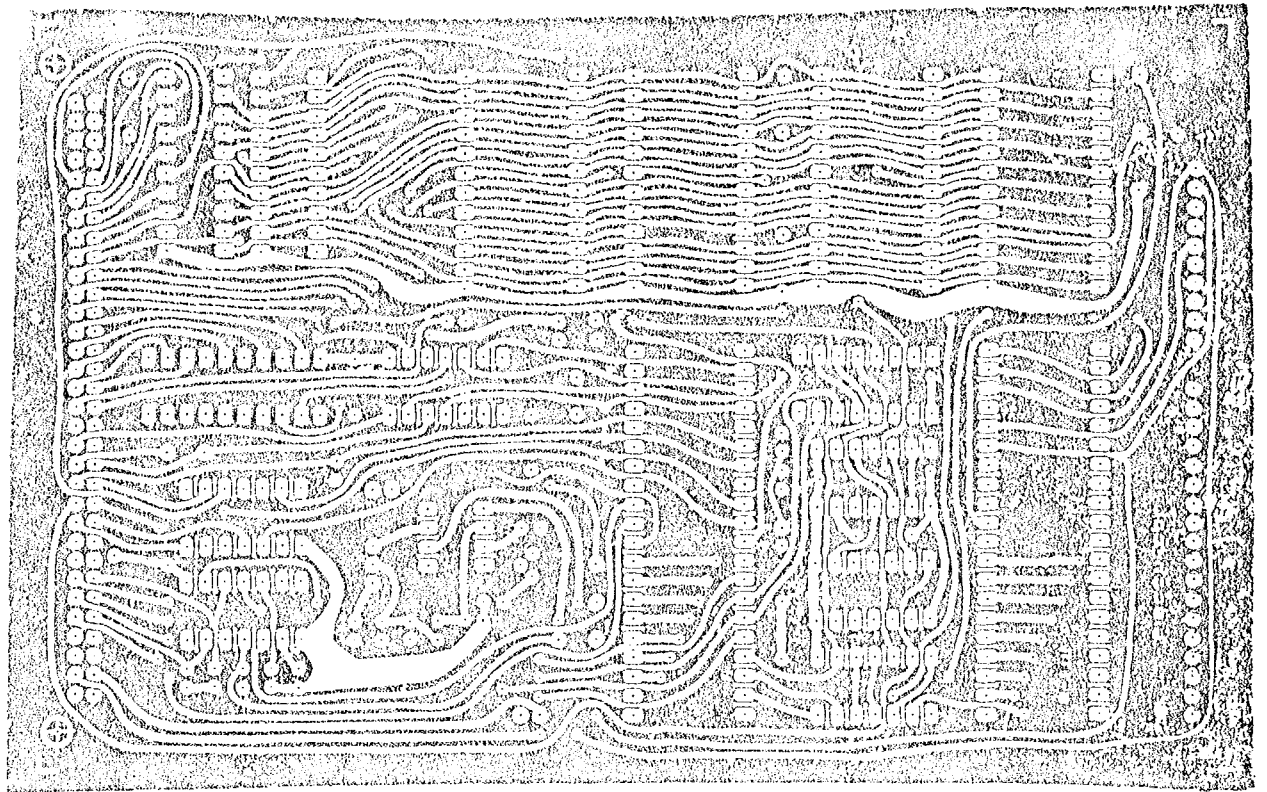
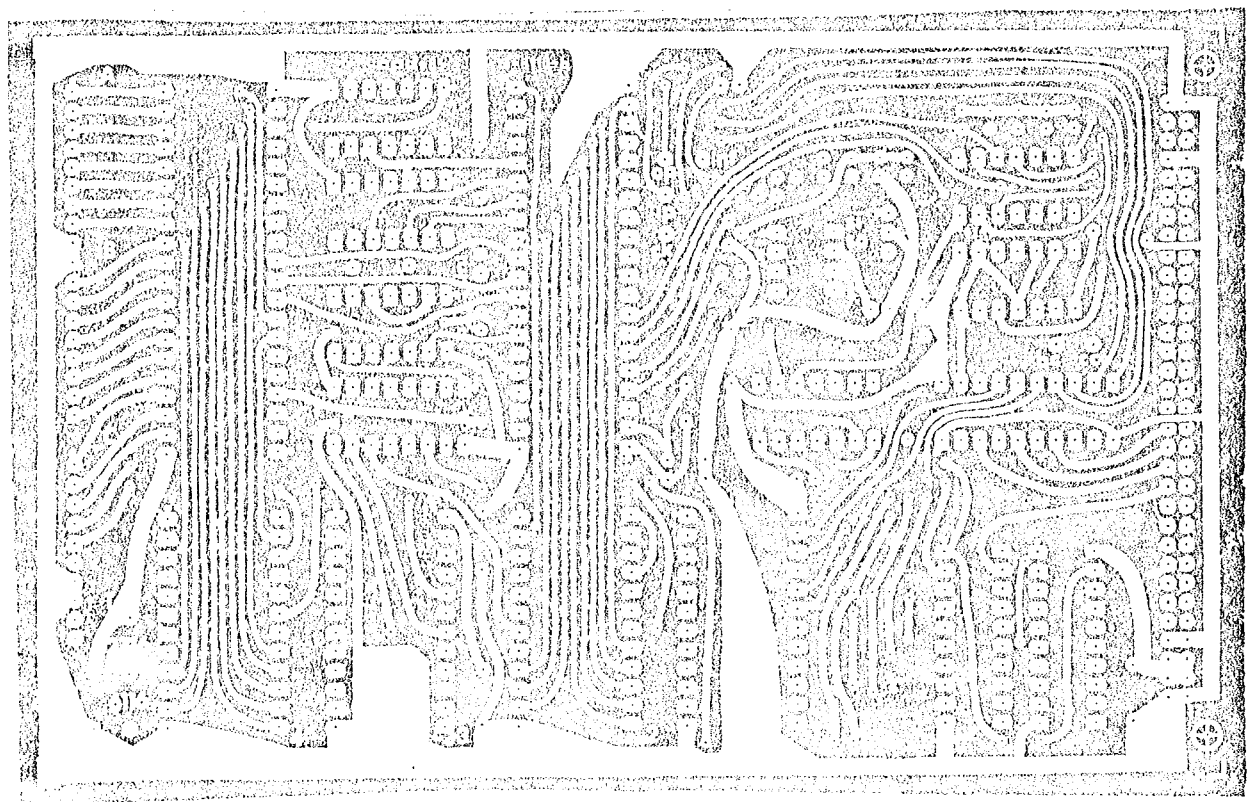


Figure A-3. Photograph of CPU Card.



(a)



(b)

Figure A-4. Printed Circuit Layout of ALP 85.

a: A-Side

b: B-Side

APPENDIX B : SYSTEM PROGRAM

8060/8085 MACRO ASSEMBLER

MODULE PAGE 1

LINE	SOURCE STATEMENT	LINE	SOURCE STATEMENT
1 ;	MAIN MACHINE TOOL CONTROL	51	PUSH D
2 ;	PROGRAM IN THREE DIMENSIONS	52	PUSH H
3 ;		53	LHLD 1B0FH
4	ORG 0800H	54	LDA 1B0EH
5 AUS	EQU 05A8H	55	MOV B,A
6 ANWEIN	EQU 0701H	56	LDA 1B11H
7 ANVAZL	EQU 07BCH	57	DAD D
8 UMY	EQU 05C9H	58	SHLD 1B12H
9 ANWAUS	EQU 07C7H	59	ADC B
10 MAIN :	LXI SP, E0B0H	60	STA 1B14H
11	MVI A, F1H	61	ANI 80H
12	OUT E1H	62	JZ LEAVE
13	CALL COORD	63	LXI H, 1B14H
14	CALL SPEED	64	CALL COMP
15	CALL BASLA	65 LEAVE :	POP H
16	MOV A, E	66	POP D
17	CPI OFH	67	POP B
18	JZ SPEED1	68 COMP :	DCX H
19	JMP ANA	69	DCX H
20 CMAIN :	CALL PARAM	70	MOV A, M
21	IN FAH	71	CMA
22	ANI OFH	72	ADI 01H
23	STA 1B36H	73	MOV M, A
24	LHLD 1B00H	74	INX H
25	SHLD 1B34H	75	MOV A, K
26	LHLD 1B03H	76	CMA
27	SHLD 1B32H	77	ACI 00H
28 LOOP :	CALL YID	78	MOV M, A
29	CALL OKEY	79	INX H
30	JZ SOBE	80	MOV A, M
31	CALL OUT	81	CMA
32	CALL INPUT	82	ACI 00H
33	JMP LOOP	83	ORI 80H
34 SOBE :	CALL BITTI	84	MOV M, A
35	HLT	85	RET
36 ;		86 CHECK :	LXI H, 1B0CH
37 ; END OF MAIN PROGRAM		87	CALL SS
38 ;		88	LXI H, 1B0FH
39 ; ADDSUB SUBROUTINE		89	CALL SS
40 ;		90	RET
41 ADDSUB :	CALL CHECK	91 SS :	MOV A, M
42	LXI H, 1B0EH	92	ANA A
43	MOV A, M	93	RNZ
44	ANI 80H	94	INX H
45	INZ COMP	95	MOV A, M
46	LXI H, 1B11H	96	ANA A
47	MOV A, M	97	RNZ
48	ANI 80H	98	INX H
49	JRZ COMP	99	SUB A
50	PUSH B	100	MOV H, A

LINE	SOURCE STATEMENT	LINE	SOURCE STATEMENT
101	RET	151	SHIFT 2 : MOV A,M
102	:	152	RAL
103	: END OF ADDSUB	153	NOP
104	:	154	MOV M,A
105	: MULT SUBROUTINE	155	INX H
106	:	156	MOV A,M
107	MULT : LDA 1B43H	157	RAL
108	ANI 80H	158	MOV M,A
109	MOV C,A	159	RET
110	LDA 1B47H	160	SUM : PUSH B
111	ANI 80H	161	LHLD 1B40H
112	XRA C	162	XCHG
113	MOV C,A	163	LHLD 1B48H
114	LDA 1B43H	164	DAD D
115	ANI 7FH	165	SHLD 1B48H
116	STA 1B43H	166	JC CAR
117	LDA 1B47H	167	CALL REPEAT
118	ANI 7FH	168	SHLD 1B4AH
119	STA 1B47H	169	POP B
120	LXI H,1B46H	170	RET
121	SUB A	171	CAR : CALL REPEAT
122	MOV M,A	172	INX H
123	INX H	173	SHLD 1B4AH
124	MOV M,A	174	POP B
125	INX H	175	RET
126	MOV M,A	176	REPEAT : LHLD 1B42H
127	INX H	177	XCHG
128	MOV M,A	178	LHLD 1B4AH
129	MVI B,20H	179	DAD D
130	STAR: LXI H,1B48H	180	RET
131	CALL SHIFT	181	:
132	LXI H,1B44H	182	:END OF MULT
133	CALL SHIFT	183	:
134	JNC CONT	184	:DIVI SUBROUTINE
135	CALL SUM	185	:
136	CONT : DCR B	186	DIVI : LHLD 1B18H
137	JNZ STAR	187	XCHG
138	LDA 1B4BH	188	LHLD 1B1CH
139	ORA C	189	MOV C,L
140	STA 1B4BH	190	MOV B,H
141	RET	191	LXI H,0000H
142	SHIFT : MOV A,M	192	CALL DIV
143	ANA A	193	XCHG
144	RAL	194	SHLD 1B2AH
145	MOV M,A	195	LHLD 1B1CH
146	INX H	196	MOV C,L
147	SHIFT1 : MOV A,M	197	MOV B,H
148	RAL	198	LXI H,0000H
149	MOV M,A	199	XCHG
150	INX H	200	CALL DIV

LINE	SOURCE STATEMENT	LINE	SOURCE STATEMENT
201	XCHG	251	COORD: CALL SIFIR
202	SHLD 1B28H	252	CALL Z
203	RET	253	CALL X
204	DIV: MOV A,B	254	CALL ONEE
205	CMA	255	CALL XX122
206	MOV B,A	256	LXI H,0000H
207	MOV A,C	257	CALL CEVIR
208	CMA	258	SHLD 1B00H
209	MOV C,A	259	CALL BASLA
210	INX B	260	MOV A,E
211	CALL LOOP1	261	CPI 0FH
212	LOOP1: MOV A,D	262	JZ COORD
213	MOV D,E	263	SHOWY1: CALL Y
214	MVI E,08H	264	CALL ONLE1
215	LOOP11: DAD H	265	CALL YY12
216	JC OVER	266	CALL CEVIR1
217	ADD A	267	SHLD 1B03H
218	JNC SUB1	268	CALL BASLA
219	INX H	269	MOV A,E
220	SUB1: PUSH H	270	CPI 0FH
221	DAD B	271	JZ SHOWY1
222	JC OK	272	SHOWX1: CALL X
223	POP H	273	CALL TWDE
224	DCR E	274	CALL XX12
225	JNZ LOOP11	275	LXI H,0000H
226	MOV E,A	276	CALL CEVIR2
227	STC	277	SHLD 1B06H
228	RET	278	CALL BASLA
229	OK: INX SP	279	MOV A,E
230	INX SP	280	CPI 0FH
231	INR A	281	JZ SHOW X1
232	DCR A	282	SHOWYZ: CALLY
233	JNZ LOOP11	283	CALL ONLE2
234	MOV E,A	284	CALL YY12
235	STC	285	CALL CEVIR
236	RET	286	SHLD 1B09H
237	OVER: ADC A	287	RET
238	JNC OSU	288	BASLA: CALL ANWEIN
239	INX H	289	JZ BASLA
240	DAD B	290	ANI 0FH
241	DCR E	291	MOV E,A
242	JNZ LOOP1	292	SIFIR: CALL ANVAZL
243	MOV E,A	293	MVI B,80H
244	STC	294	RET
245	RET	295	Z: MVI A,8FH
246	;	296	RET
247	END OF DIV	297	X: MVI A,76H
248	;	298	CALL AUS
249	COORD SUBROUTINE	299	RET
250	;	300	Y: MVI A,6EH

LINE	SOURCE STATEMENT	LINE	SOURCE STATEMENT
501	MVI A,80H	551	LHLD 1B03H
502	LHLD 1B09H	552	SHLD 1B0FH
503	LHLD 1B09H	553	SUB A
504	SHLD 1B0FH	554	STA 1B11H
505	SUB A	555	CALL ADDSUB
506	STA 1B11H	556	LHLD 1B12H
507	LDA 1B14H	557	SHLD 1B0CH
508	LDA 1B14H	558	LDA 1B14H
509	STA 1B39H	559	STA 1B0EH
510	LHLD 1B12H	560	LHLD 1B40H
511	SHLD 1B18H	561	SHLD 1B4CH
512	LHLD 1B00H	562	LHLD 1B42H
513	SHLD 1B0CH	563	SHLD 1B4EH
514	MVI A,80H	564	RET
515	STA 1B0EH	565	:
516	LHLD 1B06H	566	:END OF PARAM
517	SHLD 1B0FH	567	:
518	SUB A	568	: YID SUBROUTINE
519	STA 1B11H	569	:
520	CALL ADDSUB	570	YID: LHLD 1B34H
521	LDA 1B14H	571	SHLD 1B44H
522	LXI H,1B3AH	572	LXI H,0000H
523	MOV M,A	573	SHLD 1B46H
524	DCX H	574	CALL MULT
525	XRA M	575	LDA 1B4BH
526	MOV M,A	576	ANI 80H
527	LHLD 1B12H	577	STA 1B11H
528	XCHG	578	LHLD 1B4AH
529	LHLD 1B18H	579	SHLD 1B0FH
530	DAD D	580	CALL ADDSUB
531	SHLD 1B2EH	581	LHLD 1B12H
532	LHLD 1B18H	582	SHLD 1B30H
533	SHLD 1B50H	583	RET
534	LHLD 1B1CH	584	:
535	SHLD 1B52H	585	:END OF YID
536	CALL DIVI	586	:
537	LHLD 1B28	587	:BITTI SUBROUTINE
538	SHLD 1B40H	588	:
539	LHLD 1B2AH	589	BITTI: CALL ANWAZL
540	SHLD 1B42H	590	MVI B,20H
541	LXI H,0000H	591	MVI A,7FH
542	SHLD 1B46H	592	CALL AUS
543	LHLD 1B00H	593	MVI A,78H
544	SHLD 1B44H	594	CALL AUS
545	CALL MULT	595	MVI A,78H
546	LHLD 1B4AH	596	CALL AUS
547	SHLD 1B0CH	597	MVI A,78H
548	LDA 1B39H	598	CALL AUS
549	XRI 80H	599	MVI A,04H
550	STA 1B0EH	600	RET

LINE	SOURCE STATEMENT	LINE	SOURCE STATEMENT
601	RET	651	OUT 0F9H
602	:	652	CALL DELAY
603	: END OF BITT1	653	CALL INPUT
604	:	654	RET
605	: OKEY SUBROUTINE	655	031: MVI A,08H
606	:	656	OUT 0F9H
607	OKEY:LDA 1B06 H	657	MVI A,0CH
608	LXI H,1B34 H	658	MOV E,A
608	LXI H,1B34 H	658	OUT 0F9H
609	XRA M	659	CALL DELAM
610	RNZ	660	CALL INPUT
611	LDA 1B07 H	661	RET
612	INX H	662	000: CALL COMPAR
613	XRA M	663	000: JNZ 041
614	RNZ	664	042: MVI A,02H
615	DCX H	665	OUT 0F9H
616	DCX H	666	MVI A,03H
617	LDA 1B31H	667	MOV E,A
618	XRA M	668	OUT 0F9H
619	RNZ	669	CALL DELAM
620	DCX H	670	CALL INPUT
621	LDA 1B30H	671	RET
622	XRA M	672	041: MVI A,08H
623	RET	673	OUT 0F9H
624	:	674	MVI A,0CH
625	:END OF OKEY	675	MOV E,A
626	:	676	OUT 0F9H
627	:OUT SUBROUTINE	677	CALL DELAY
628	:	678	CALL INPUT
629	OUT: LXI H,1B39H	679	RET
630	MOV A,M	680	001: CALL COMPAR
631	RLC	681	JNZ 011
632	MOV B,A	682	012: MVI A,02H
633	INX H	683	OUT 0F9H
634	MOV A,M	684	MVI A,03H
635	RLC	685	MOV E,A
636	RLC	686	OUT 0F9H
637	ORA B	687	CALL DELAY
638	STA 1B54H	688	CALL INPUT
639	ANA A	689	RET
640	JZ 000	690	011: MVI A,00H
641	CPI 01H	691	OUT 0F9H
642	JZ 001	692	MVI A,04H
643	CPI 02H	693	MOV E,A
644	JZ 002	694	OUT 0F9H
645	003: CALL COMPAR	695	CALL DELAY
646	JNZ 031	696	CALL INPUT
647	032: MVI A,00H	697	RET
648	OUT 0F9H	698	002: CALL COMPAR
649	MVI A,01H	699	JNZ 021
650	MOV E,A	700	022: MVI A,00H

LINE	SOURCE STATEMENT	LINE	SOURCE STATEMENT
701	OUT OF9H	751	YINCE : LHL 1B32H
702	MVI A,01H	752	INX H
703	MOV E,A	753	SHLD 1B32H
704	OUT OF9H	754	RET
705	CALL DELAY	755	YDEC: LHL 1B32H
706	CALL INPUT	756	DCX H
707	RET	757	SHLD 1B32H
708	021: MVI A,08H	758	RET
709	OUT OF9H	759	STEST: CPI 01H
710	MVI A,0CH	760	RZ
711	MOV E,A	761	ANI 04H
712	OUT OF9H	762	JZ XDEC
713	CALL DELAY	763	XINCE LHL 1B34H
714	CALL INPUT	764	INX H
715	RET	765	SHLD 1B34H
716	COMPAR: LHL 1B30H	766	RET
717	XCHG	767	XDEC: LHL 1B34H
718	LHL 1B32H	768	DCX H
719	MOV A,L	769	SHLD 1B34H
720	SUB E	770	RET
721	RNZ	771	INPUT1: IN OFAH
722	MOV A,H	772	ANI OFH
723	SBB D	773	MOV D,A
724	RET	774	ANI OCH
725	DELAY: LDA 1B38H	775	MOV B,A
726	ANA A	776	LXI H,1B36H
727	MOV C,A	777	MOV A,D
728	CALL DL	778	XRA M
729	MOV A,E	779	ANI 03H
730	ANI 0AH	780	ORA B
731	OUT OF9H	781	MOV M,D
732	MOV A,C	782	RET
733	CALL DL	783	:
734	RET	784	:END OF INPUT
735	DL: MVI B,0DAH	785	:
736	ID1: DCR B	786	ANA: LDA 1B03H
737	JNZ ID1	787	MOV B,A
738	DCR A	788	LDA 1B09H
739	JNZ DL	789	ORA B
740	RET	790	JNZ CMAIN
741	:	791	LDA 1B00H
742	:END OF OUT	792	MOV B,A
743	:	793	LDA 1B06H
744	:INPUT SUBROUTINE	794	ORA B
745	:	795	JZ CMAIN
746	INPUT: CALL INPUT1	796	LHL 1B03H
747	CPI 02H	797	XCHG
748	JZ STEST	798	LHL 1B09H
749	CPI 08H	799	MOV A,L
750	JZ YDEC	800	SUB E

LINE	SOURCE STATEMENT	LINE	SOURCE STATEMENT
801	MOV A,H	851	LDA 1B39H
802	SBB D	852	MOV B,A
803	JNC ONYP	853	ANA A
804 ONYN:	CALL OKEY	854	RAR
805	JZ SOEEN	855	ANA A
806	MVI A,00H	856	RAR
807	OUT OF9H	857	ANA A
808	MVI A,04H	858	RAR
809	OUT OF9H	859	MOV C,A
810	CALL DELAY	860	ANA A
811	MVI A,00H	861	RAR
812	OUT OF9H	862	ADD C
813	CALL DELAY	863	ADD B
814	LHLD 1B32H	864	STA 1B38H
815	DCX H	865	RET
816	SHLD 1B32H	866 BBOL2:	CFI 0E8H
817	JMP ONYN	867	JNC BBOL3
818 ONYP:	CALL OKEY	868	LDA 1B38H
819	JZ SOEEN	869	MOV B,A
820	MVI A,08H	870	ANA A
821	OUT OF9H	871	RAR
822	CALL DELAY	872	ANA A
823	MVI A,08H	873	RAR
824	OUT OF9H	874	ADD
825	LHLD 1B32H	875	STA 1B38H
826	INX H	876	RET
827	SHLD 1B32H	877 BBOL3:	LDA 1B38H
828	JMP ONYP	878	MOV B,A
829 SPED:	LHLD 1B50H	879	ANA A
830	XCHG	880	RAR
831	LHLD 1B52H	881	MOV C,A
832	MOV A,L	882	ANA A
833	SUB E	883	RAR
834	MOV A,H	884	ADD C
835	SBB D	885	ADD B
836	JNC BOL1	886	STA 1B38H
837	SHLD 1B18H	887	RET
838	XCHG	888	END
839	SHLD 1B1CH		
840	CALL DIV1		
841	JMP CBOL1		
842 BOL 1:	SHLD 1B1CH		
843	XCHG		
844	SHLD 1B18H		
845	CALL DIV1		
846 CBOL1:	LDA 1B29H		
847	CFI 80H		
848	RC		
849	CFI 0C8H		
850	JNC BBOL2		

APPENDIX C : OPERATING INSTRUCTIONS OF SYSTEM

- S.1 Turn on the supply voltages.
- S.2 Start the execution of the system program:
 - S.2.1 Press the "R" key and enter Command 7. Give the beginning address (0800H) of the system program and start the execution by pressing on the terminal key, ([↓] or [↑]) [ECB].
 - S.2.2 "H1 =-----.-----" will be displayed on the seven segment displays of the microcomputer.
- S.3 Enter the z coordinate (Z) of the beginning point of the motion.
- S.4 Enter the x coordinate (X1) of the beginning point of the motion.
 - S.4.1 If the entered value of X1 is desired to be changed, press the key "F". This will bring the software to step S.2.2. In other words "H1 =-----.-----" will appear again on the seven segment displays and a new X1 value is expected ones more.
 - S.4.2. After pressing the termination key, the entered value of X1 is stored and "Y=-----.-----" is displayed on the seven segment displays.
- S.5 Enter the y coordinate (Y1) of the beginning point of the motion:
 - S.5.1 Press the "F" key if you want to change the entered Y1 value as in step S.4.1.
 - S.5.2 Press the termination key and obtain "H2=-----.-----" on the displays.
- S.6 Enter the x coordinate (X2) of the destination point of the motion:
 - S.6.1 "F" key is used, if necessary, for changing the entered X2 value.
 - S.6.2 Pressing the termination key displays "Y=-----.-----"

- S.7 Enter the y coordinate (Y2) of the destination point of the motion.
- S.7.1 Press the key "F" if Y2 value is desired to be changed.
- S.7.2 Press the termination key. Then, "H=-----" will appear on the displays.
- S.8 Enter the feedrate.
- S.9 Start the motion.
- S.10 End of motion.
- S.10.1 A "Bitti" message on the seven segment displays will indicate that the motion is completed.
- S.11 Go to step S.2 if cutting is to be proceeded between new coordinates.

APPENDIX D : PHOTOGRAPH OF SYSTEM SET

